

Wireless Control Networks

Modeling, Synthesis, Robustness, Security



George J. Pappas
Joseph Moore Professor
School of Engineering and Applied Science
University of Pennsylvania
pappasg@seas.upenn.edu

Many thanks



PRECISE
PENN RESEARCH IN EMBEDDED COMPUTING AND INTEGRATED SYSTEMS ENGINEERING

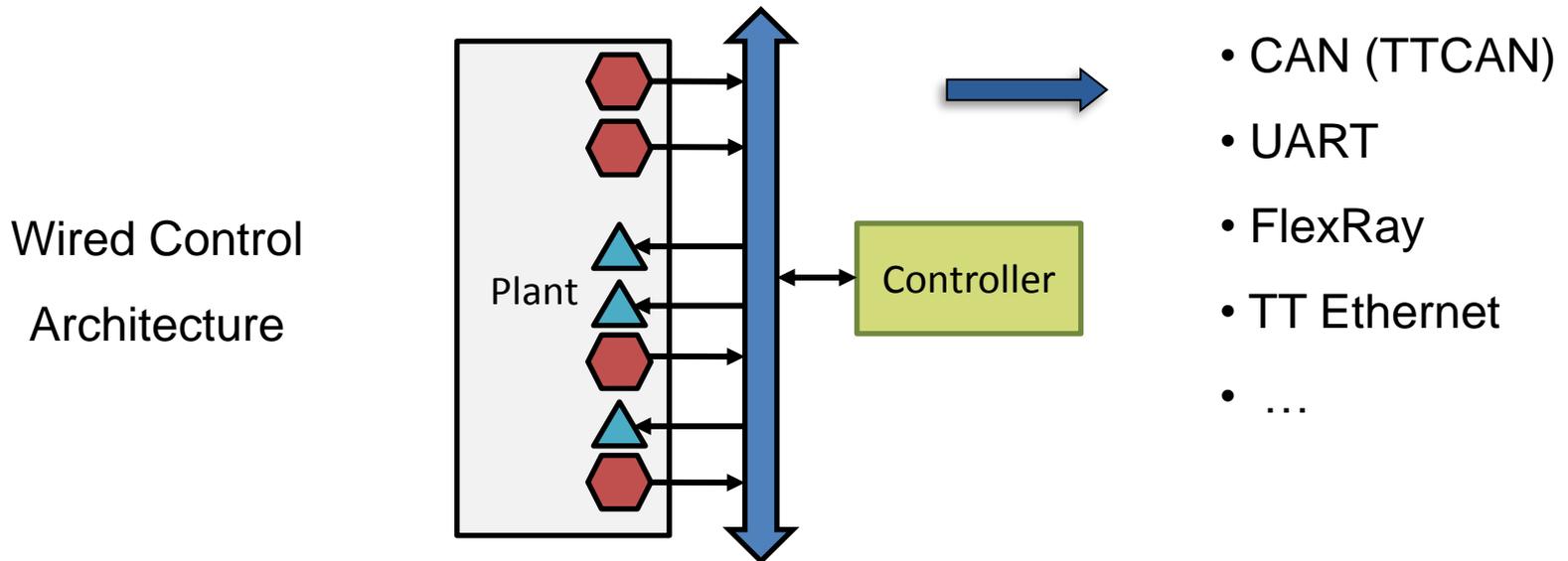
ABB
Honeywell



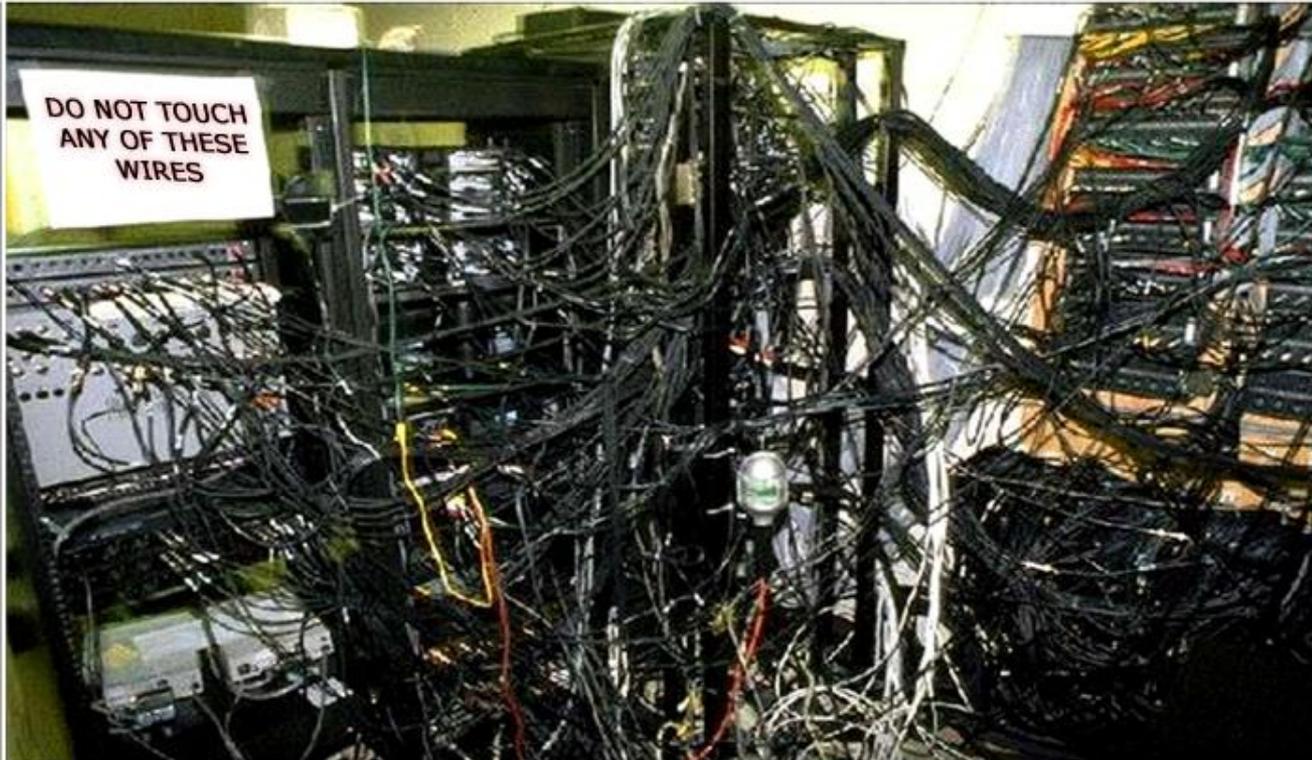
Industrial Control Systems: \$120Billion/Year market



- Sensors () and Actuators () are installed on a plant
- Communicate with controller () over a wired network

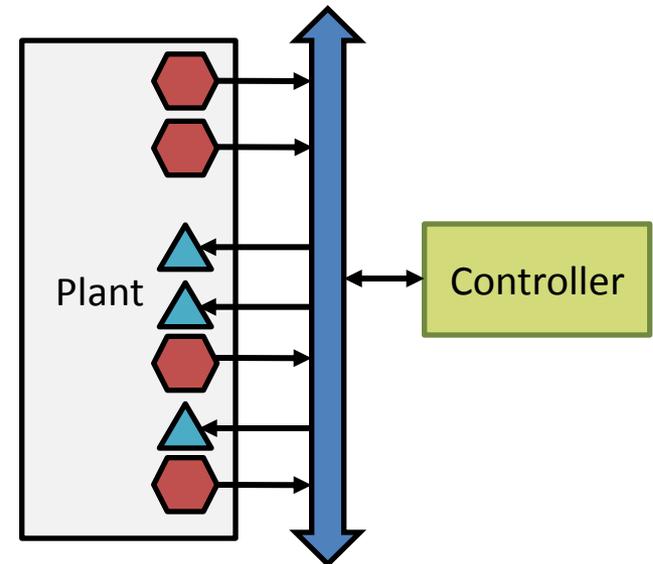
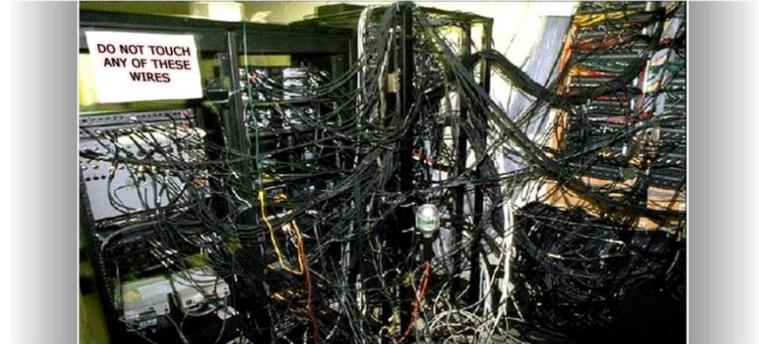


- Control is typically PID loops running on PLC
- Communication protocols are increasingly time-triggered

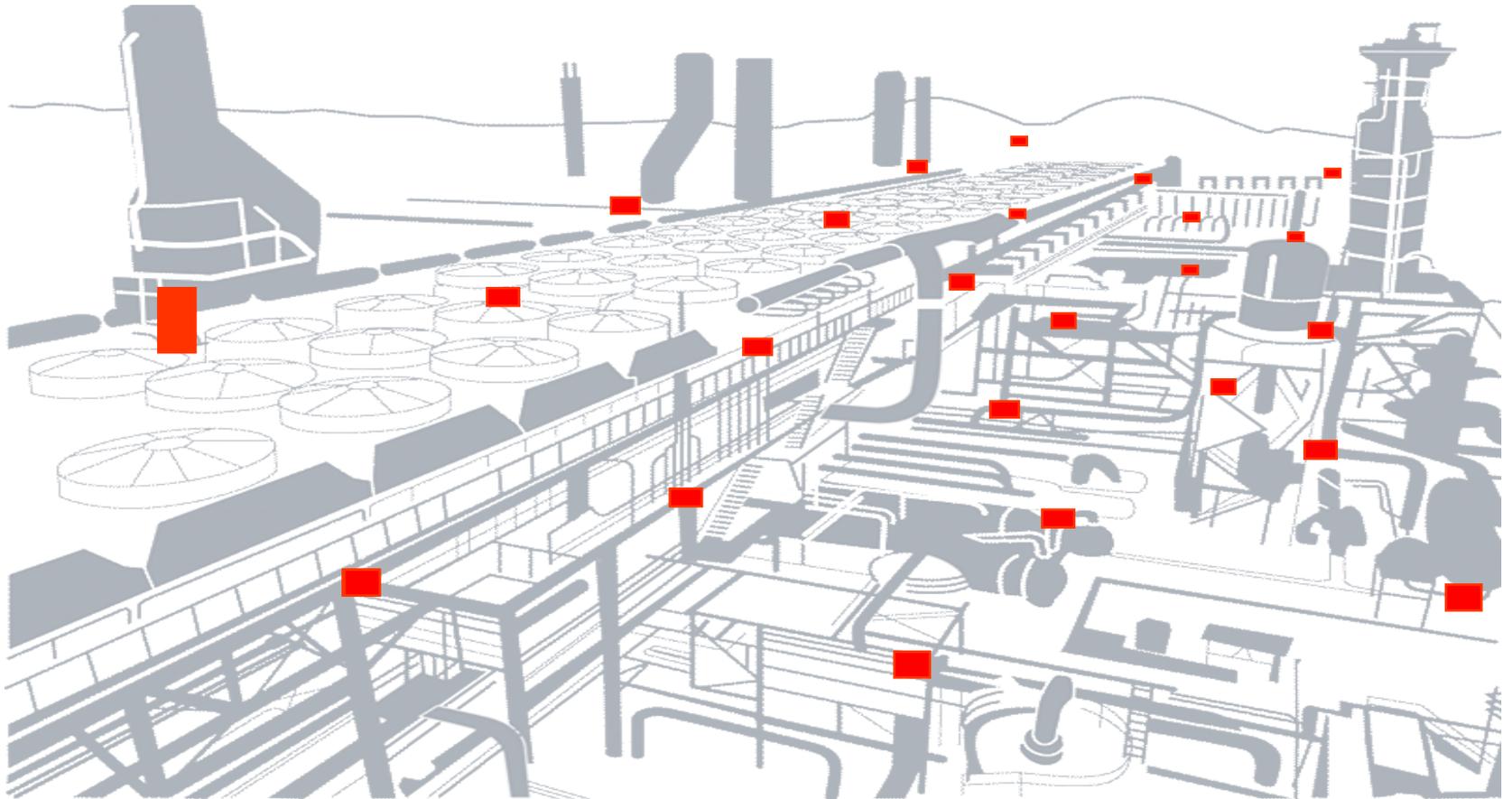


Courtesy of **Honeywell**

- Wires are expensive
 - Wires as well as installation costs
 - Wire/connector wear and tear
- Lack of flexibility
 - Wires constrain sensor/actuator mobility
 - Limited reconfiguration options
- Restricted control architectures
 - Centralized control paradigm

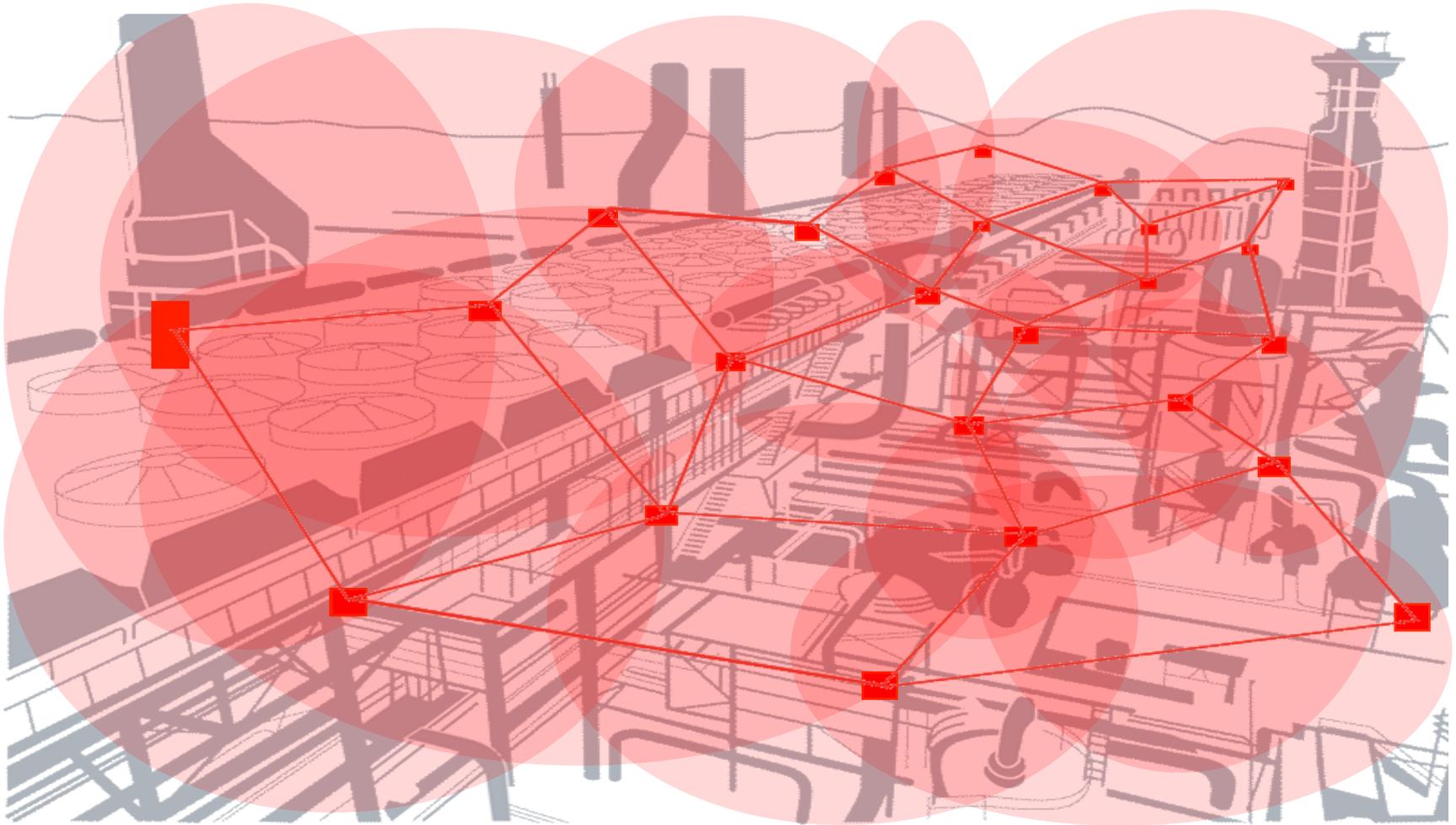


The promise: Wireless Control Systems



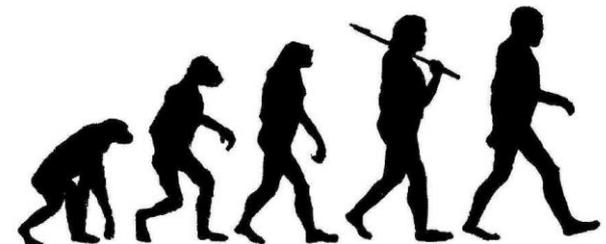
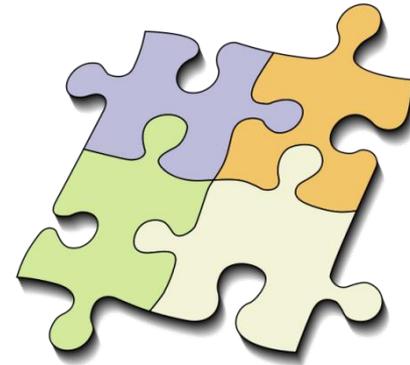
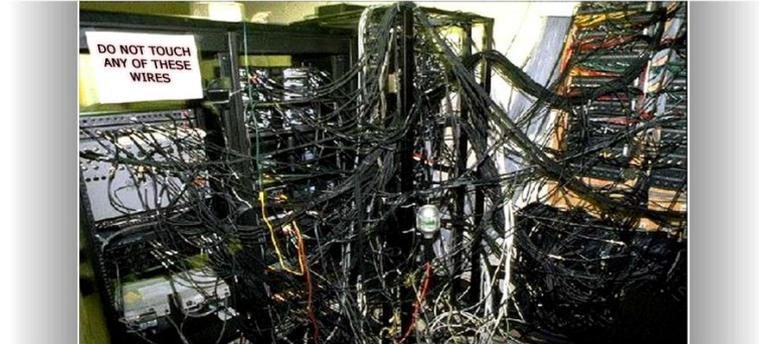
Courtesy of **Honeywell**

The promise: Wireless Control Systems

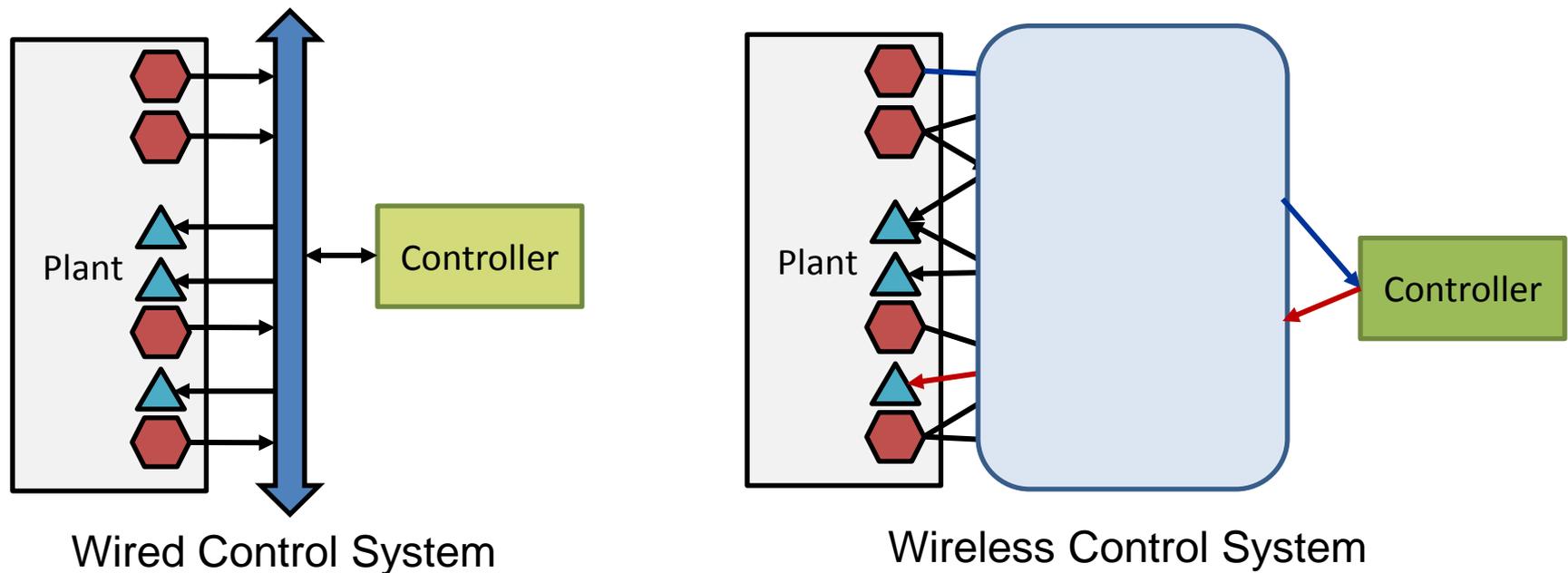


Courtesy of **Honeywell**

- Lower costs, easier installation
 - Suitable for emerging markets
- Broadens scope of sensing and control
 - Easier to sense/monitor/actuate
 - New application domains
- Compositionality
 - Enables system evolution through logical expansion/contraction of plants and controllers with composable control systems.
- Runtime adaptation
 - Control stability and performance are maintained in the presence of node, link and topological changes.

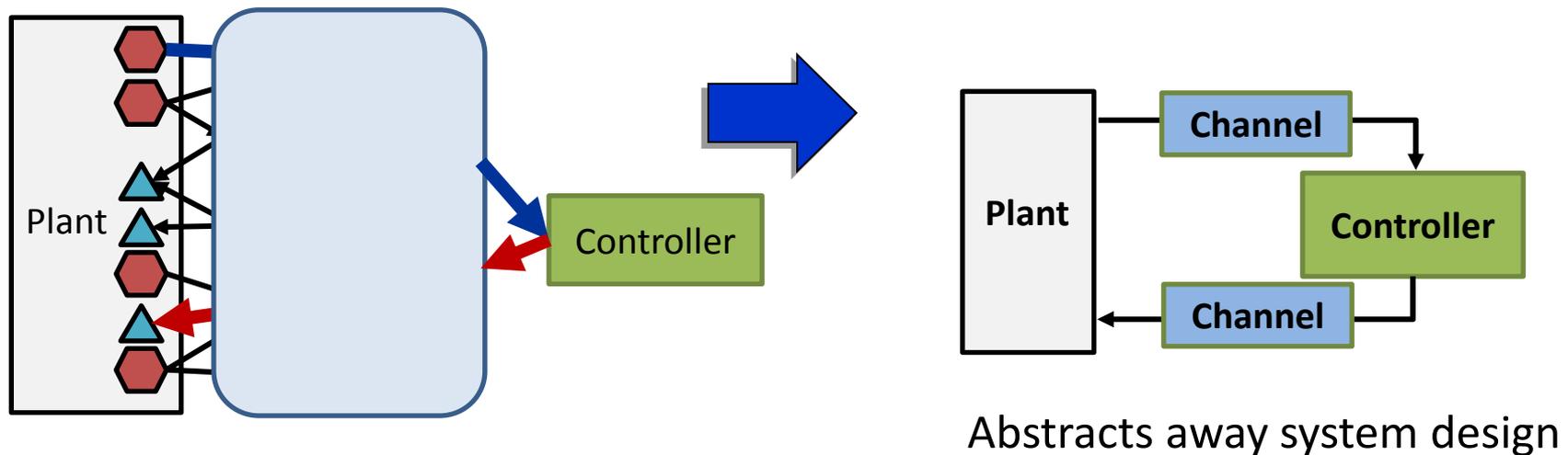


- Paradigm shift towards wireless control architectures



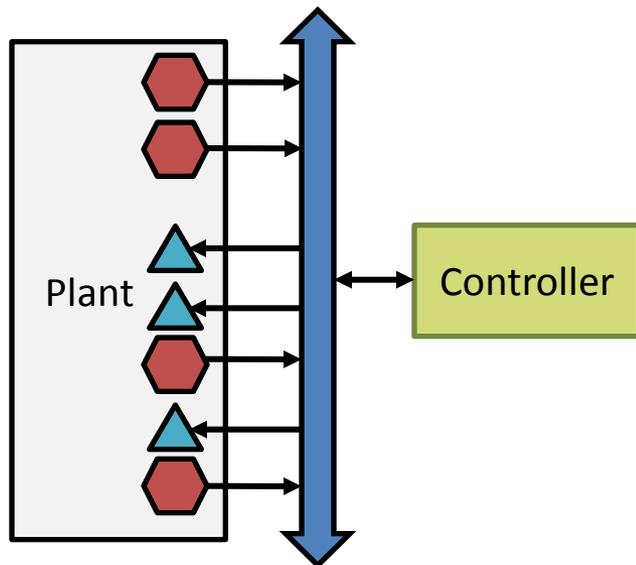
- **Single-hop** and **multi-hop** communication networks

- Standard Wireless Control Systems employ packet routing to deliver information to centralized controllers
 - Control performance depends on the network's QoS

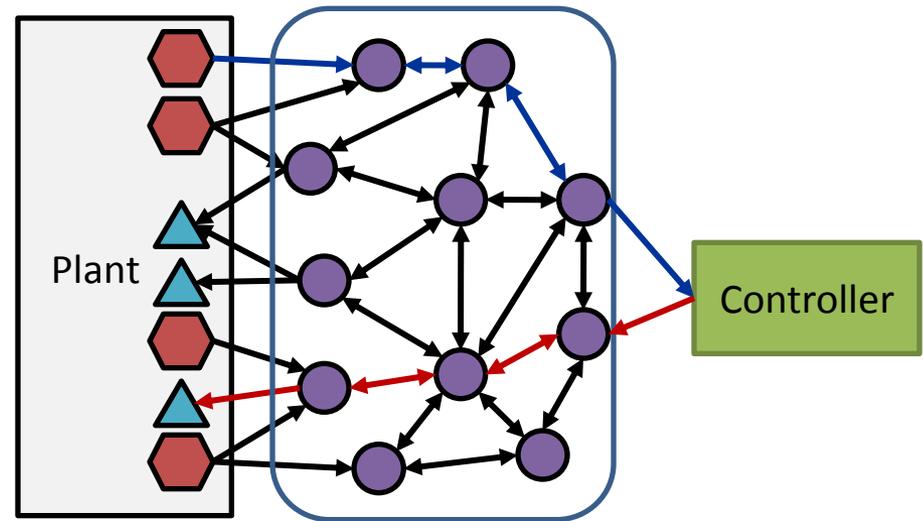


- General challenges include network-induced delay, single-packet vs. multi-packet transmission systems, dropping of communication packets
- Single-hop vs multi-hop networks

- Paradigm shift towards multi-hop control architectures

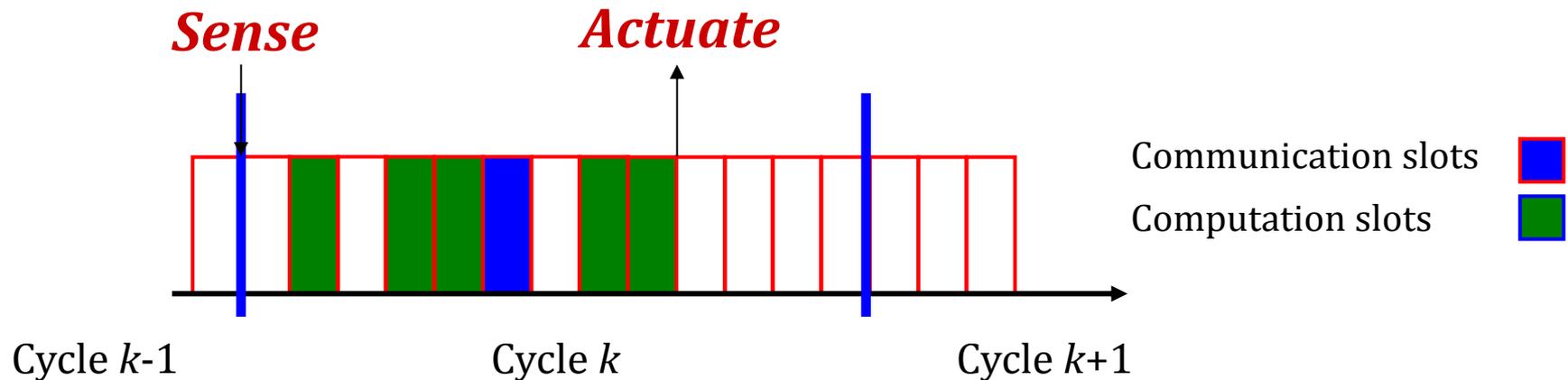


Wired Control System



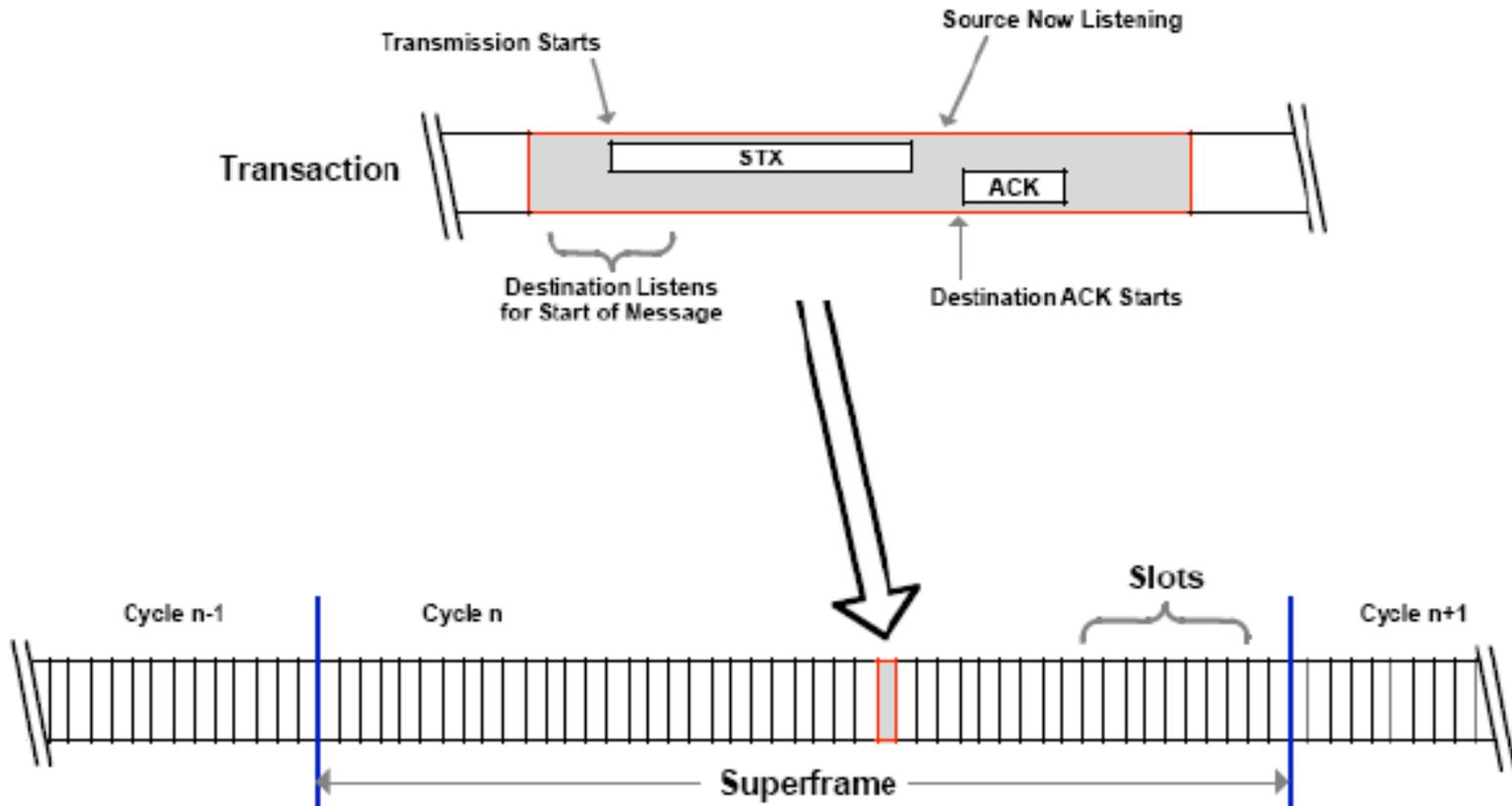
Wireless Control System

- Widely used for time-critical industrial control applications
- Instead of mapping control computation and communication to periodic-tasks, we allocate them to precise time-slots



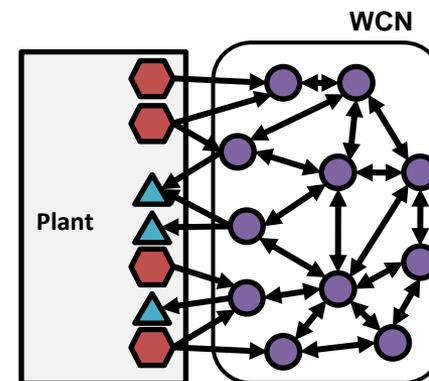
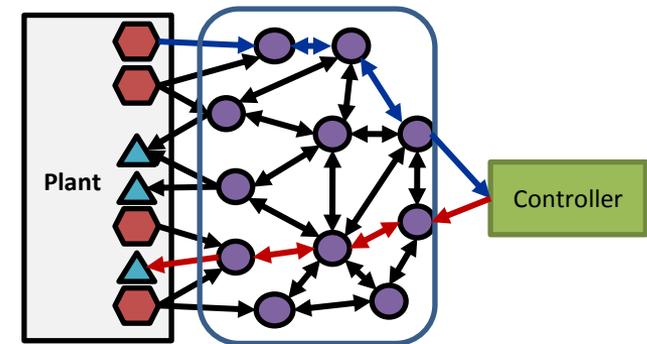
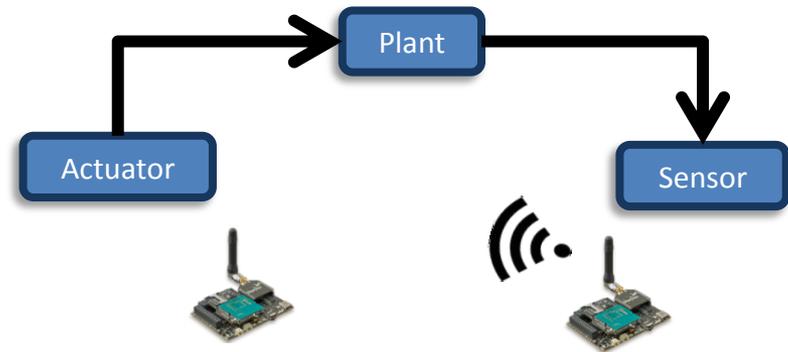
- Wireless time-triggered standards (ISA100, WirelessHART)

- TTA Architecture (TDMA – FDMA), 10ms slots

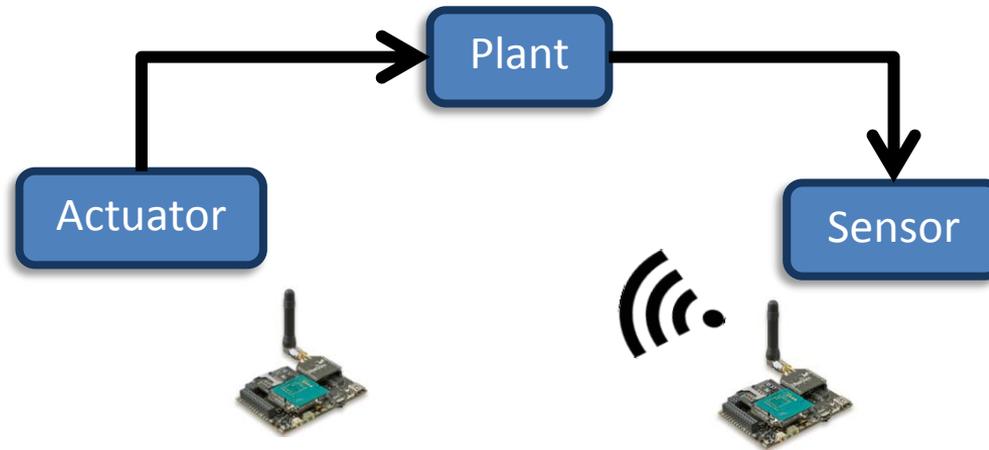


- **Modeling**
 - Holistic modeling of control, communication, computation
 - Interfaces between control and time-triggered communication
- **Analysis**
 - Impact of TDMA-based wireless on control performance
 - Compositional scheduling of multiple control loops
- **Synthesis**
 - Control-scheduling co-design
 - Controller design incorporating TDMA-based properties
 - Network topology design based on physical plant properties
- **Robustness**
 - Robustness analysis with respect to packet losses, node failures
 - Robustness with respect to faulty or malicious nodes

- Optimal Power Management in Wireless Control Systems
 - Power-aware control over single-link networks
- Control with multi-hop wireless networks
 - Routing-based control over time-triggered networks
- Wireless Control Networks
 - A simple decentralized approach for *in-network* control



- Optimal Power Management in Wireless Control Systems*



- Control over a single wireless link
- Separation & optimal plant control
- Optimal and suboptimal communication policies

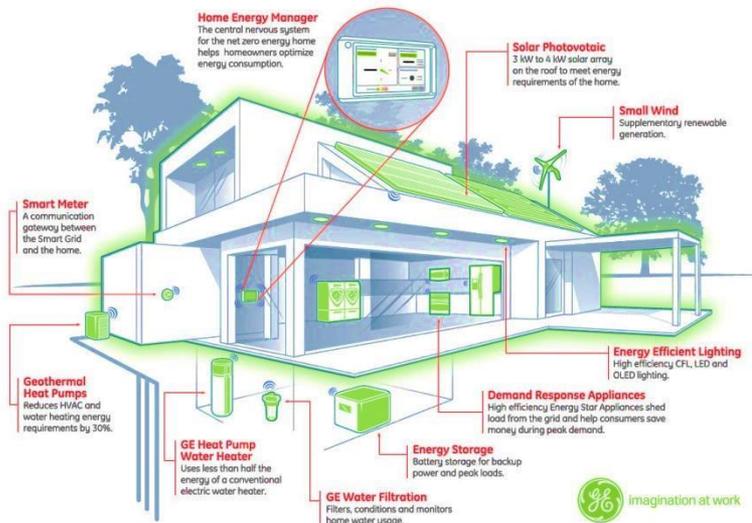
*K. Gatsis, M. Pajic, A. Ribeiro, and G.J. Pappas. *Power-aware communication for wireless sensor-actuator systems*, IEEE Conference on Decision and Control, submitted.

K. Gatsis, A. Ribeiro, G.J. Pappas, *Optimal power management in wireless control systems*, American Control Conference, 2013.

K. Gatsis, A. Ribeiro, G.J. Pappas, *Optimal power management in wireless control systems*. *IEEE Transactions on Automatic Control*, submitted

- Control systems with **power-constrained wireless sensors**, e.g. HVAC, building/industry automation
- Power regulation:  sensor lifetime
- Impact & trade-offs with closed-loop control task

GE Targets Net Zero Energy Homes by 2015

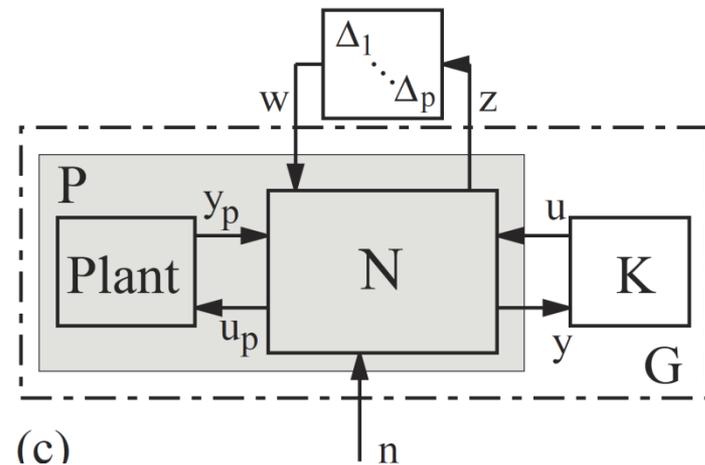
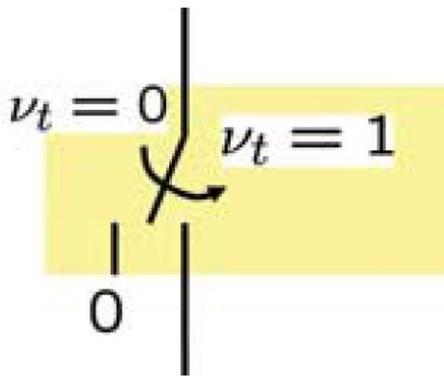


- Common mathematical framework for control/wireless communications
 - unpredictable wireless conditions
 - online power adaptation (PHY layer)
 - timely & reliable information delivery
 - controller design
 - Methodology for (co-) design power & plant control mechanisms
 - Advantages & new insights – in contrast to “control-only” or “communication-only” perspectives
- Communications
- Control

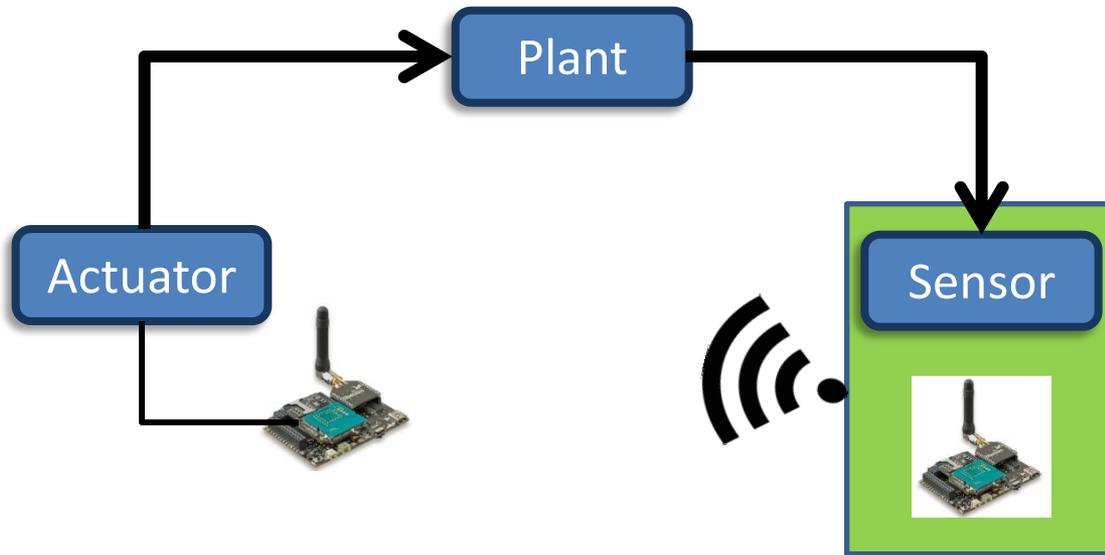
- **Communication as a constraint/disturbance**

- Estimation and Control under packet drops [Hespanha et al 2007], [Sinopoli et al 2004], [Schenato et al 2007], [Gupta et al 2007], [Imer et al 2006]
- Communication as model uncertainty (robust control techniques) [Elia 2005], [Braslavsky et al 2007]

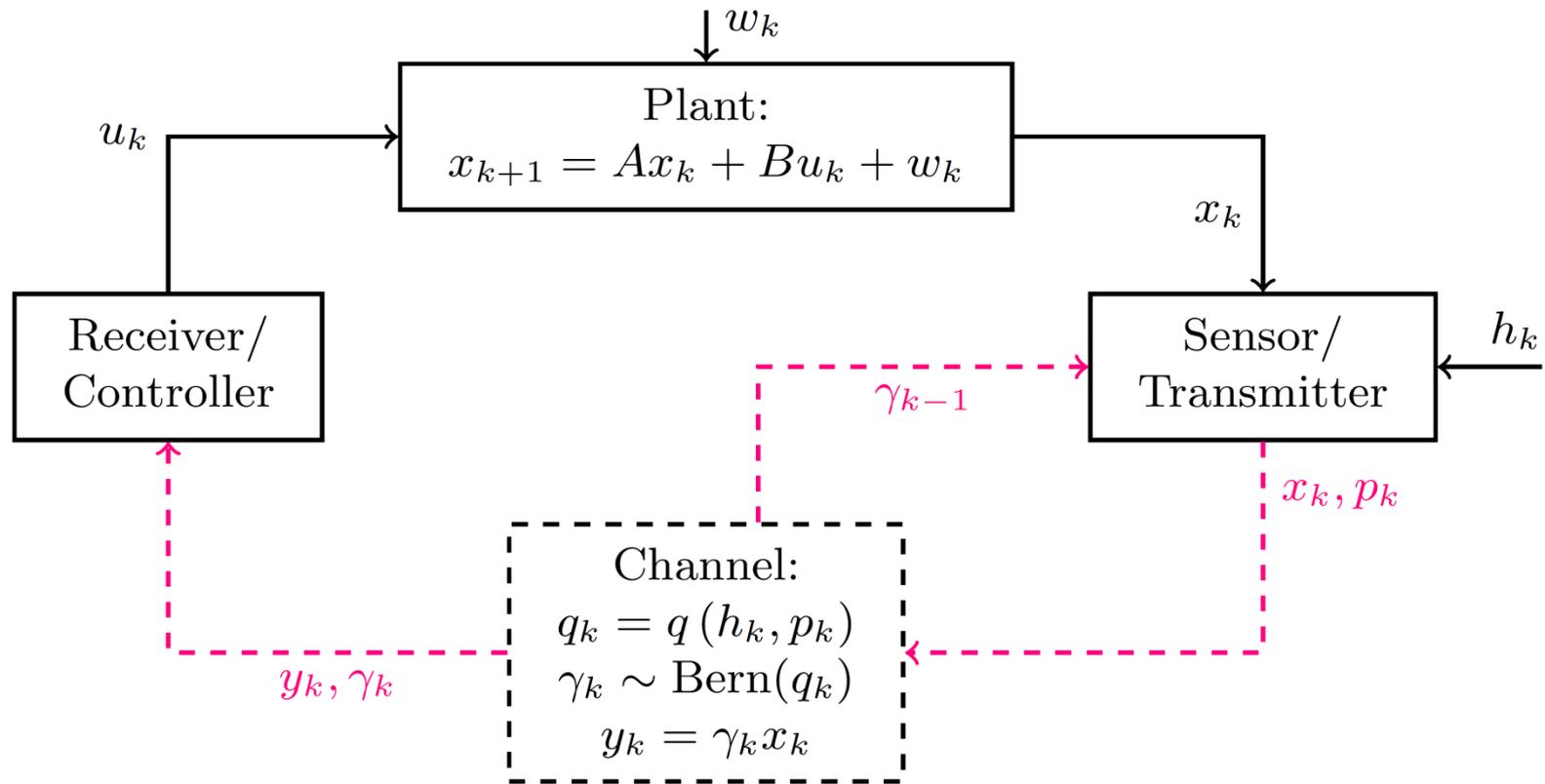
➤ **Communication not part of the design**



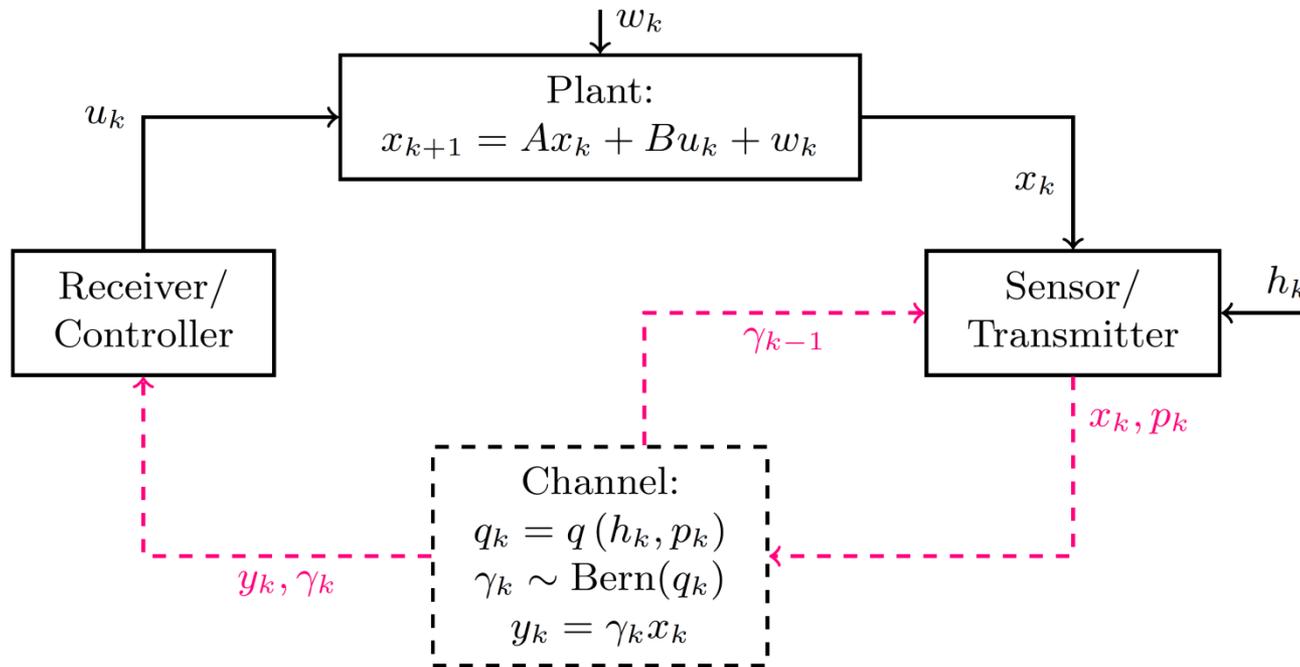
- **Communication with data-rate constraints: coding & control design**
 - [Tatikonda, Mitter 2004], [Nair et al 2007], including power [Quevedo et al 2010]
 - **Communication design: encoding & bit-rate for stability**
- **Event-based paradigm: sensor (actuator) decides whether to transmit (actuate) or not**
 - Estimation [Xu, Hespanha 2004], [Cogill et al 2007], [Mesquita et al 2012], [Li, Lemmon 2011]
 - Control [Tabuada 2007], [Anta, Tabuada 2010], [Rabi, Johansson 2009], [Molin, Hirche 2009], [Donkers et al 2011]
 - **Communication cost: average number of transmissions**



- Single loop with power-constrained sensor/transmitter & power-free receiver/actuator
- Goal: design power control & plant control mechanisms
- **On-line** by adapting to both wireless **channel conditions** and **plant state**
 - Less power when plant 'close' to stability
 - Good channel - cheap to transmit vs. bad channel - costly



- **Channel state** information h_k available at transmitter
- **Power adaptation** p_k to both channel h_k & plant x_k
- Packet drops capture both effects of random wireless channel & protection by power



Decision variables

$$\pi := \{p_0, p_1, \dots\}$$

$$\theta := \{u_0, u_1, \dots\}$$

- Performance: Joint average **linear quadratic** and **power** costs

$$\text{minimize } J(\pi, \theta) := \lim_{N \rightarrow \infty} \frac{1}{N} \sum_{k=0}^{N-1} \mathbb{E}^{\pi, \theta} \left[x_k^T Q x_k + u_k^T R u_k + \lambda p_k \right]$$

Wireless Communication Model

Decoding depends on power and channel



- Received signal-to-noise ratio

$$SNR_k = \frac{h_k p_k}{N_0}$$

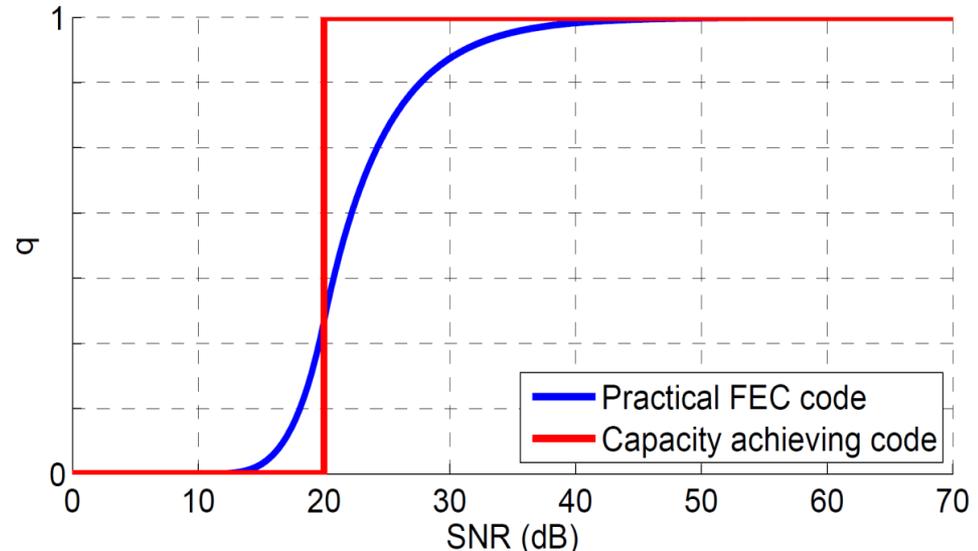
- $p_k \in [0, p_{\max}]$
- h_k block fading, i.i.d.
- N_0 : AWGN power level

- Probability of successful decoding

$$q_k = q(SNR_k)$$

- determined experimentally
- depends on error-correcting code

- Combine in $q_k = q(h_k, p_k)$

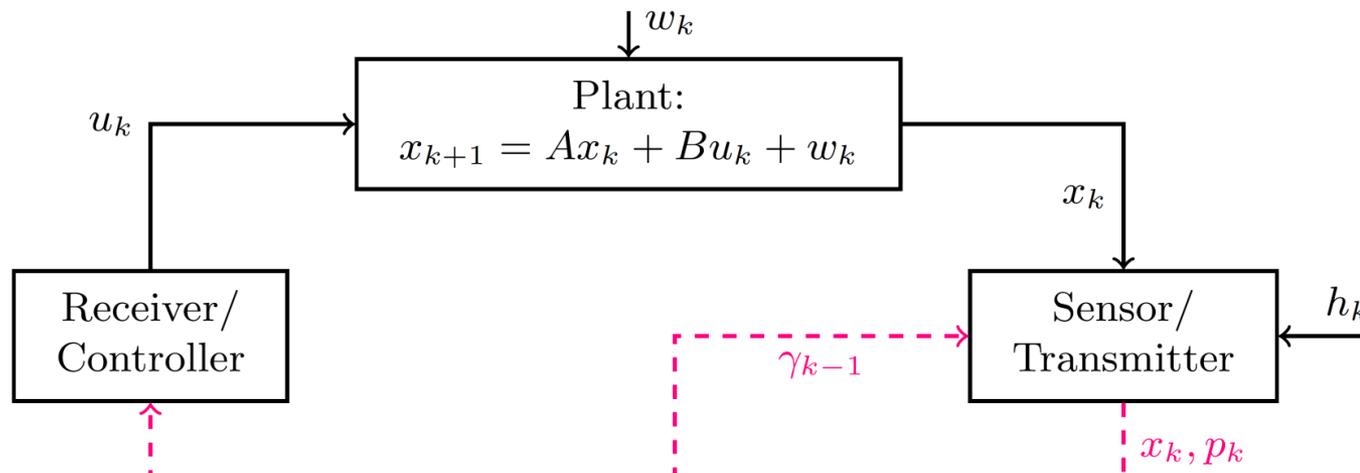


- Generalizes standard Bernoulli packet drops
 - Wireless effects are explicitly captured
 - Bernoulli successes are actively controlled by power
- Generalizes event-triggered transmissions
 - Decision depends also on wireless conditions
 - Communication cost is power consumption vs. transmission rate
- Packet-based communication: unlike data-rate constraints & coding

$$\text{minimize } J(\pi, \theta) := \lim_{N \rightarrow \infty} \frac{1}{N} \sum_{k=0}^{N-1} \mathbb{E}^{\pi, \theta} \left[x_k^T Q x_k + u_k^T R u_k + \lambda p_k \right]$$

- Information structure couples decisions:

Control action u_k affects power decision p_{k+1} through x_{k+1}



- Controller keeps estimate*

$$\hat{x}_k = \begin{cases} x_k & \text{if } \gamma_k = 1, \\ A\hat{x}_{k-1} + Bu_{k-1} & \text{if } \gamma_k = 0 \end{cases}$$

- Innovation terms at sensor/transmitter (known by ACK):

$$\varepsilon_k := x_k - \underbrace{(A\hat{x}_{k-1} + Bu_{k-1})}_{\text{controller's best estimate}}$$

- Restrict available information: innovation and channel

$$\{x_0, h_0, \gamma_0, \dots, \gamma_{k-1}, x_k, h_k\} \rightarrow \{\varepsilon_0, h_0, \dots, \varepsilon_k, h_k\}$$

- Control input does not affect transmitter - **no effect on quality of future plant state estimation**

* Optimal if information from lost packets is removed

- Adapt power to innovation and channel

$$\Pi = \left\{ \begin{array}{l} p_k = p(\varepsilon_k, h_k), \\ \text{such that } \|\varepsilon\| \geq L \Rightarrow p(\varepsilon, h) = p_{max} \end{array} \right.$$

Theorem: For any communication policy $\pi \in \Pi$

$$\min_{\theta} J(\pi, \theta) = Tr(PW) + \underbrace{\lim_{N \rightarrow \infty} \frac{1}{N} \sum_{k=0}^{N-1} \mathbb{E}^{\pi} [e_k^T \tilde{P} e_k + \lambda p_k]}_{J_{COMM}(\pi)}$$

where P is the algebraic Riccati equation solution, $\tilde{P} := A^T P A + Q - P$, and the optimal control is the standard LQR

$$\theta^* : u_k = K \hat{x}_k$$

- Assumptions: (A, B) controllable, $(A, Q^{1/2})$ observable, and for every channel h *

$$q(h, p_{\max}) > 1 - \frac{1}{\lambda_{\max}(A)^2}$$

- relates to stability of the jump estimation errors when transmitter uses full power

$$e_k = \begin{cases} 0 & \text{if } \gamma_k = 1, \\ Ae_{k-1} + w_{k-1} & \text{if } \gamma_k = 0 \end{cases}$$

- guarantees that for any $\pi \in \Pi$

there exists a finite uniform bound $\mathbb{E}^\pi e_k^T e_k \leq M$

* Can be relaxed – in expectation over h

- Finite horizon N - standard LQR Bellman equation & solution

$$V_k := \min_{u_k} \mathbb{E}^\pi \left[x_k^T Q x_k + u_k^T R u_k + V_{k+1} | G_k \right], \quad V_N := 0$$

$$V_k = \mathbb{E}^\pi \left[x_k^T P_{N-k} x_k | G_k \right] + \sum_{j=k}^{N-1} \text{Tr}(P_{N-j-1} W) + \sum_{j=k}^{N-1} \mathbb{E}^\pi \left[e_j^T \tilde{P}_{N-j} e_j | G_k \right]$$

since plant input has no effect on future plant estimates, and $u_k = -K_{N-k} \hat{x}_k$

with standard Riccati recursion

$$K_{k+1} := (R + B^T P_k B)^{-1} B^T P_k A,$$

$$\tilde{P}_{k+1} := A^T P_k B K_{k+1},$$

$$P_{k+1} := A^T P_k A + Q - \tilde{P}_{k+1}, \quad P_0 := 0$$

} Converge by controllability/
observability assumptions

- Limit of finite horizon optimal cost

$$\inf_{\theta \in \Theta} J_{\text{LQR}}^N(\pi, \theta) = \mathbb{E}^\pi V_0 = x_0^T P_0 x_0 + \underbrace{\sum_{k=0}^{N-1} \text{Tr}(P_{N-k-1} W)}_{\text{average converges to } \text{Tr}(PW)} + \underbrace{\mathbb{E}^\pi \sum_{k=0}^{N-1} e_k^T \tilde{P}_{N-k} e_k}_{\text{average converges: } e_k \text{ bounded}}$$

- Optimal communication: estimation vs. power

$$J_{\text{COMM}}^* = \min_{\pi \in \Pi} \lim_{N \rightarrow \infty} \frac{1}{N} \sum_{k=0}^{N-1} \mathbb{E}^{\pi} \left[e_k^T \tilde{P} e_k + \lambda p_k \right]$$

- Reduces to a Markov Decision Process: state (ε, h) , action p
- **Existence** of solution to Bellman equation is shown

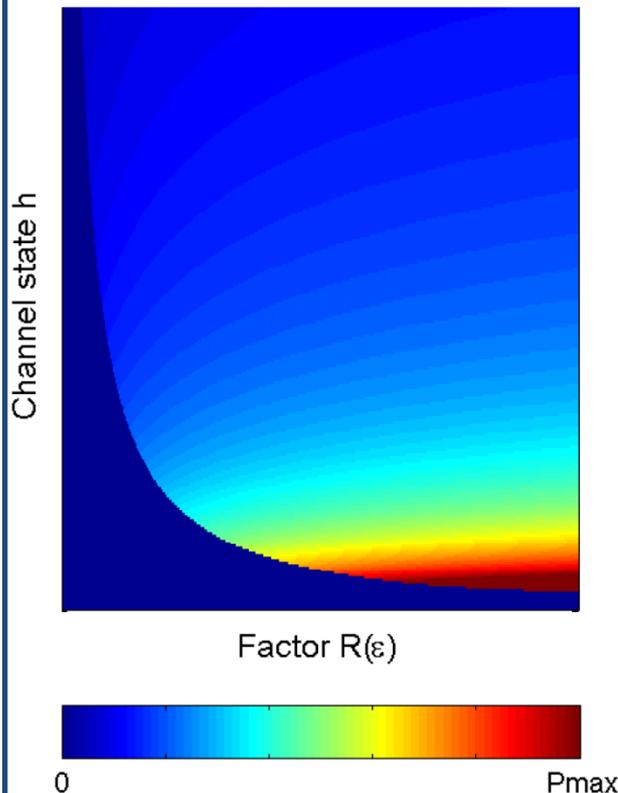
$$\begin{aligned} & V(\varepsilon, h) + J_{\text{COMM}}^* \\ &= \min_{p \in [0, p_{\max}]} (1 - q(h, p)) \varepsilon^T \tilde{P} \varepsilon + \lambda p + \mathbb{E} \left[V(\varepsilon^+, h^+) | \varepsilon, h, p \right] \end{aligned}$$

- **Not computationally tractable** due to continuous state space

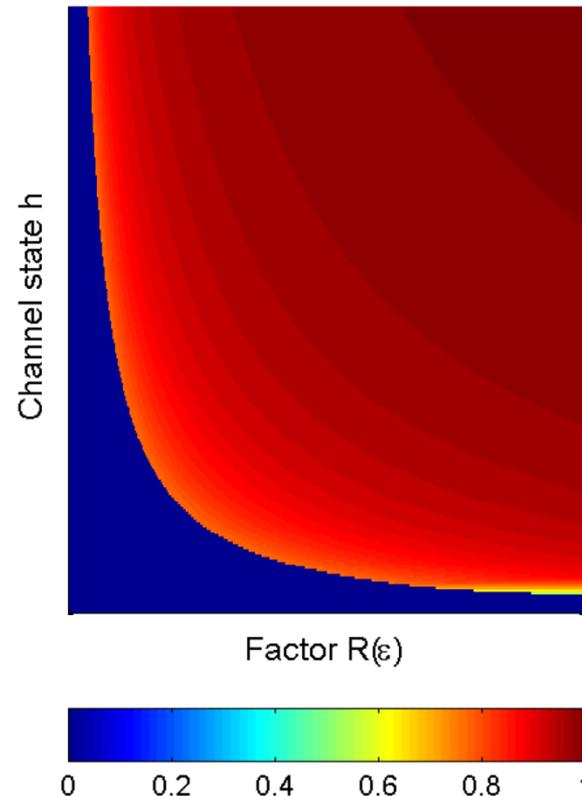
- Optimal power allocation in terms of an unknown penalty on innovation

$$p^*(\varepsilon, h) := \operatorname{argmin}_{p \in [0, p_{\max}]} (1 - q(h, p)) R(\varepsilon) + \lambda p$$

Optimal power allocation p^*



Optimal decoding probability q^*



- Zero power when error small or channel fading low
- Area depends on weight λ
- Outside zero-power region adapt power to both plant and channel
- “Soft” event-triggering

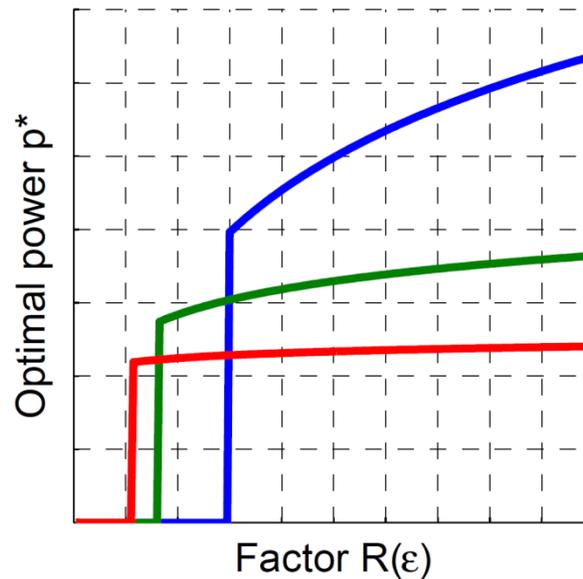
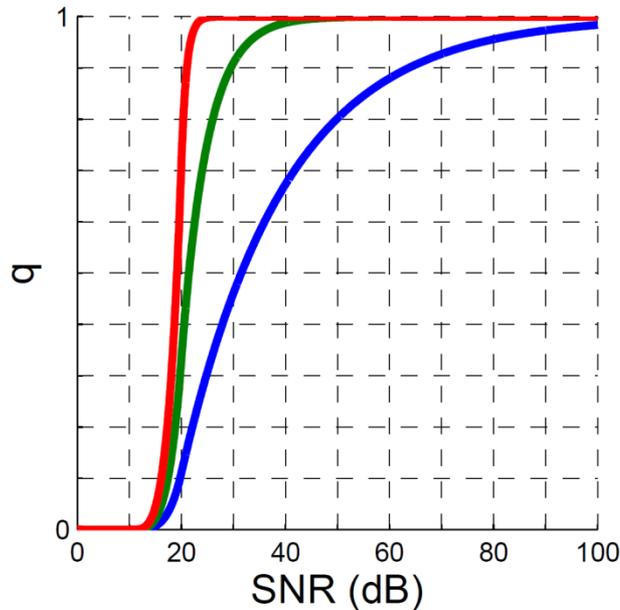
Characterization of Optimal Communication Policy

Dependence on error-correcting code



- Effect of different error-correcting codes

$$p^*(\varepsilon, h) := \operatorname{argmin}_{p \in [0, p_{\max}]} (1 - q(h, p)) R(\varepsilon) + \lambda p$$



- Fixed channel h
- Zero-power region depends on shape of function $q(h, p)$
- “Soft” event-triggering: power adapts to plant when transmitting

- Model capacity achieving codes by indicator

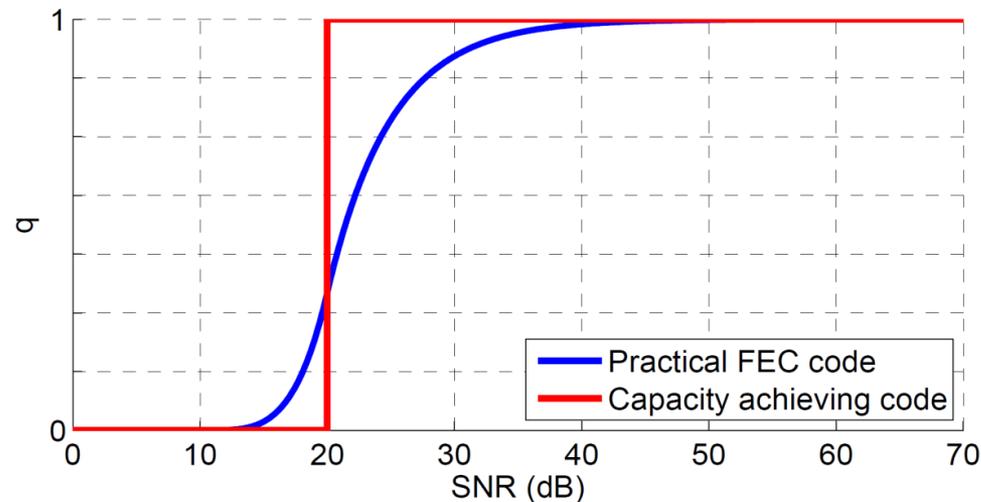
$$q(h, p) = \mathbb{I}(p h \geq p_0)$$

- Optimal (not tractable)

$$p^{CA}(\varepsilon, h) := \begin{cases} 0 & \text{if } h R(\varepsilon) \leq \lambda p_0 \\ p_0/h & \text{otherwise} \end{cases}$$

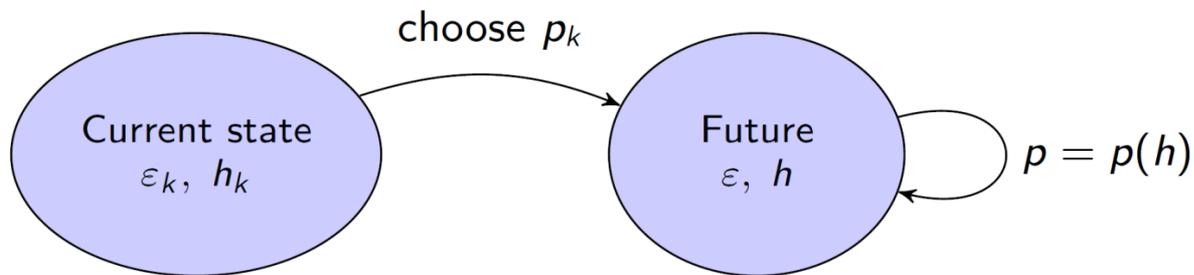
- Packet success $q_k = 0$ or 1

➤ *Recover standard event-triggered transmit-or-not policies, trigger depends on channel and estimation error!*



Rollout policy (model predictive):

“optimize current power as if future policy is some reference”



- Reference policy adapting only to channel $p(h)$
- Bernoulli packet success $\bar{q} := \mathbb{E}_h q(h, p(h))$
- Quadratic cost-to-go $H = (1 - \bar{q})(A^T H A + \tilde{P})$

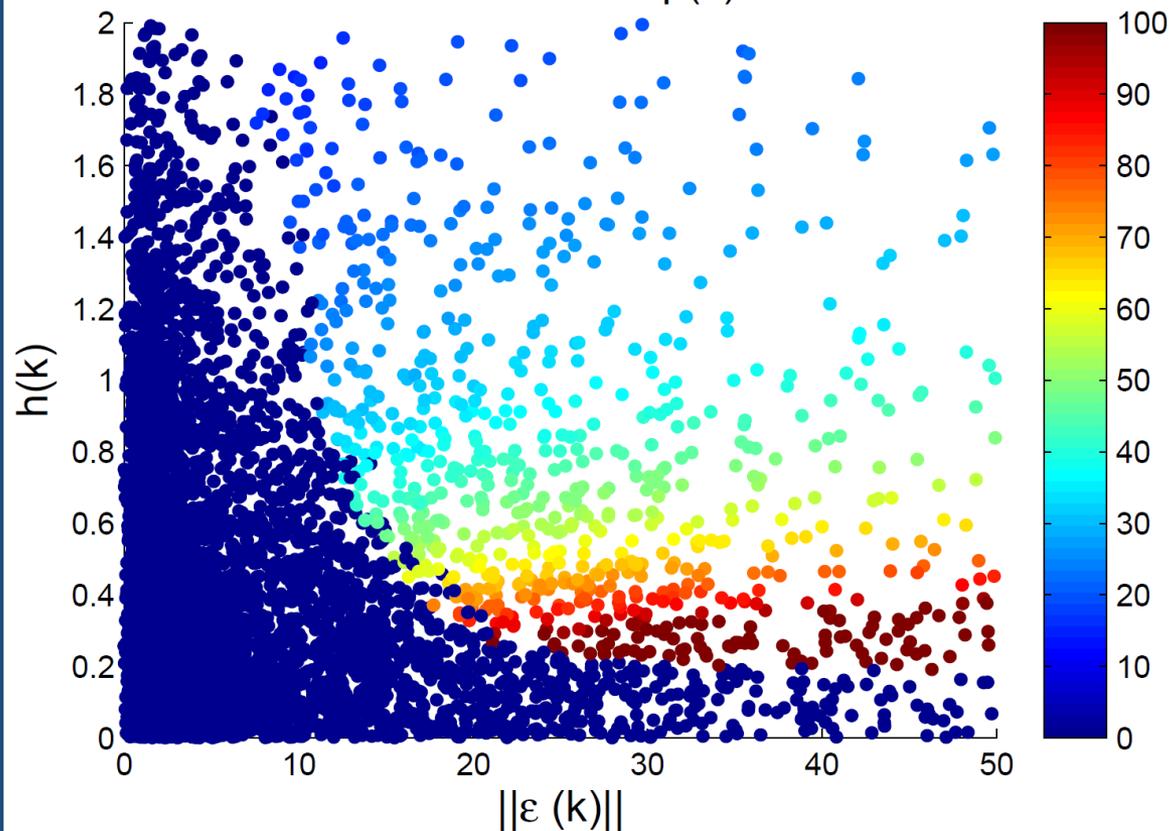
$$p^{\text{roll}}(\varepsilon, h) = \underset{p \in [0, p_{\max}]}{\operatorname{argmin}} (1 - q(h, p)) \frac{\varepsilon^T H \varepsilon}{1 - \bar{q}} + \lambda p$$

Approximates optimal $R(\varepsilon)$

Simulation of Suboptimal Policies – General Codes

$$A = \begin{bmatrix} 2 & 0 \\ 1 & 0.8 \end{bmatrix}, B = \begin{bmatrix} 2 \\ 1 \end{bmatrix}$$

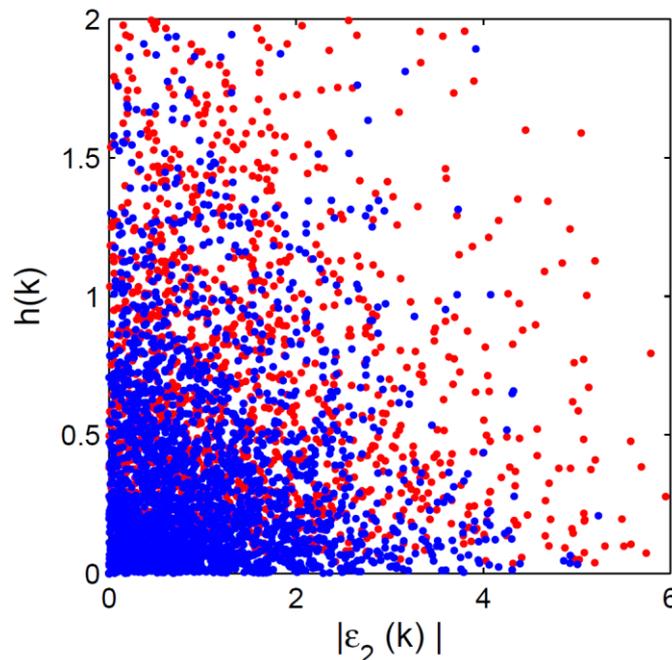
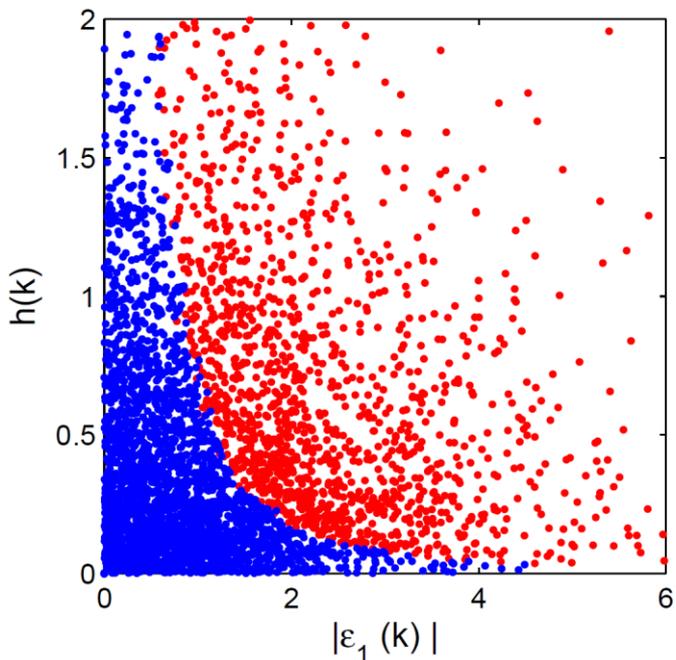
Power choice $p(k)$



- Quadratic penalty on error
- Characteristics similar to the optimal policy

$$A = \begin{bmatrix} 2 & 0 \\ 1 & 0.8 \end{bmatrix}, \quad B = \begin{bmatrix} 2 \\ 1 \end{bmatrix}$$

Blue: don't transmit,
Red: transmit

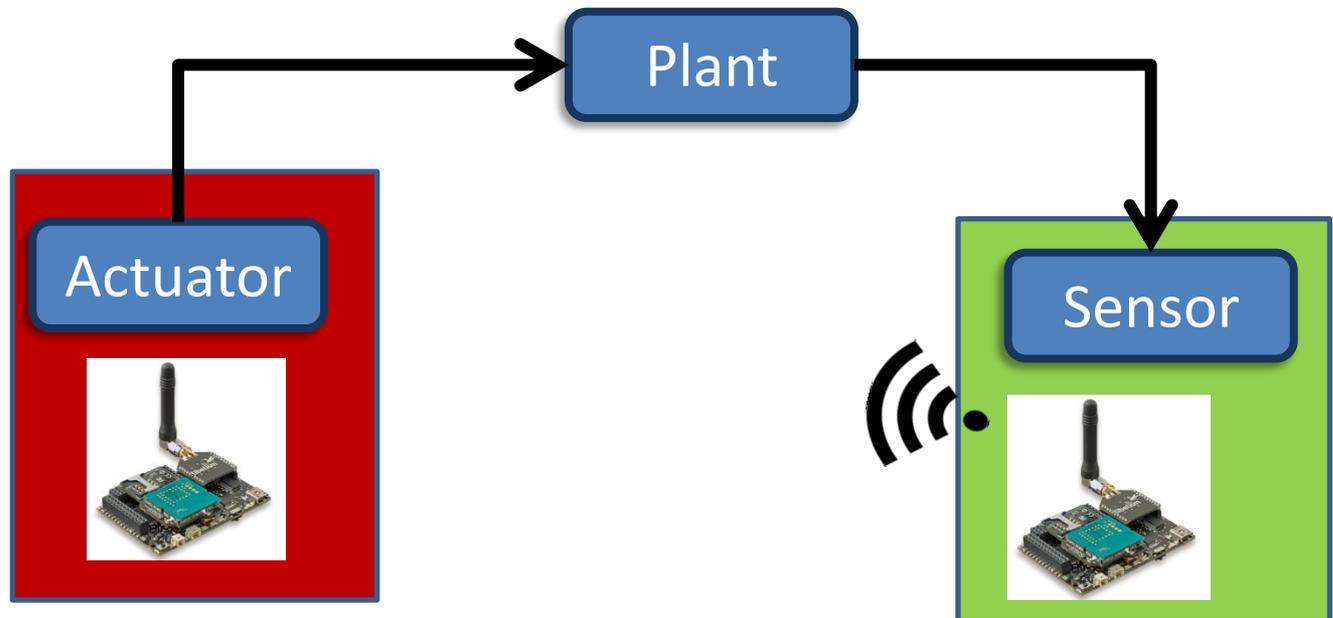


- Policies become event-triggered
- Rollout policy adapts to plant structure

- Richer communication model:
 - captures **uncertainties** of wireless & **power adaptation**
- Communication/control **separation** can be established (suboptimal but otherwise joint cost hard to analyze)
- Optimal communication is ‘soft’ event-triggered
 - zero power if error small or channel adverse
 - power **adaptation to both plant and channel** states otherwise
- Communication policies can be designed by ADP techniques

- **Model, analysis, communication/control co-design of complex wireless sensor & actuator networks**
 - Multiple or distributed plants
 - Shared wireless channels (interference)
 - Optimal **control-aware resource allocation**, e.g. power, scheduling
 - Economic **resource-aware controller synthesis**

- Architecture limitation:
 - wireless receiver/controller *always listens*
 - comparable power consumption at both ends
 - common in any event-based scheme over wireless





- Ideally: Turn off receiver *between transmissions...*
 - inconsistent with event-triggering
- Our approach: **coordination protocol**
 - Devices turn off and agree on next wake-up time (self-triggered* step)
 - Upon wake-up sensor decides whether to transmit or not (event-triggered step)
 - How to '*predict*' when next event will occur?
 - Consider power costs at both ends, current channel & plant states

*[Anta, Tabuada 2010]

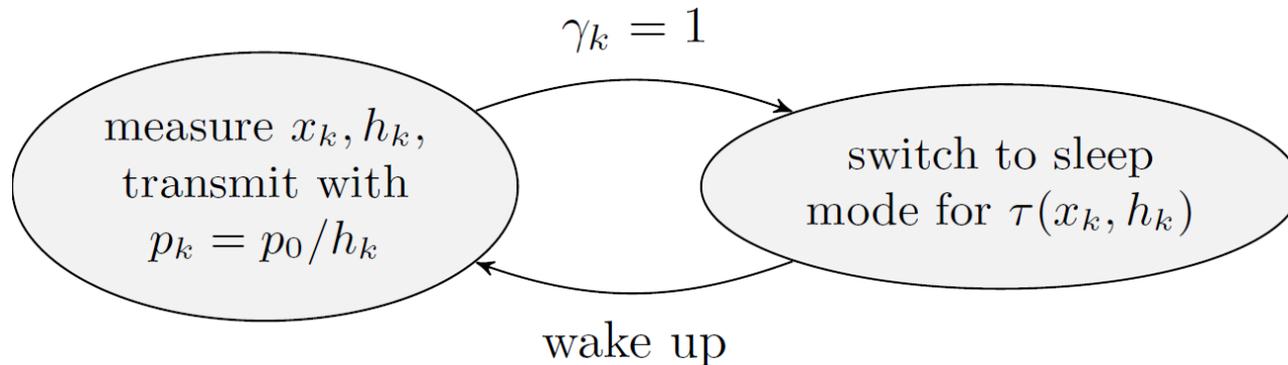
- **Markov fading channel (finite states, irreducible, aperiodic)**
 - possibility of *predicting* good channels
- Capacity achieving code
- **Constant power penalty p_a for awake receiver, p_k for transmitter as before**

$$p'_k := \begin{cases} p_a + p_k & \text{if awake at } k, \\ 0 & \text{if in sleep mode at } k \end{cases}$$

- Fixed LQR controller $u_k = K \hat{x}_k$
- Trade-off estimation error vs. power at both ends

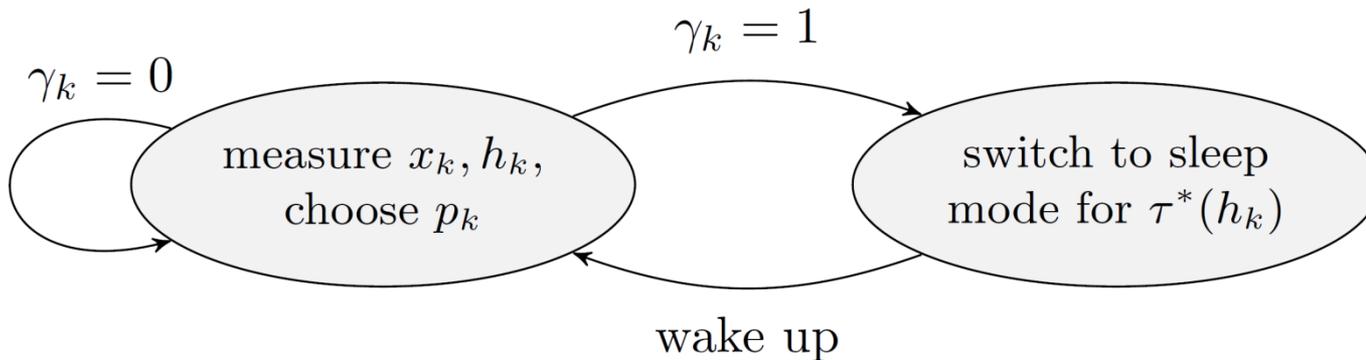
$$J := \lim_{N \rightarrow \infty} \frac{1}{N} \sum_{k=0}^{N-1} \mathbb{E} [e_k^T P e_k + \lambda p'_k]$$

- Self-triggered protocol:



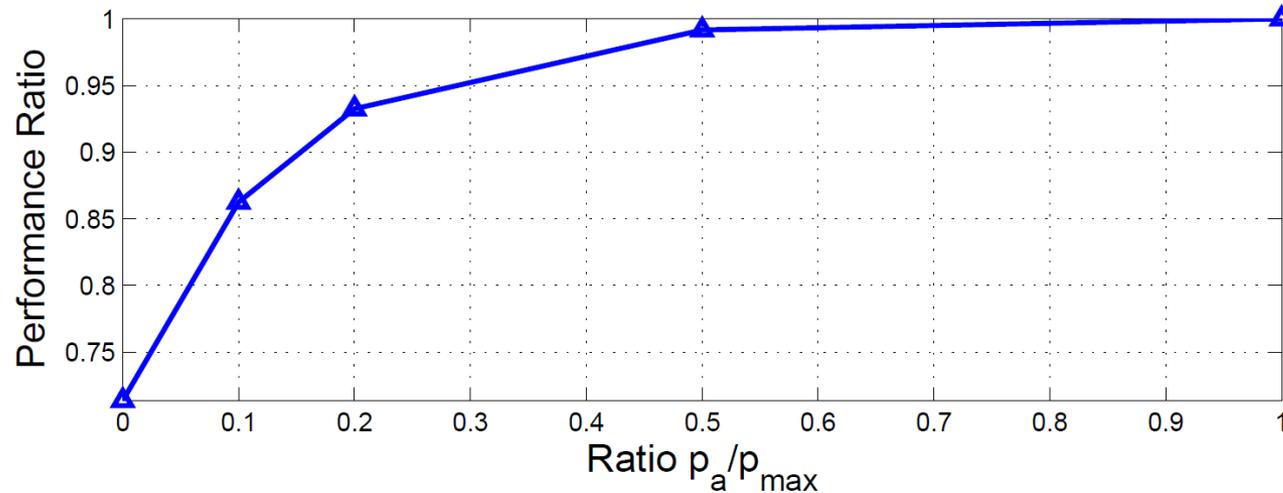
- Cost **independent of plant state** : estimation error is reset on every transmission
- Sleep-time need only **depend on channel state** : predict when channel suitable and estimation error not too large
- **Optimal** computed by analogy to a MDP (tractable for finite channel states)

- Proposed protocol – model predictive



- **Upon wake-up decide** whether to transmit & sleep according to optimal self-triggered, or skip current step
- Current decision based on **modeling future behavior & cost**
- Guaranteed to perform **not worse than** optimal self-triggered
- Injects event-triggered steps between sleep

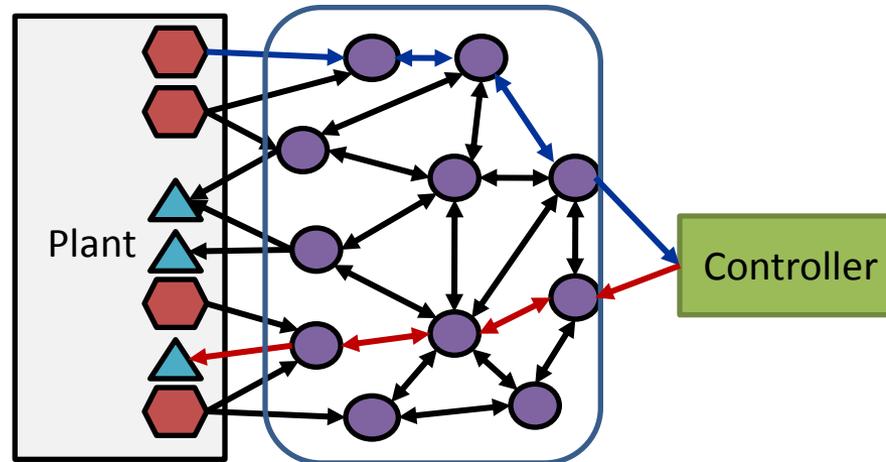
- Ratio of proposed protocol / optimal self-triggered as receiver's constant power increases



- If power for receiver to stay awake dominates power for transmitter to communicate, self-triggered performs best

- New Paradigm for Control/ Wireless Networking
 - Model capturing explicitly **wireless fading channel** effects and **power allocation & interaction with control** task
 - Novel Physical Layer design: Characterization of **optimal power adaptation** to channel & plant conditions
 - **Receiver** power considerations via a coordination protocol
- Future work
 - **Medium Access Control** for multiple closed-loops over a shared wireless channel
 - **Control-aware Resource Allocation**, e.g. scheduling, power, in wireless networked control systems

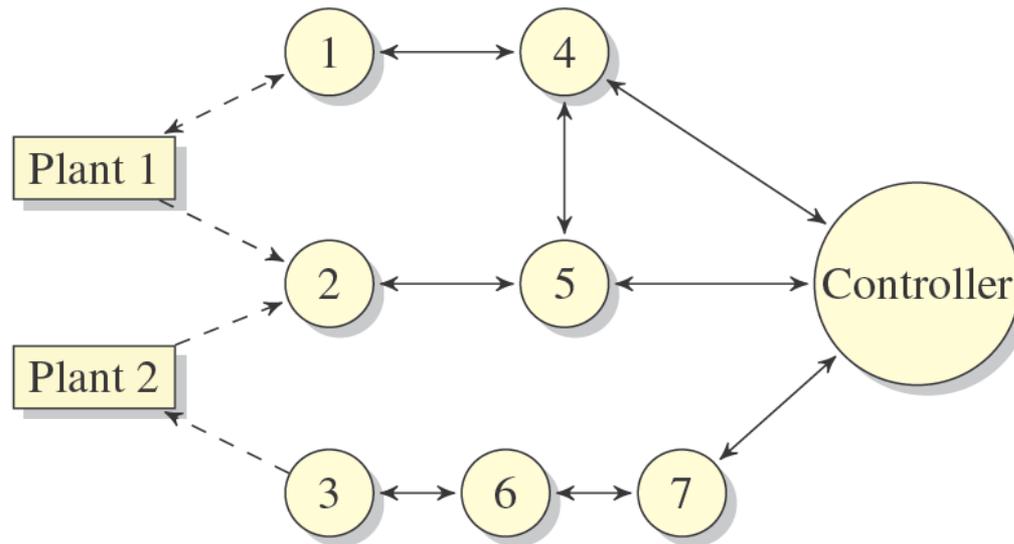
- Control with multi-hop wireless networks*



- **Formal modeling**
- Analysis & synthesis
- Compositional analysis
- Industrial case study

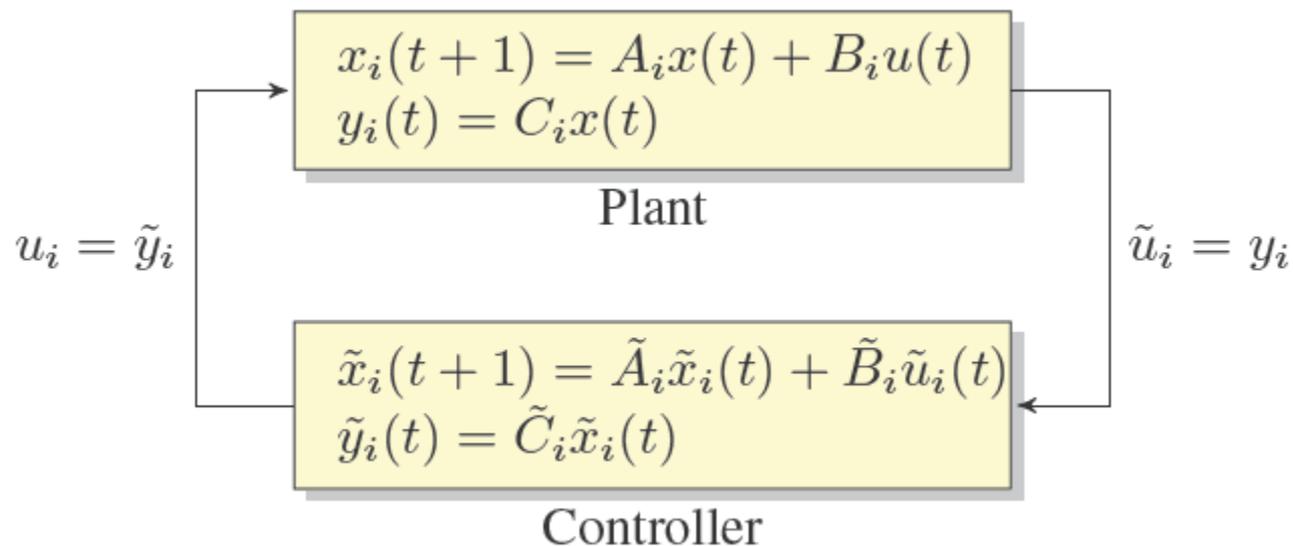
*R. Alur, A. D’Innocenzo, K.H. Johansson, G. Pappas, G. Weiss *Compositional modeling and analysis of multi-hop networks*, IEEE Transactions on Automatic Control, October 2011

- A multi-hop wireless networked system



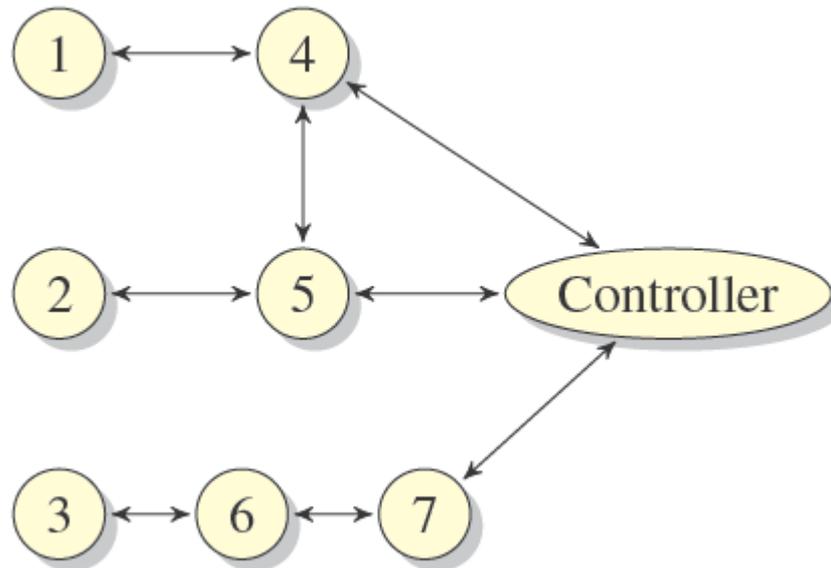
- Assumptions:
 - Plants/controllers are discrete-time linear systems
 - Multi-hop network runs time-triggered protocol

- Plants/controllers are discrete-time linear systems

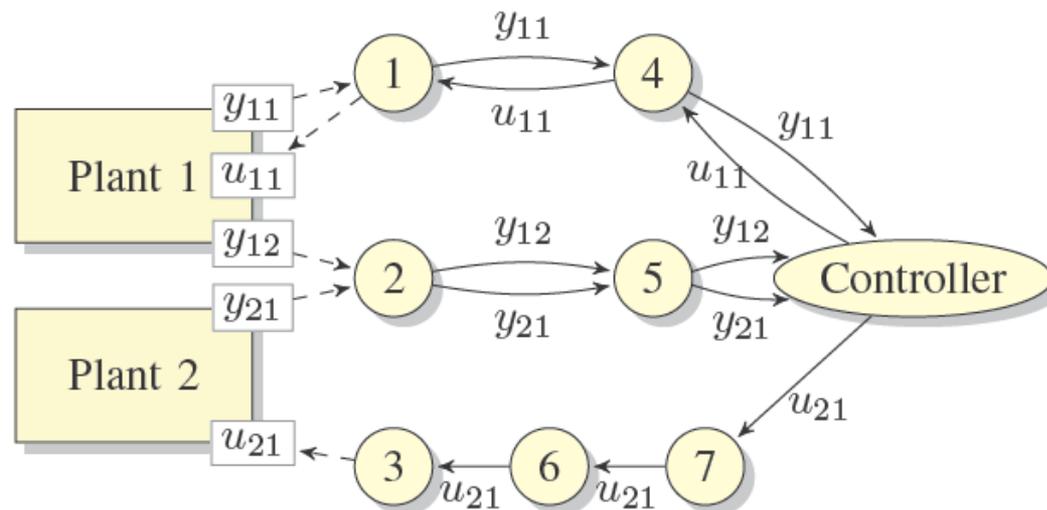


- Controllers are designed to achieve suitable performance

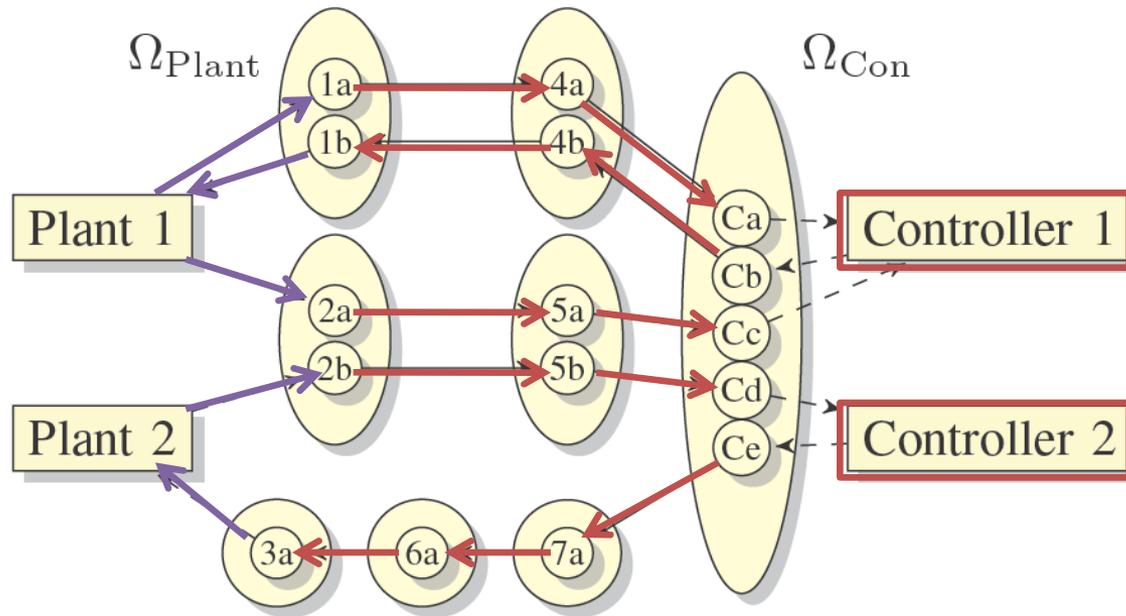
- Plants/controllers are discrete-time linear systems
- Graph $G = (V, E)$ where V is the set of nodes and E is the radio connectivity graph



- Plants/controllers are discrete-time linear systems
- Graph $G = (V, E)$ where V is the set of nodes and E is the radio connectivity graph
- Routing $R : I \cup O \rightarrow 2V^* \setminus \{\emptyset\}$ associates to each pair sensor-controller or controller-actuator a set of allowed routing paths



Communication and computation schedule



Communication schedules: $\eta: \mathbb{N} \rightarrow 2^{E \times (\mathbb{I} \cup \mathbb{O})}$

Computation schedules: $\mu_i: \mathbb{N} \rightarrow \{\text{Idle}, \text{Active}\}$

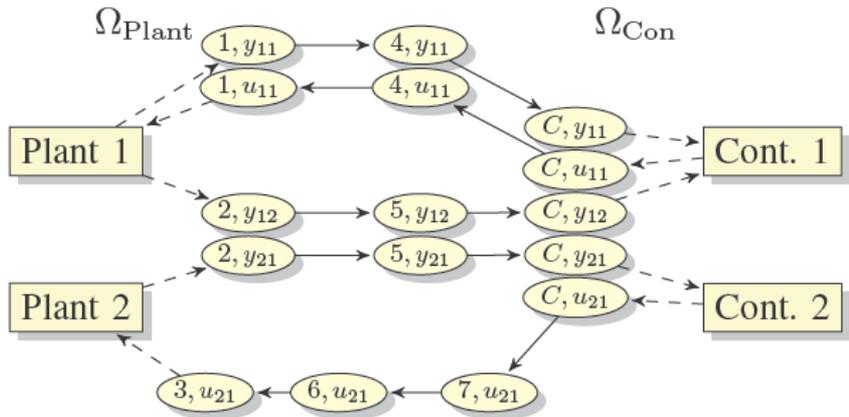
1a,4a | 2a,5a | 4a,Ca | 5a,Cc | 2b,5b | 5b,Cd | Cb,4b | 4b,1b | Ce,7a | 7a,6a | 6a,3a | ...

Communication schedule

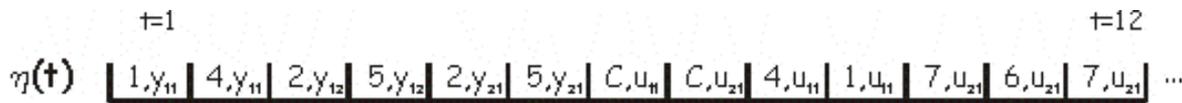
| | | | Cont 1 | Cont 2 | | | | ...

Computation schedule

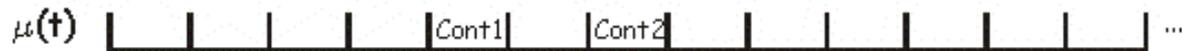
Evolution in each time step



$$T[\eta(t), \mu(t)] = \begin{pmatrix} A_{\text{Plant}} & B_{\text{Plant}} (0 \ 0 \ 1 \ 0) & 0 \\ \begin{pmatrix} 1 \\ 0 \\ 0 \\ 0 \end{pmatrix} C_{\text{Plant}} & \text{Adjacency matrix of } \eta(t) & \begin{pmatrix} 0 \\ 0 \\ 0 \\ 1 \end{pmatrix} C_{\text{Controller}}^{\mu(t)} \\ 0 & B_{\text{Controller}}^{\mu(t)} (0 \ 1 \ 0 \ 0) & A_{\text{Controller}}^{\mu(t)} \end{pmatrix}$$



communication schedule



computation schedule

Given communication and computation schedules, the closed loop multi-hop control system is a switched linear system

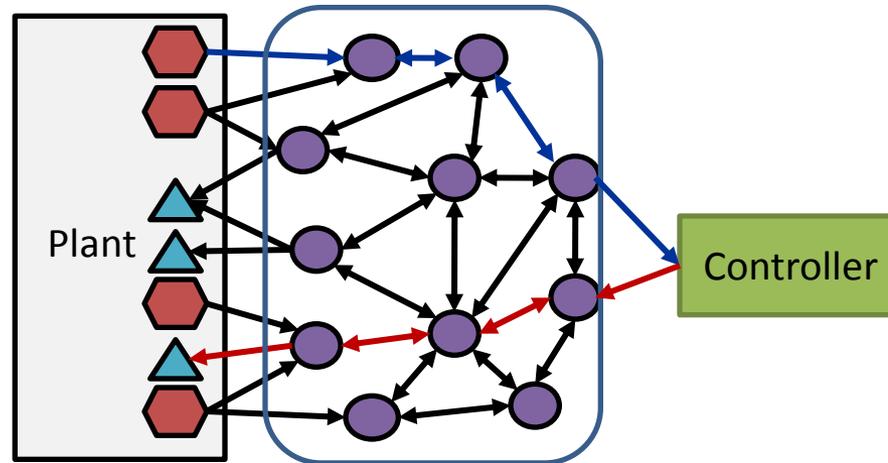
$$x(t+1) = T[\eta(t), \mu(t)]x(t)$$

where the schedule (discrete switching signal) is either:

1. Deterministic and periodic
2. Nondeterministic and periodic
3. Stochastic due to packet loss, failures

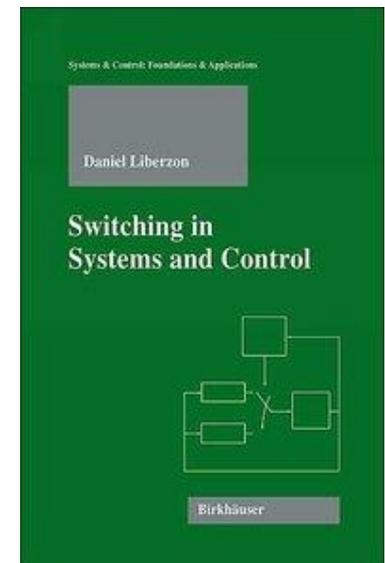
Modeling the multi-hop control network as a hybrid system!

- Control with multi-hop wireless networks

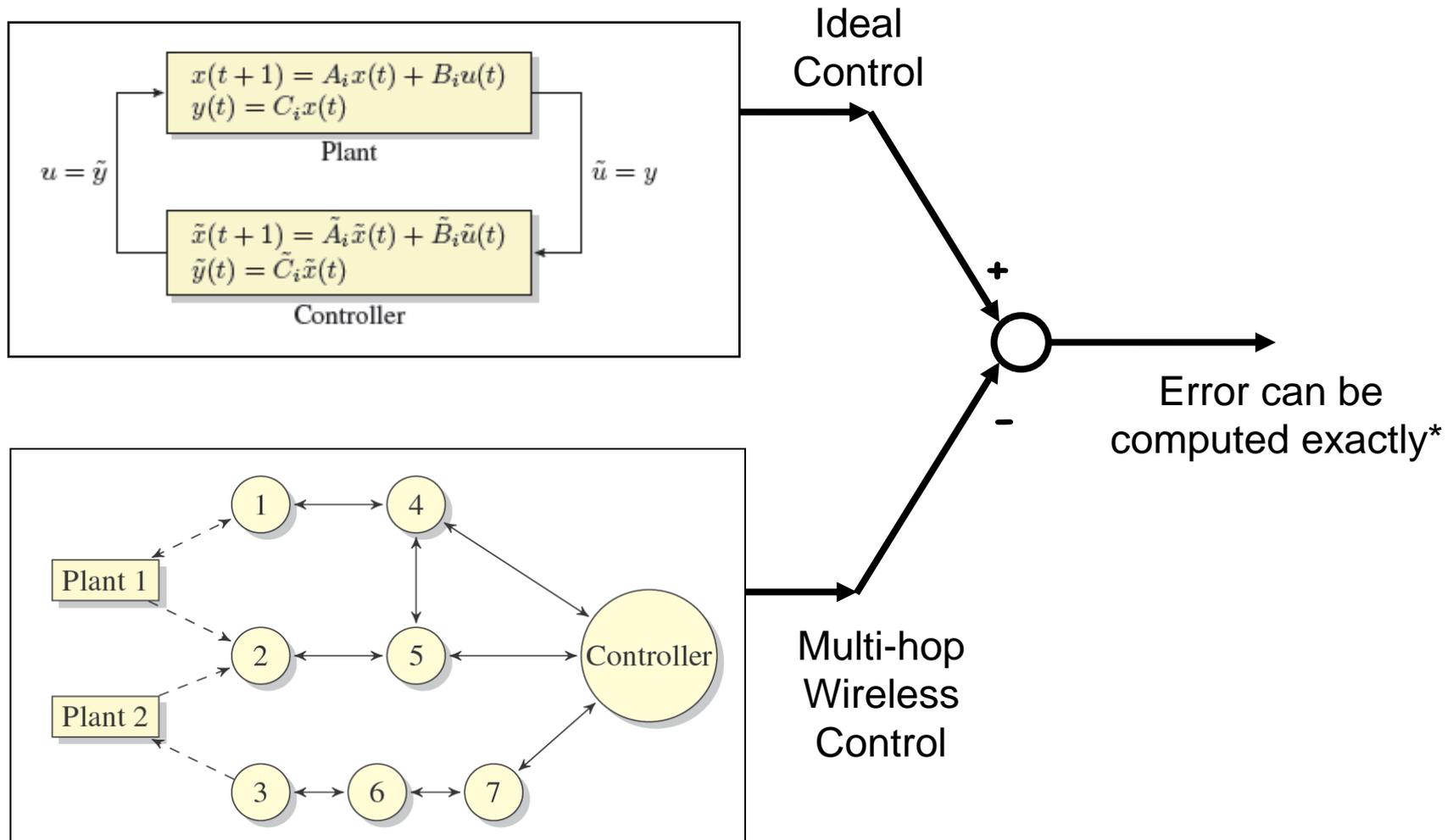


- Formal modeling
- **Analysis & synthesis**
- Compositional analysis
- Industrial case study

- Periodic deterministic schedule (static routing, no TX errors):
 - Theory of periodic time varying linear systems applies
 - Schedule is a fixed string in the alphabet of edges/controllers
 - Nghiem, Pappas, Girard, Alur – EMSOFT 2006, ACM TECS 2012
- Periodic non-deterministic schedule (dynamic routing):
 - Theory of switched/hybrid linear system can be applied
 - Schedule is an automaton over edges/controllers
 - Alur, Weiss – HSCC 2007
- Stochastic analysis (stochastic packet loss, failures):
 - Theory of discrete time Markov jump linear systems applies
 - Schedule is a Markov Chain over edges/controllers
 - Alur, D’Innocenzo, K.H. Johansson, Pappas, Weiss, IEEE CDC 2009, IEEE TAC 2011



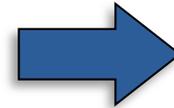
Periodic deterministic schedules



*T. Nghiem, G. Pappas, A. Girard, R. Alur, *Time triggered implementations of dynamic controllers*, ACM Transactions on Embedded Computing Systems, 2012, In press

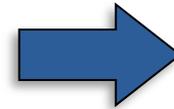
We consider 3 types of failure models:

Long communication disruptions (w.r.t the speed of the control system)



Permanent link failures

Typical packet transmission errors (errors with short time span)

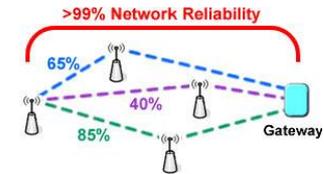


Independent Bernoulli Failures

A general failure model where errors have random time span



A Markov model



Permanent failures are modeled by a function $F : E \rightarrow [0,1]$
 $F(v_1, v_2)$ models the probability that the link (v_1, v_2) fails.

Decision problem: Given a permanent failure model, determine if

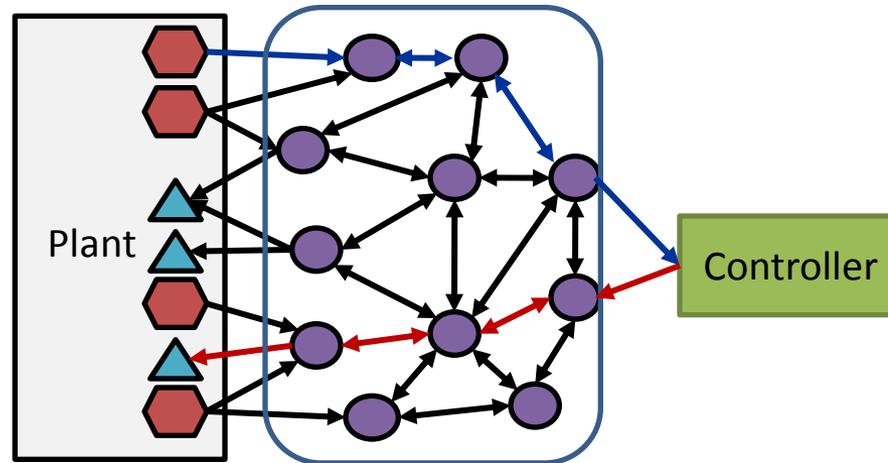
$$P_{stable} \geq a$$

where P_{stable} - probability that the multi-hop control is stable.

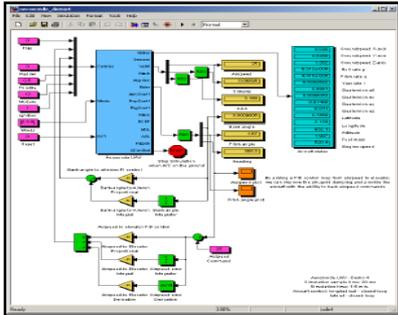
Permanent failure decision problem is **NP-hard** (CDC 2009)

Works for small networks/control loops

- Control with multi-hop wireless networks

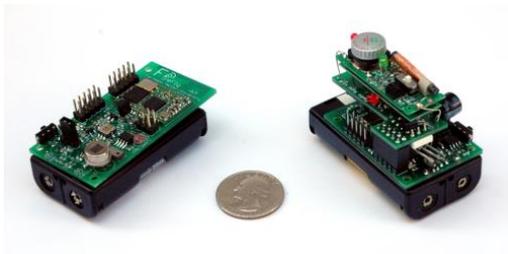


- Formal modeling
- Analysis & synthesis
- **Compositional analysis**
- Industrial case study

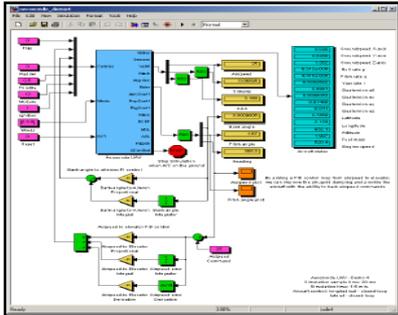


Control Design
Sampling frequency
Delays, jitter

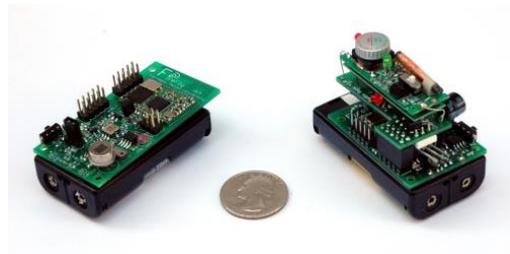
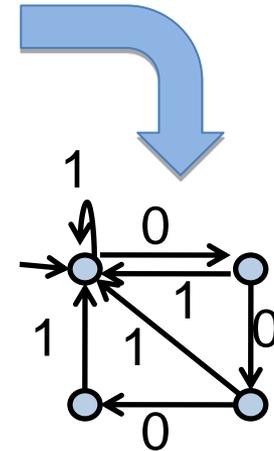
Problems
Impact of scheduling on control
Composing schedules



Scheduling
WCET
RM, EDF



Control Design
Control loop must get
at least one slot in a
superframe of 4 slots



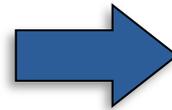
Scheduling
Non-deterministic schedules
for time-triggered platforms



*R. Alur and G. Weiss, *Automata-based interfaces for control and scheduling*, HSCC 2007

- Stability Control Specifications

$$x(t+1) = T[\eta(t), \mu(t)]x(t)$$



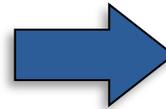
Automata specifying schedules that guarantee stability

- Periodic Control Specifications on TTA

Sample every 100 seconds

If not sampled in the last 200 seconds, sample every 10 seconds for the next minute

Specifications of maximal time delays between events



Automata that specify valid periodic schedules

- Consider control plant with resource constraints on actuator

$$\dot{x} = Ax + \left(\sum_{i=1}^N \beta_i(t) B_i \right) u(t) + w.$$

- Time-dependent switching signal allows only one actuator active at any time

$$\beta(t) \in \{0, 1\}^N \text{ such that } \sum_{i=1}^N \beta_i(t) \leq 1$$

- Many related approaches by Hristu/Brockett '95, Lincoln and Bernhardsson 2000, Zhang, Hu, Abate 2010 etc.
- Generally discrete-time, computationally intensive search for switching signal.

*J. Le Ny, E. Feron, and G. J. Pappas, *Resource constrained LQR control under fast sampling*, HSCC 2011

- Minimize steady state LQR cost over control input and switching signal

$$J_T(\beta, u) = \frac{1}{T} E \left\{ \int_0^T x^T Q x + \left(\sum_i \beta_i(t) u^T R_i u \right) dt + x^T(T) Q_f x(T) \right\}$$

$$J(\beta, u) = \lim_{T \rightarrow \infty} \sup J_T(\beta, u).$$

- Subject to constraints

$$\dot{x} = Ax + \left(\sum_{i=1}^N \beta_i(t) B_i \right) u(t) + w.$$

$$\beta(t) \in \{0, 1\}^N \text{ such that } \sum_{i=1}^N \beta_i(t) \leq 1$$

- Given switching signal and T , LQR controller is optimal. Hence

$$J(\beta) = \limsup_{T \rightarrow \infty} J_T(\beta) = \limsup_{T \rightarrow \infty} \frac{1}{T} \int_0^T \text{Tr}[P_\beta(t; T)W] dt$$

- Optimize above cost over steady-state average utilizations per input

$$b_i \iff \limsup_{T \rightarrow \infty} \frac{1}{T} \int_0^T \beta_i(t) dt$$

- We are keeping average utilization but we are ignoring order

- Compute performance bound using semi-definite programming

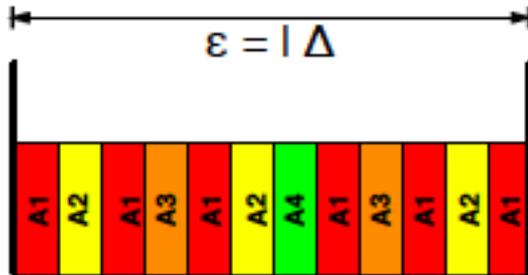
$$\begin{aligned} & \min_{X \succ 0, \{b_i\}_{1 \leq i \leq N}} \text{Tr}[X^{-1}W] \\ & \text{s.t. } XA^T + AX - \left(\sum_{i=1}^N b_i B_i R_i^{-1} B_i^T \right) + XQX \preceq 0 \\ & \sum_{i=1}^N b_i \leq 1, \quad 0 \leq b_i, i = 1, \dots, N. \end{aligned}$$

- Optimize above cost over steady-state average utilizations per input
- Theorem (HSCC 2011): In the limit of arbitrarily fast switching , these policies are asymptotically optimal.

- For simple system with three inputs, SDP provides optimal utilization rates

$$b_1 \approx 0.54, b_2 \approx 0.44, b_3 \approx 0.02$$

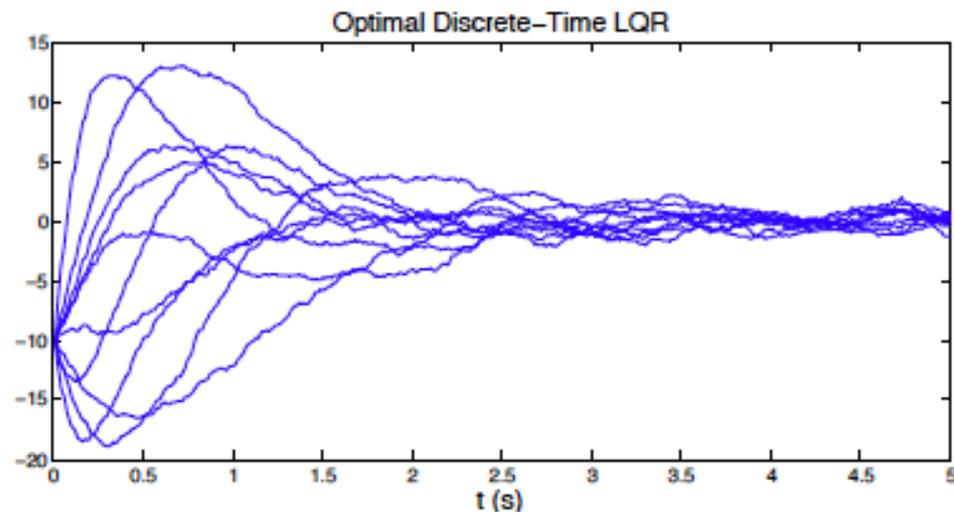
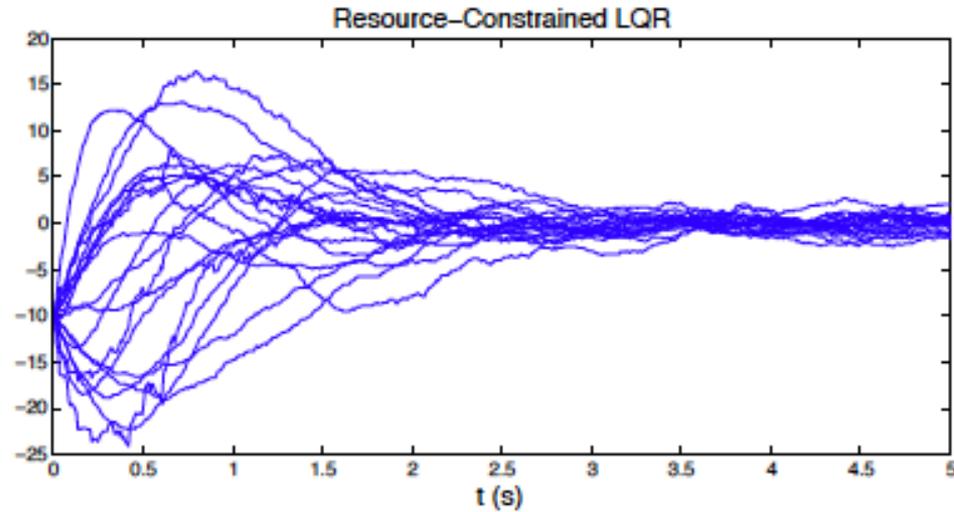
- Approximate optimal utilization rates



$$b_i \approx \frac{l_i}{l}, \quad i = 1, \dots, N, \quad \sum_{i=1}^N l_i \leq l$$

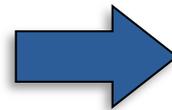
- In a schedule of 100 slots, 54 slots go to input 1, 44 to input 2, etc
- Tradeoff between length of schedule and approximation of utilization

Sample system realizations (10ms slots)



- Stability Control Specifications

$$x(t+1) = T[\eta(t), \mu(t)]x(t)$$



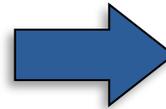
Automata specifying schedules that guarantee performance

- Periodic Control Specifications on TT

Sample every 100 seconds

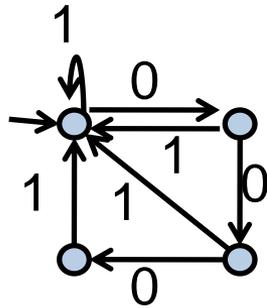
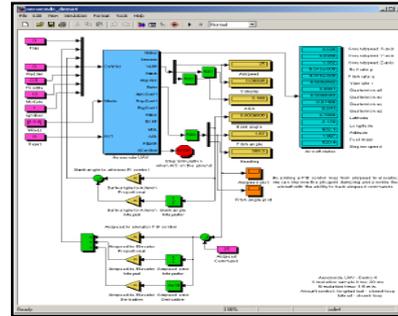
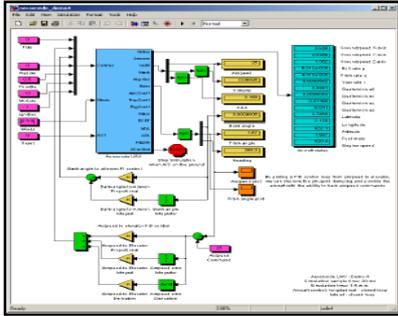
If not sampled in the last 200 seconds, sample every 10 seconds for the next minute

Specifications of maximal time delays between events

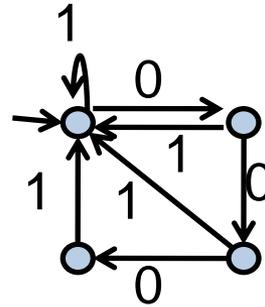


Automata that specify valid periodic schedules

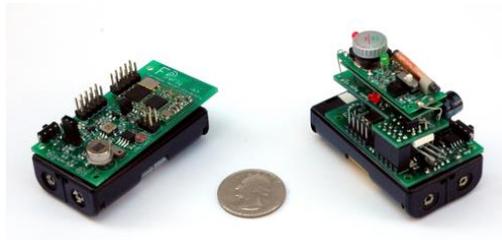
Automata are compositional



\supseteq

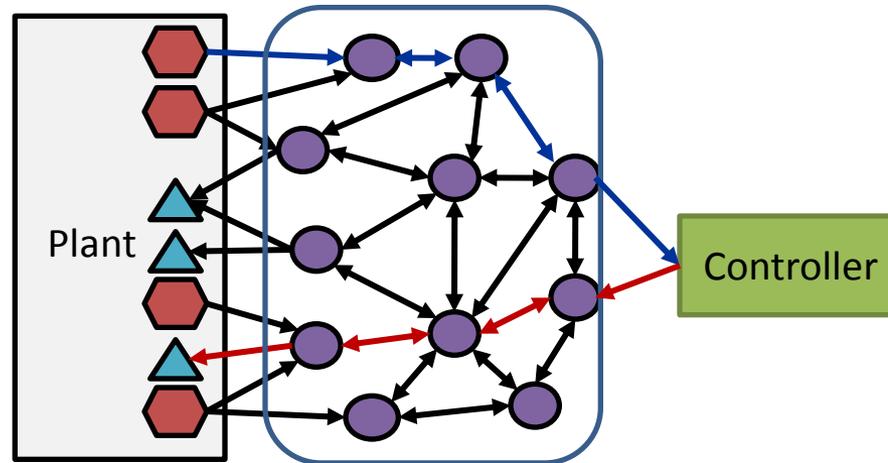


$= ?$



- The more robust the controller, the larger the automaton that can be tolerated with acceptable performance loss.
- The larger the automaton that can be tolerated, the more composable our limited resources will be.
- Tradeoff between control performance and composability

- Control with multi-hop wireless networks



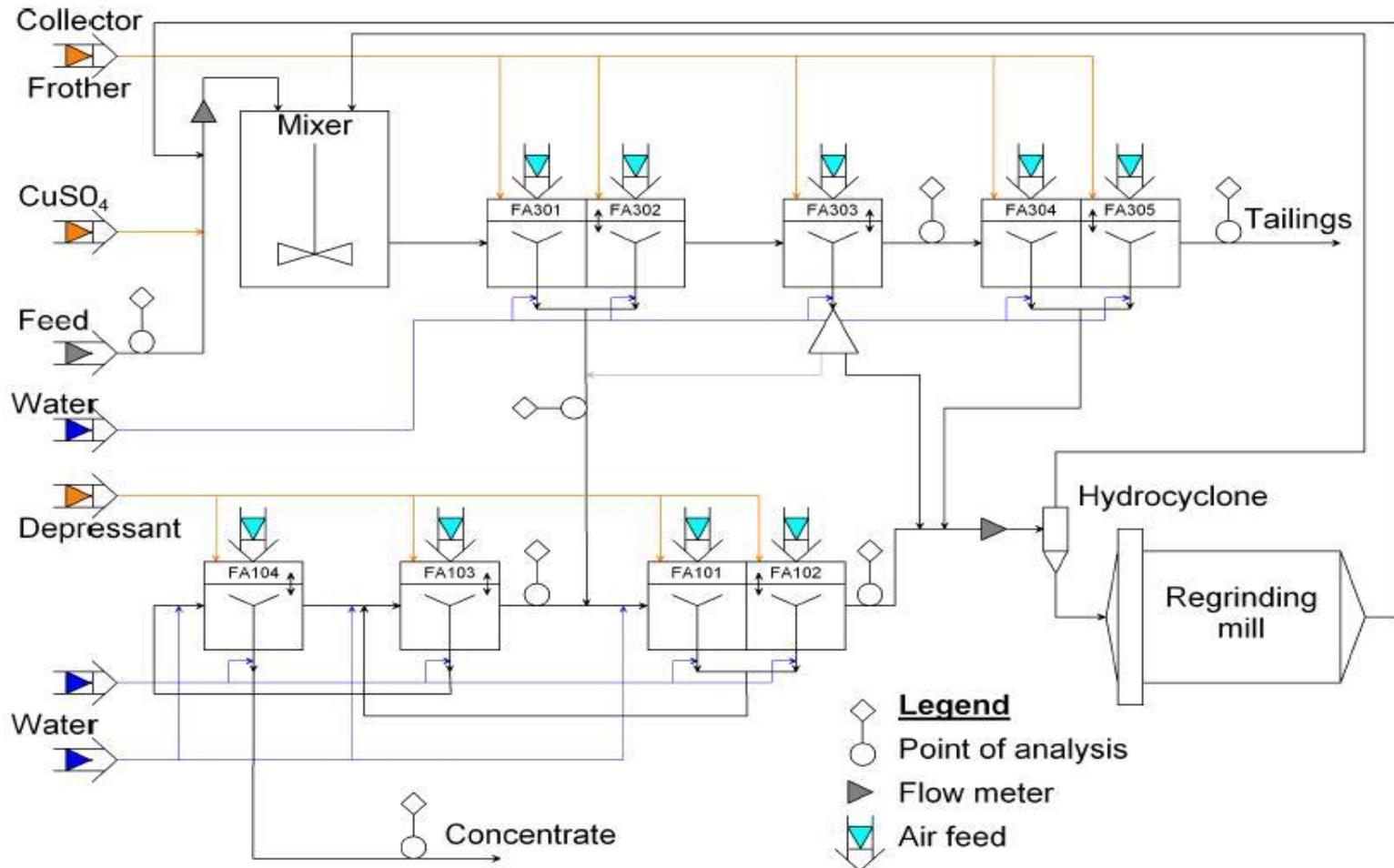
- Formal modeling
- Analysis & synthesis
- Compositional analysis
- **Industrial case study**

Boliden mine in Garpenberg, Sweden

- Mining phases:
 - Drilling and blasting
 - Ore transportation
 - Ore concentration



Flotation bank control problem



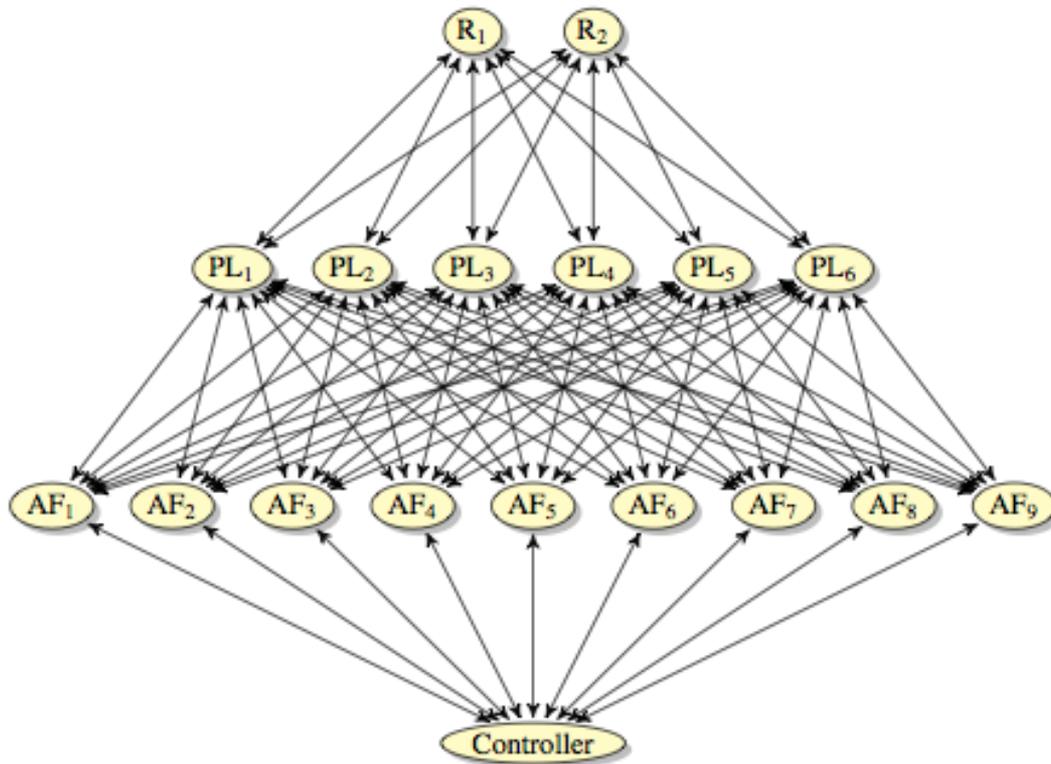
H. Lindvall, "Flotation modelling at the Garpenberg concentrator using Modelica/Dymola," 2007.

Process Time Scales: Zn Flotation

Loop category	# of loops in category	Loop name	Sampling interval (T_s)
Air flow	9	FA301_FC1	2
		FA302_FC1	2
		FA303_FC1	2
		FA304_FC1	2
		FA305_FC1	2
		FA101_FC1	2
		FA102_FC1	2
		FA103_FC1	2
		FA104_FC1	2
		Level	6
FA303_LC1	1		
FA305_LC1	8		
FA102_LC1	8		
FA103_LC1	8		
FA104_LC1	8		

Loop category	# of loops in category	Loop name	Sampling interval (T_s)
Reagents	2	BL031_FC1	2
		FA300_FC2	1

- Each controlled variable represents a control loop
- Only the main control loops:
 - air flow, pulp level and reagent
- Each loop abstracted by a time constraint (the sampling interval)
 - specifies the maximum delay between sensing and actuation
- The sampling interval used as a constraint for defining the set of “good” schedules



Using SMV to compose schedules

```

MODULE loop2(bus)
VAR
  cnt:0..6;
ASSIGN
  init(cnt):=0;
  next(cnt):=case
bus=e2to5 & cnt=0 : 1;
bus=e5toc & cnt=1 : 2;
bus=bus & cnt=2 : 3;
bus=ecto7 & cnt=3 : 4;
bus=e7to6 & cnt=4 : 5;
bus=e6to3 & cnt=5 : 6;
1:cnt;
  esac;
DEFINE
  done := cnt=6;
    
```

progress
counters

```

MODULE loop1(bus)
VAR
  in1:0..2;
  in2:0..2;
  out1:0..3;
ASSIGN
  init(in1):=0;
  init(in2):=0;
  init(out1):=0;
  next(in1):=case
bus=e1to4 & in1=0 : 1;
bus=e4toc & in1=1 : 2;
1:in1;
  esac;
  next(in2):=case
bus=e2to5 & in2=0 : 1;
bus=e5toc & in2=1 : 2;
1:in2;
  esac;
  next(out1):=case
bus=bus & allin & out1= 0 :1;
bus=ecto4 & allin & out1= 1 : 2;
bus=e4to1 & allin & out1= 2 : 3;
1 : out1;
  esac;
DEFINE
  allin := in1=2 & in2=2;
  done := out1=3;
    
```

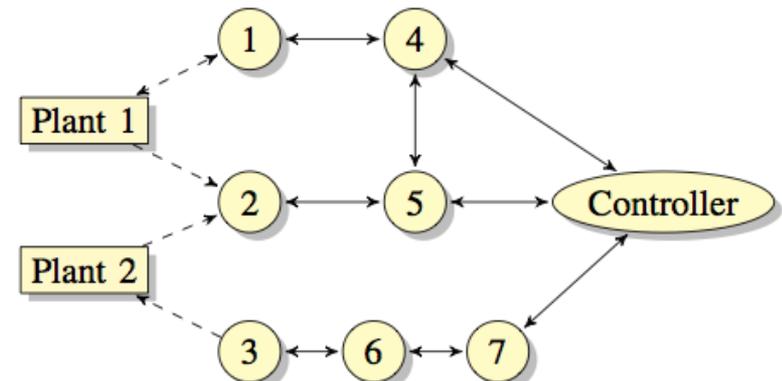
```

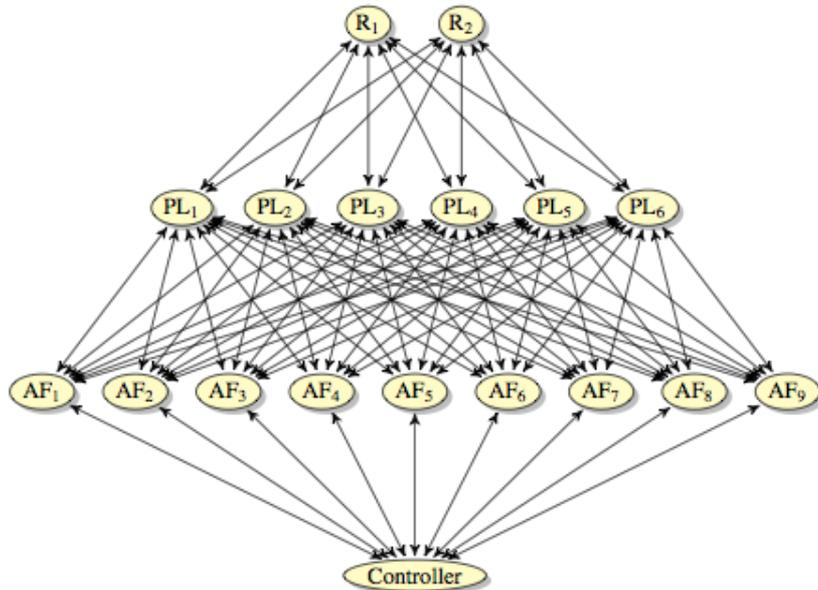
MODULE main
VAR
  bus:{e1to4, e2to5, e4to1, e4toc, e5toc, e6to3, e7to6, ecto4, ecto7, idle};
  I1:loop1(bus);
  I2:loop2(bus);
SPEC
  AG !(I1.done & I2.done);
    
```

Req. For Plant 2:
e2to5, e5toC, ...,e6to3
must be a subsequence
of the schedule

Req. For Plant 1:
more involved because it
has two inputs

We are looking for a schedule that satisfies both requirements which comes as a counter-example to the claim that there is no such schedule





17 single-input-single-output loops
Timing constraints
At most one message in a time slot



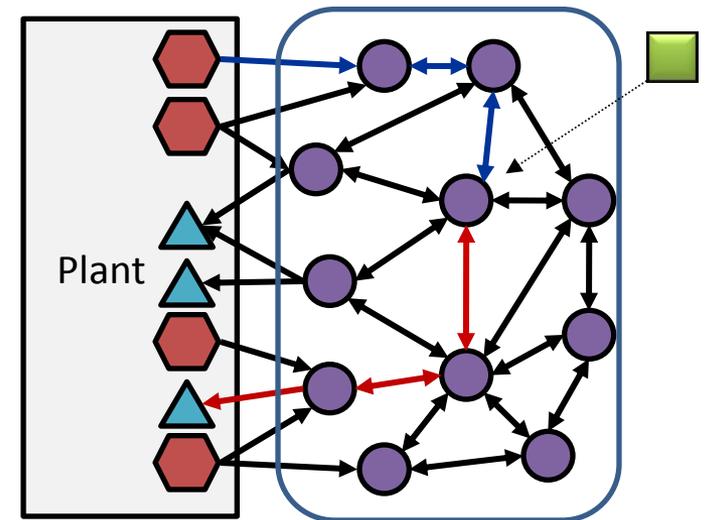
SMV code with 18 modules
272 lines
BDD nodes allocated: 26797



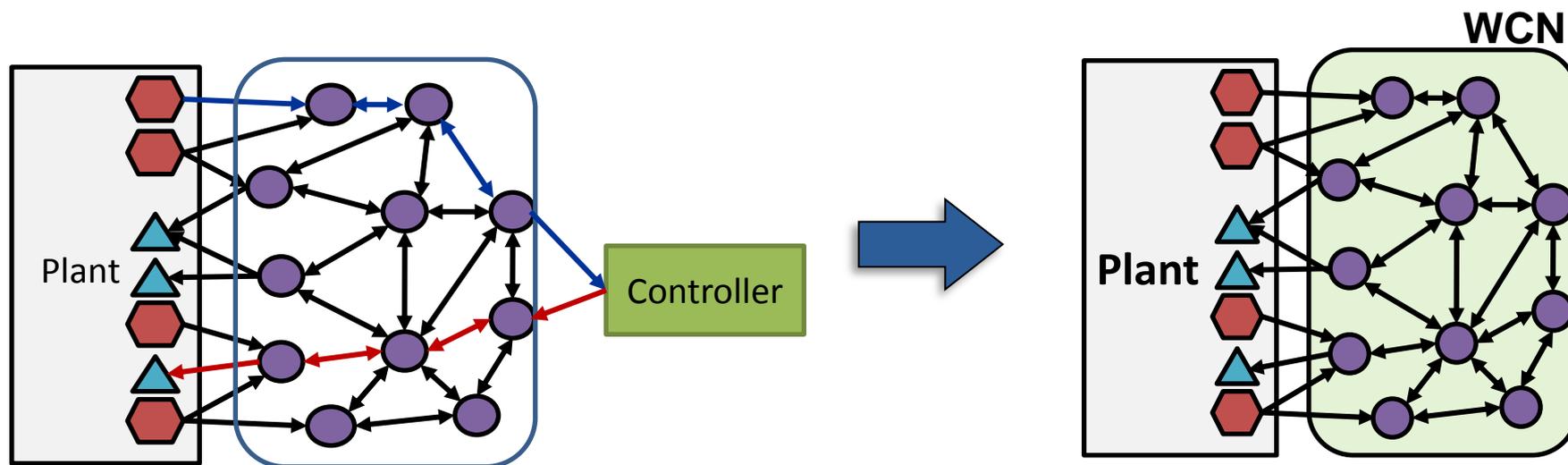
~2 minutes

Shortest schedule that satisfy the
constraints posed by all 17 loops
37 time slots

- Time-triggered architectures not optimal for event-based systems
 - Hybrid TDMA/CSMA or LTTA architectures
 - Event-based sensing and control
- Time-synchronization for large networks
 - Model TDMA clock drift using timed automata
 - Scheduling by composing timed-automata
- Wireless models are not precise
 - On-line adaptation of packet drop probability
 - Robust/adaptive control
- Control over virtual network computation
 - Runtime control reconfiguration in presence of node failures
 - Embedded virtual machines for control [Pajic, Mangharam 2012]

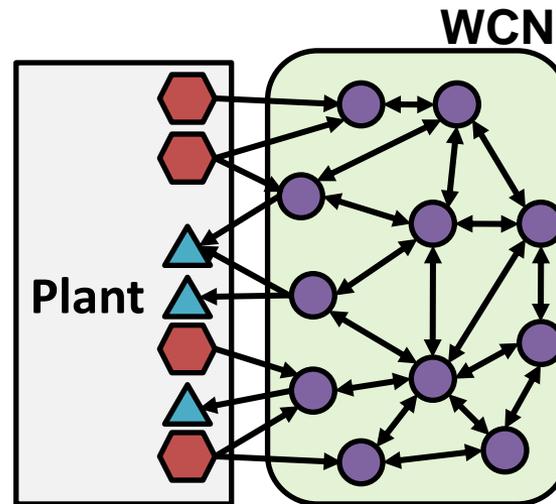


- In multi-hop control, nodes route information to controller



- Can we leverage computation of the network?
- Can we distribute the controller to nodes of the network?
- Reminiscent of network coding

- Wireless control network*



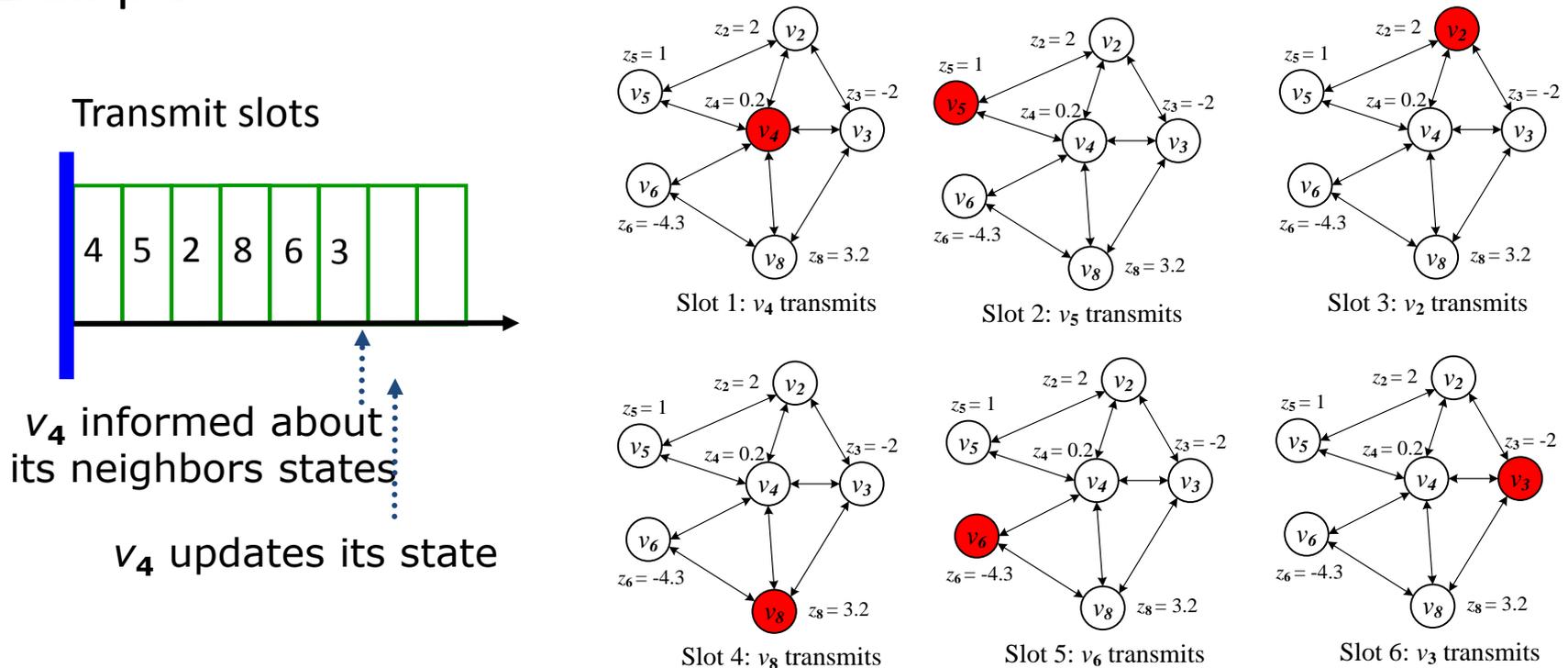
- **Modeling**
- Controller synthesis
- Robustness & security

*M. Pajic, S. Sundaram, G. Pappas, R. Mangharam, *Wireless Control Network: A New Approach for Control over Network*, IEEE Transactions on Automatic Control, 2011.

M. Pajic, R. Mangharam, G.J. Pappas, S. Sundaram, Topological Conditions for In-Network Stabilization of Dynamical Systems, IEEE Journal on Selected Areas of Communication, 2013

M. Pajic, S. Sundaram, J. Le Ny, G.J. Pappas, R. Mangharam, Closing the Loop: A Simple Decentralized Method for Control over Wireless Networks, IPSN'12

- Each node maintains its (possible vector) state
 - Transmits state exactly once in each step (per frame)
 - Updates own state using linear iterative strategy
- Example:



- Discrete-time plant $\mathbf{x}[k + 1] = \mathbf{A}\mathbf{x}[k] + \mathbf{B}\mathbf{u}[k] + \mathbf{B}_w \mathbf{u}_w[k]$
 $\mathbf{y}[k] = \mathbf{C}\mathbf{x}[k],$

- Node state update procedure:

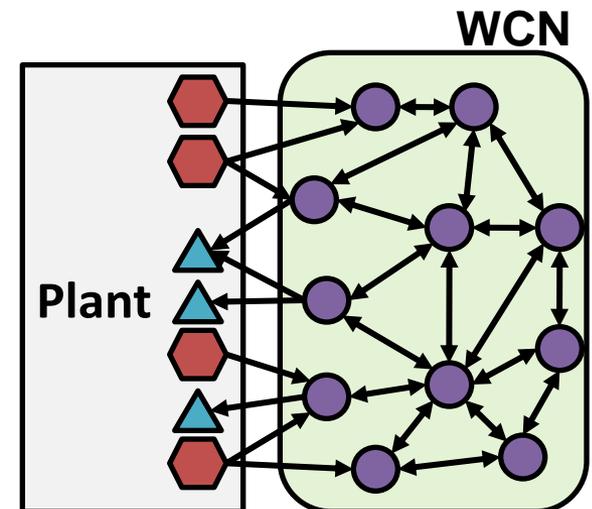
$$z_i[k + 1] = w_{ii}z_i[k] + \sum_{v_j \in \mathcal{N}_{v_i}} w_{ij}z_j[k] + \sum_{s_j \in \mathcal{N}_{v_i}} h_{ij}y_j[k]$$

From neighbors
From sensors

- Actuator update procedure:

$$u_i[k] = \sum_{j \in \mathcal{N}_{a_i}} g_{ij}z_j[k]$$

From actuator's neighbors

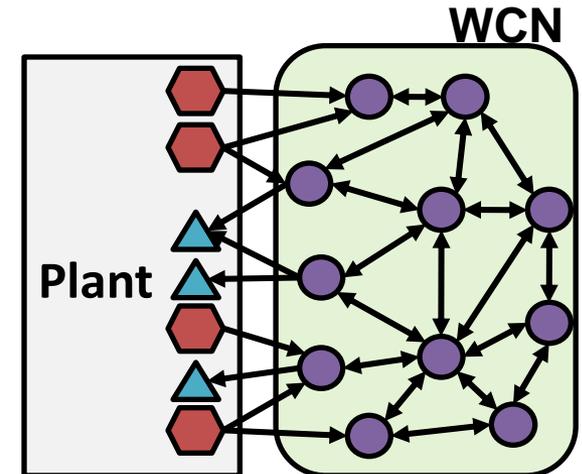


- Network acts as a linear dynamical compensator

$$\mathbf{z}[k+1] = \underbrace{\begin{bmatrix} w_{11} & w_{12} & \cdots & w_{1N} \\ w_{21} & w_{22} & \cdots & w_{2N} \\ \vdots & \vdots & \ddots & \vdots \\ w_{N1} & w_{N2} & \cdots & w_{NN} \end{bmatrix}}_{\mathbf{W}} \mathbf{z}[k] + \underbrace{\begin{bmatrix} h_{11} & h_{12} & \cdots & h_{1p} \\ h_{21} & h_{22} & \cdots & h_{2p} \\ \vdots & \vdots & \ddots & \vdots \\ h_{N1} & h_{N2} & \cdots & h_{Np} \end{bmatrix}}_{\mathbf{H}} \mathbf{y}[k]$$

$$\mathbf{u}[k] = \underbrace{\begin{bmatrix} g_{11} & g_{12} & \cdots & g_{1N} \\ g_{21} & g_{22} & \cdots & g_{2N} \\ \vdots & \vdots & \ddots & \vdots \\ g_{m1} & g_{m2} & \cdots & g_{mN} \end{bmatrix}}_{\mathbf{G}} \mathbf{z}[k]$$

Structural constraints: Only elements corresponding to existing links (link weights) are allowed to be non-zero

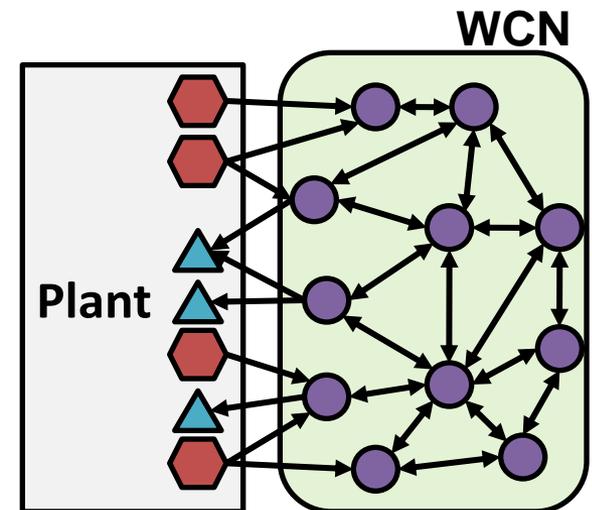


- Overall system state: $\hat{\mathbf{x}}[k] = \begin{bmatrix} \mathbf{x}[k] \\ \mathbf{z}[k] \end{bmatrix}$
 - ← plant
 - ← network

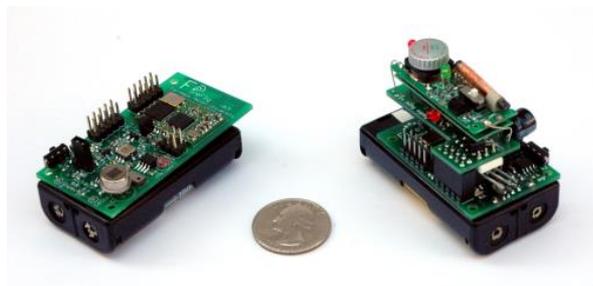
- Closed-loop system:

$$\hat{\mathbf{x}}[k + 1] = \underbrace{\begin{bmatrix} \mathbf{A} & \mathbf{BG} \\ \mathbf{HC} & \mathbf{W} \end{bmatrix}}_{\hat{\mathbf{A}}} \underbrace{\begin{bmatrix} \mathbf{x}[k] \\ \mathbf{z}[k] \end{bmatrix}}_{\hat{\mathbf{x}}[k]} + \underbrace{\begin{bmatrix} \mathbf{B}_w \\ \mathbf{0} \end{bmatrix}}_{\hat{\mathbf{B}}} \mathbf{u}_w$$

- Matrices \mathbf{W} , \mathbf{G} , \mathbf{H} are structured
- Sparsity constraints imposed by topology!



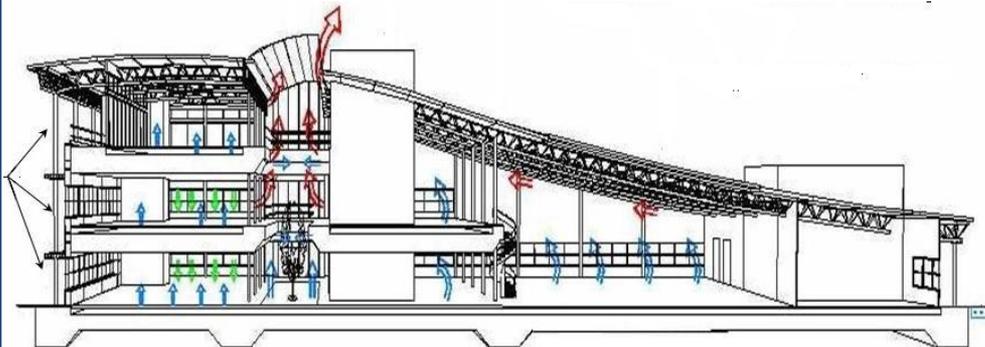
- Low overhead
 - Each node only calculates linear combination of its states and state of its neighbors
 - Suitable even for resource constrained nodes
 - Easily incorporated into existing wireless networks (e.g., systems based on the ISA100.11a or wirelessHART)
 - Backup mechanism in ‘traditional’ networked control systems; used for graceful degradation



- Simple scheduling
 - Each node needs to transmit only once per frame
 - Static (conflict-free) schedule
- No routing!
- Multiple sensing/actuation points
 - Geographically distributed sensors/actuators

Process control

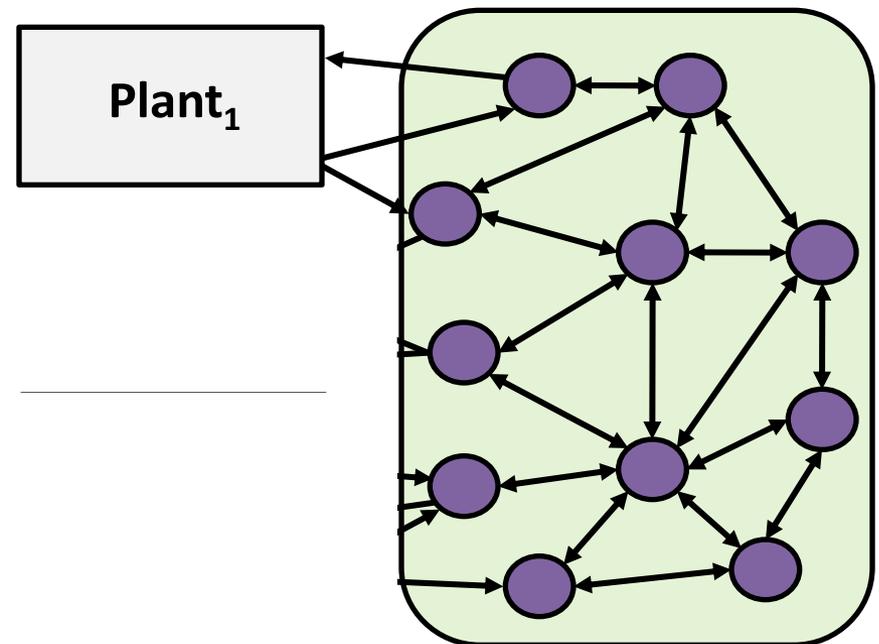
Building automation



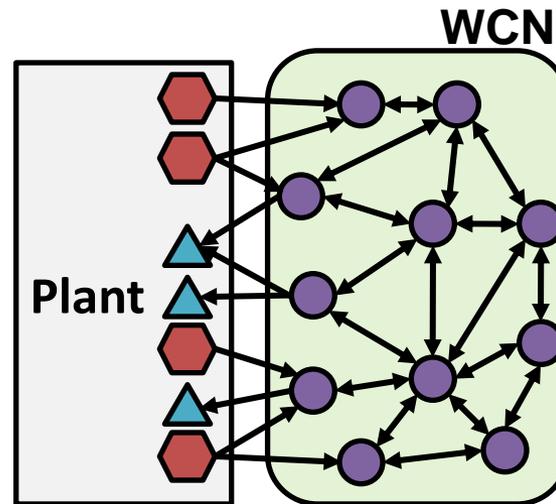
- Adding new control loops is easy!
 - Does not require any communication schedule recalculation
- WCN configurations can be combined

Stable configuration

$$(\mathbf{W}_1, \mathbf{H}_1, \mathbf{G}_1) \in \Psi$$

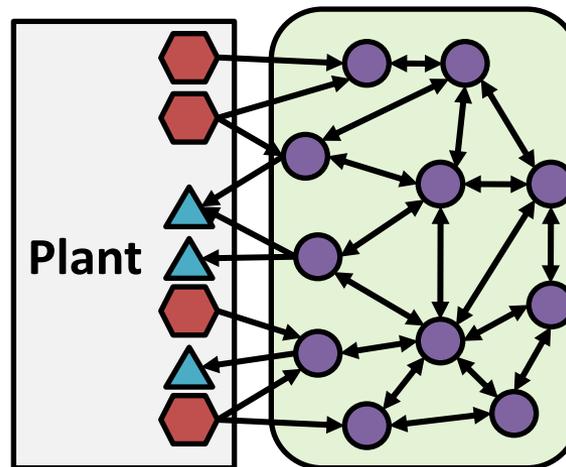


- Wireless control network



- Modeling
- **Controller synthesis**
- Robustness & security

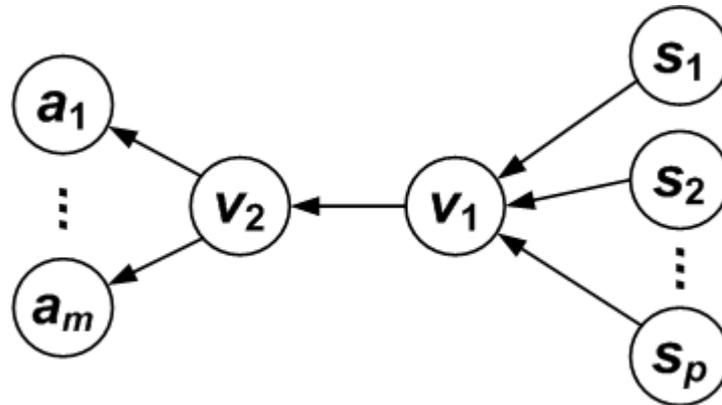
- Use WCN to stabilize the closed-loop system
 - Synthesis of optimal WCN configurations



- Does the plant influence the WCN network topology?
 - How many nodes? How to interconnect them?
- Given network topology, design distributed controller
 - Extracting a stabilizing closed loop configuration

- The objective of the network is systems stabilization!

- Example:

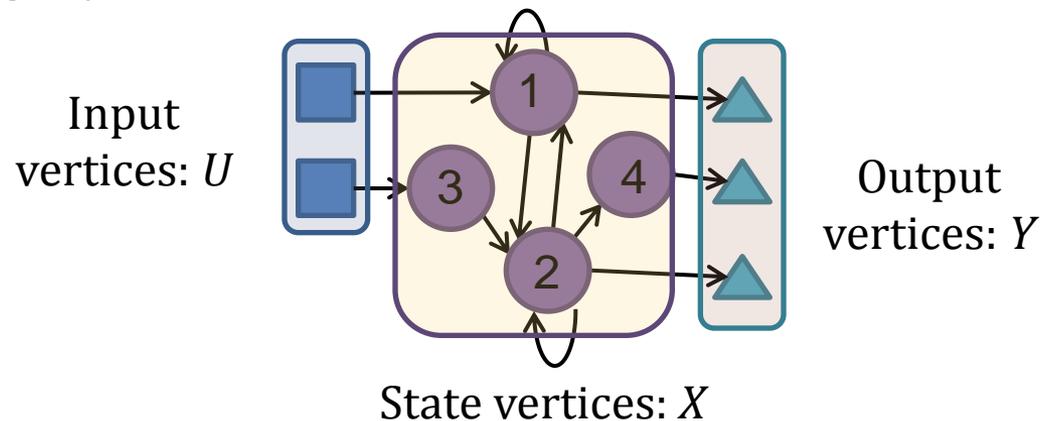


- This network is not capable of delivering all of the source information to all of the sinks at each time-step
- That is not necessarily a cause for concern when the main objective is to stabilize the system.

- Structured system theory: Systems represented as graphs
- Linear system

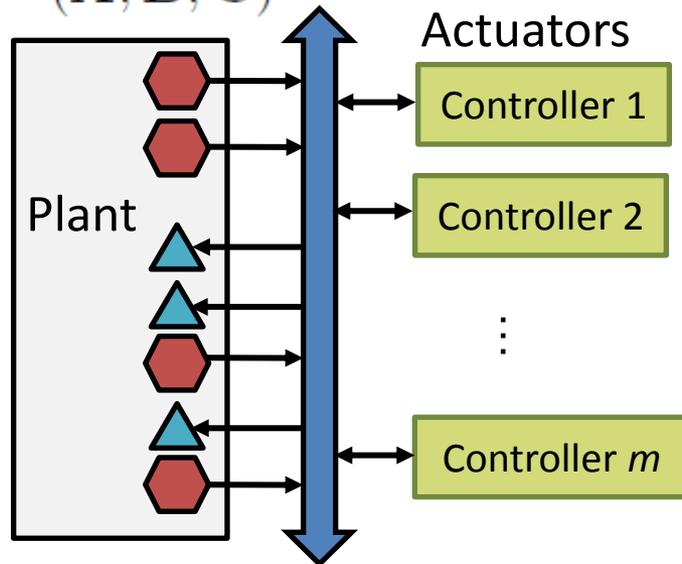
$$x[k+1] = \begin{bmatrix} \lambda_1 & \lambda_2 & 0 & 0 \\ \lambda_3 & \lambda_4 & \lambda_5 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & \lambda_6 & 0 & 0 \end{bmatrix} x[k] + \begin{bmatrix} \lambda_7 & 0 \\ 0 & 0 \\ 0 & \lambda_8 \\ 0 & 0 \end{bmatrix} u[k], \quad y[k] = \begin{bmatrix} \lambda_9 & 0 & 0 & 0 \\ 0 & \lambda_{10} & 0 & 0 \\ 0 & 0 & 0 & \lambda_{11} \end{bmatrix} x[k]$$

- Associated graph H



- Properties of graph are generic properties of structured system

$$\Sigma = (A, B, C)$$



Decentralized control system

From feedback patterns J_1, J_2, \dots, J_m

$$\mathbf{K}_f = \{ \mathbf{K} \in \mathbb{R}^{m \times p} \mid k_{ij} = 0 \text{ if } j \notin J_i \}$$

Fixed Modes [Wang & Davison, 1973; Siljak, 1981]

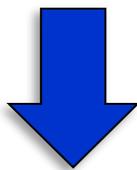
$$\Lambda_f = \bigcap_{\mathbf{K} \in \mathbf{K}_f} \Lambda(A + \mathbf{BK}\mathbf{C})$$

Indicate whether the system can be stabilized

The plant \leftrightarrow network model

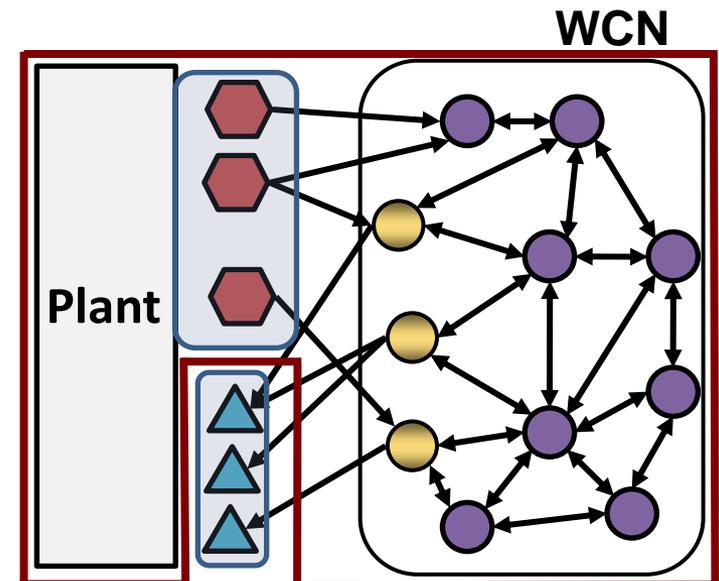
$$\hat{\mathbf{x}}[k+1] = \begin{bmatrix} \mathbf{x}[k+1] \\ \mathbf{z}[k+1] \end{bmatrix} = \underbrace{\begin{bmatrix} \mathbf{A} & \mathbf{0} \\ \mathbf{HC} & \mathbf{W} \end{bmatrix}}_{\tilde{\mathbf{A}}} \begin{bmatrix} \mathbf{x}[k] \\ \mathbf{z}[k] \end{bmatrix} + \underbrace{\begin{bmatrix} \mathbf{B} \\ \mathbf{0} \end{bmatrix}}_{\tilde{\mathbf{B}}} \mathbf{u}[k]$$

$$\hat{\mathbf{y}}[k] = \underbrace{\begin{bmatrix} \mathbf{0} & \mathbf{E}_{\mathcal{V}_A} \end{bmatrix}}_{\tilde{\mathbf{C}}} \begin{bmatrix} \mathbf{x}[k] \\ \mathbf{z}[k] \end{bmatrix},$$



New plant: Plant & WCN

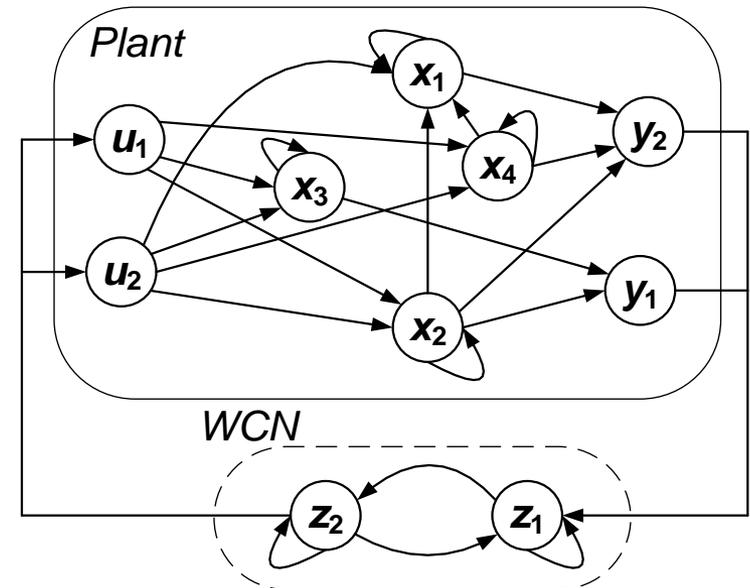
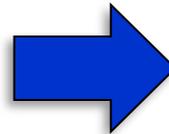
Controlled by controllers at the actuators



- Use structured system theory and decentralized control on the WCN and network

$$x[k+1] = \begin{bmatrix} 2 & 0 & 1 & -3 \\ 0 & 2 & 10 & -4 \\ 0 & 0 & 2 & 0 \\ 0 & 0 & 0 & 3 \end{bmatrix} x[k] + \begin{bmatrix} 0 & 1 \\ 1 & 1.6 \\ -0.5 & 4 \\ 2 & 5 \end{bmatrix} u[k]$$

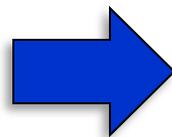
$$y[k] = \begin{bmatrix} 1 & 0.3 & 2 & 0 \\ 0 & 0.1 & 0 & 1 \end{bmatrix} x[k]$$



- Can we stabilize the plant with 2 nodes?

- Consider a numerically specified system
- Example: A system with integrators

$$A = \begin{bmatrix} 2 & 0 & 1 & -3 \\ 0 & 2 & 10 & -4 \\ 0 & 0 & 2 & 0 \\ 0 & 0 & 0 & 3 \end{bmatrix}$$



Eigenvalues are 2,2,2,3

$\Lambda=2$ has geometric multiplicity $d=2 (\geq 1)$

Network condition: Let d denote the largest geometric multiplicity of any unstable eigenvalue of the plant. If

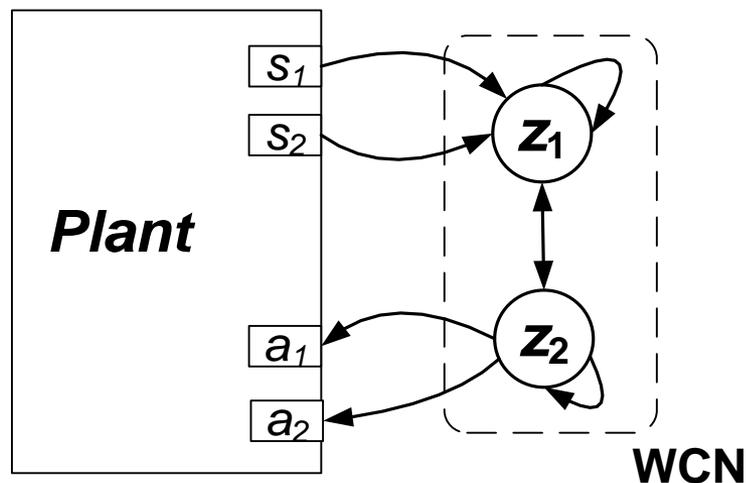
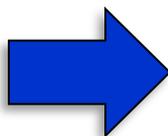
- 1) connectivity of the network is at least d , and
- 2) each actuator has at least d nodes in neighborhood

then there exists a stabilizing configuration for WCN

- Use structured system theory on WCN and network

$$x[k+1] = \begin{bmatrix} 2 & 0 & 1 & -3 \\ 0 & 2 & 10 & -4 \\ 0 & 0 & 2 & 0 \\ 0 & 0 & 0 & 3 \end{bmatrix} x[k] + \begin{bmatrix} 0 & 1 \\ 1 & 1.6 \\ -0.5 & 4 \\ 2 & 5 \end{bmatrix} u[k]$$

$$y[k] = \begin{bmatrix} 1 & 0.3 & 2 & 0 \\ 0 & 0.1 & 0 & 1 \end{bmatrix} x[k]$$

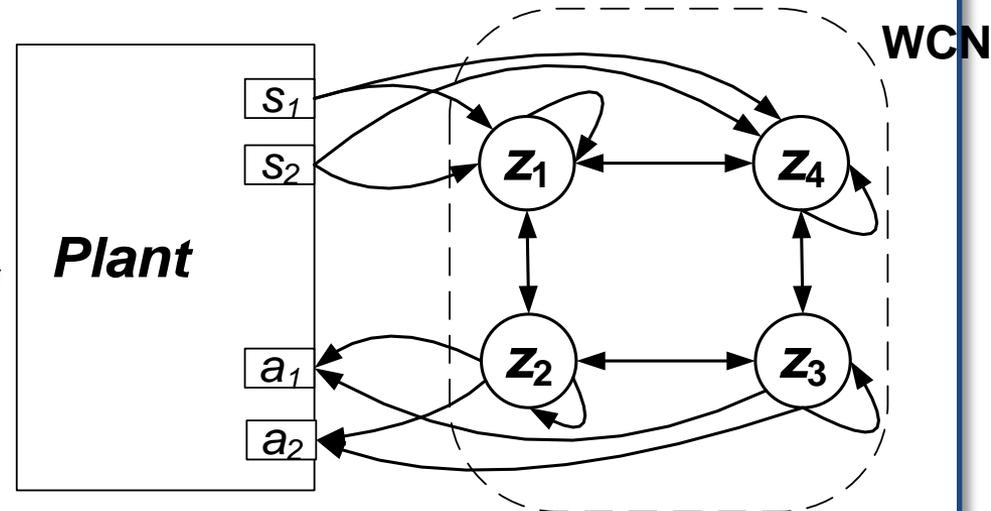
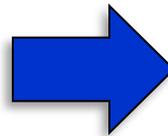


- We **cannot** stabilize with with 2 nodes!

- Use structured system theory on WCN and network

$$x[k+1] = \begin{bmatrix} 2 & 0 & 1 & -3 \\ 0 & 2 & 10 & -4 \\ 0 & 0 & 2 & 0 \\ 0 & 0 & 0 & 3 \end{bmatrix} x[k] + \begin{bmatrix} 0 & 1 \\ 1 & 1.6 \\ -0.5 & 4 \\ 2 & 5 \end{bmatrix} u[k]$$

$$y[k] = \begin{bmatrix} 1 & 0.3 & 2 & 0 \\ 0 & 0.1 & 0 & 1 \end{bmatrix} x[k]$$



- We **cannot** stabilize with with 2 nodes!
- But we can stabilize plant with 4 nodes

- Is fully connected network sufficient?

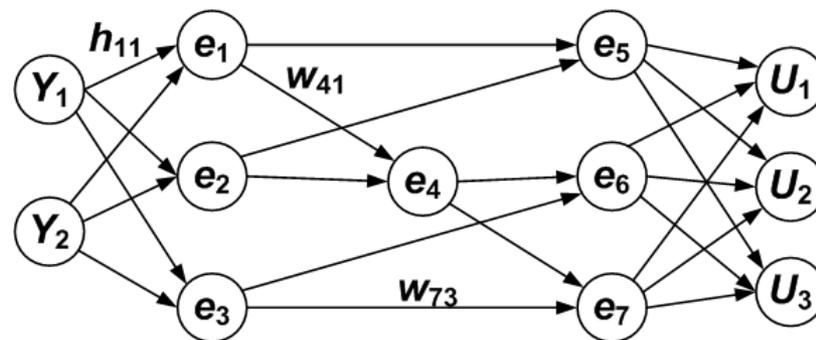
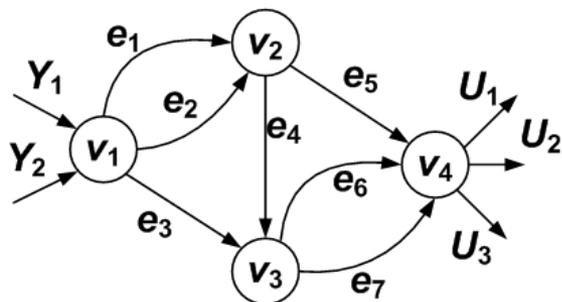
Sufficient condition: If

- 1) Geometric multiplicity is 1 for all unstable eigenvalues,
- 2) System is controllable and and observable,

then it can be stabilized with a strongly connected network, where each sensor and actuator is connected to the network.

Generic condition!

- **Problem:** network synthesis for stabilization when *network coding over point-to-point communication links* is used
- Example: Point to point communication in a simple network



Direct labeled line graph

- Algebraic approach to network coding (Koetter, Medard, 2005)
 - each link in the initial graph is mapped to a unique vertex in the line graph
- The labeled line graph directly corresponds to the WCN model!

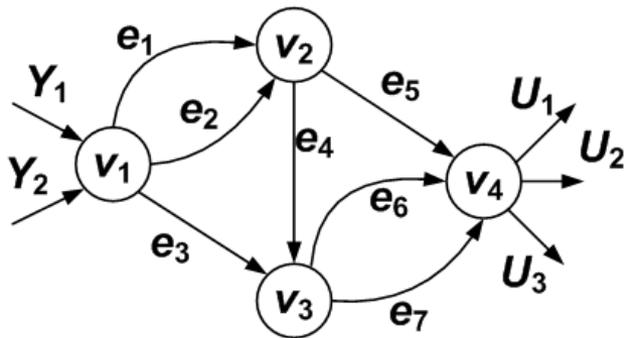
- Consequently, the same reasoning can be used for point-to-point networks

Sufficient condition when point-to-point networks with linear network coding are used for communication:

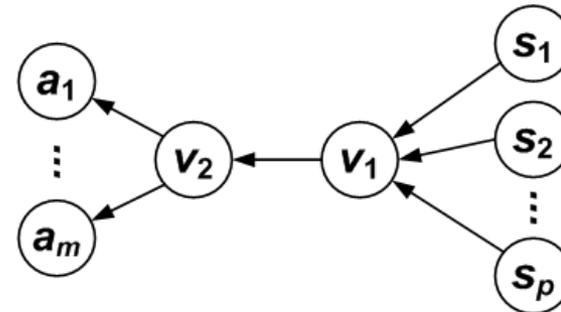
Let d denote the largest geometric multiplicity of any unstable e-value of a detectable and stabilizable plant. If ***edge connectivity*** of the network **between sensors and actuators** is at least d then the system can be stabilized using dynamic compensators at actuators.

The equivalent generic condition also holds!

- Problem: network synthesis for stabilization, in the case where ***network coding over point-to-point communication links*** is used
- Examples: Point-to-point communication in simple networks

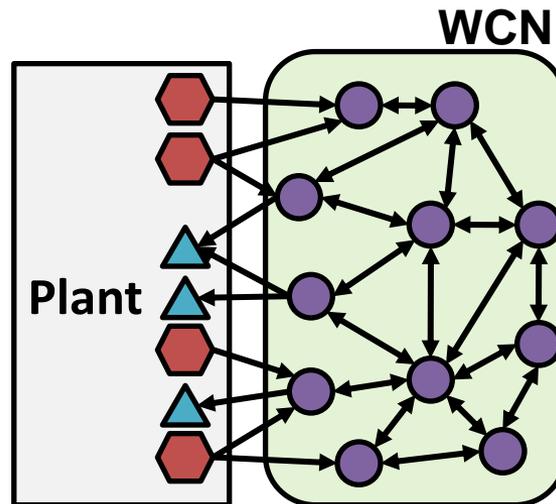


Stabilizable for $d \leq 3$



Stabilizable for $d \leq 1$

- Use WCN to stabilize the closed-loop system



- For a specific WCN network topology
 - How to stabilize the closed-loop system

- **Problem:** Find numerical matrices **W**, **H**, **G** satisfying structural constraints such that

$$\hat{\mathbf{A}} = \begin{bmatrix} \mathbf{A} & \mathbf{BG} \\ \mathbf{HC} & \mathbf{W} \end{bmatrix} \quad \text{is stable}$$

- **Solution:** Formulate Lyapunov function and try to solve using Linear Matrix Inequalities (LMIs)
 - Find positive definite matrix **P** such that

$$\mathbf{P} - \hat{\mathbf{A}}^T \mathbf{P} \hat{\mathbf{A}} > \mathbf{0}$$

- Schur complement:

$$\begin{array}{l} \mathbf{Q} - \mathbf{R}^T \mathbf{S}^{-1} \mathbf{R} > 0 \\ \mathbf{S} > 0 \end{array} \Leftrightarrow \begin{bmatrix} \mathbf{Q} & \mathbf{R}^T \\ \mathbf{R} & \mathbf{S} \end{bmatrix} > 0$$

- Standard application to stability:

$$\begin{array}{l} \mathbf{P} - \hat{\mathbf{A}}^T \mathbf{P} \hat{\mathbf{A}} > 0 \\ \Leftrightarrow \mathbf{P} - \hat{\mathbf{A}}^T \mathbf{P} \mathbf{P}^{-1} \mathbf{P} \hat{\mathbf{A}} > 0 \\ \Leftrightarrow \begin{bmatrix} \mathbf{P} & \hat{\mathbf{A}}^T \mathbf{P} \\ \mathbf{P} \hat{\mathbf{A}} & \mathbf{P} \end{bmatrix} > 0 \end{array}$$

- **Bilinear matrix inequality** (free variables in $\hat{\mathbf{A}}$ multiply free variables in \mathbf{P})
- Not a problem when \mathbf{W} , \mathbf{H} and \mathbf{G} are unstructured -> a change of variables produces an LMI

- Change of variables no longer works when $\hat{\mathbf{A}}$ is structured
- Alternative approach [*de Oliveira et. al, CDC'00*]:

$$\begin{aligned}
 \mathbf{P} - \hat{\mathbf{A}}^T \mathbf{P} \hat{\mathbf{A}} > 0 &\Leftrightarrow \begin{bmatrix} \mathbf{P} & \hat{\mathbf{A}}^T \\ \hat{\mathbf{A}} & \mathbf{P}^{-1} \end{bmatrix} > 0 \quad \text{linear in } \hat{\mathbf{A}} \\
 &\Leftrightarrow \begin{bmatrix} \mathbf{P} & \hat{\mathbf{A}}^T \\ \hat{\mathbf{A}} & \mathbf{Q} \end{bmatrix} > 0, \quad \boxed{\mathbf{QP} = \mathbf{I}} \quad \text{nonconvex constraint}
 \end{aligned}$$

- Problem is still nonconvex,
 - This form appears frequently in design of static output feedback controllers

- Various methods developed to deal with constraint $\mathbf{QP} = \mathbf{I}$
- Use approach by [*El Ghaoui et al., TAC, 1997*]:
 - Positive definite $n \times n$ matrices \mathbf{P} and \mathbf{Q} satisfy $\mathbf{QP} = \mathbf{I}$ if and only if they are optimal solutions to the problem

$$\begin{aligned} & \min \operatorname{tr}(\mathbf{QP}) \\ & s.t. \begin{bmatrix} \mathbf{P} & \mathbf{I} \\ \mathbf{I} & \mathbf{Q} \end{bmatrix} \geq 0 \end{aligned}$$

and the minimum cost is n .

- Still nonlinear \rightarrow linearize around a feasible point $\mathbf{P}_0, \mathbf{Q}_0$

Find feasible points

P_0, Q_0, W_0, H_0, G_0

Solve the LMI problem,
from P_k, Q_k find

$P_{k+1}, Q_{k+1}, W_{k+1}, H_{k+1}, G_{k+1}$

Function Linearization

$$\min tr(\mathbf{P}_k \mathbf{Q}_{k+1} + \mathbf{Q}_k \mathbf{P}_{k+1})$$

$$\begin{bmatrix} \mathbf{P}_{k+1} & \hat{\mathbf{A}}_{k+1}^T \\ \hat{\mathbf{A}}_{k+1} & \mathbf{Q}_{k+1} \end{bmatrix} > 0, \quad \begin{bmatrix} \mathbf{P}_{k+1} & \mathbf{I} \\ \mathbf{I} & \mathbf{Q}_{k+1} \end{bmatrix} \geq 0,$$

$$\hat{\mathbf{A}}_{k+1} = \begin{bmatrix} \mathbf{A} & \mathbf{B}\mathbf{G}_{k+1} \\ \mathbf{H}_{k+1}\mathbf{C} & \mathbf{W}_{k+1} \end{bmatrix},$$

$$(\mathbf{W}_{k+1}, \mathbf{H}_{k+1}, \mathbf{G}_{k+1}) \in \Psi, \quad \mathbf{P}_{k+1} > 0, \quad \mathbf{Q}_{k+1} > 0$$

no

System stable

yes

Configure
WCN

Goal: A WCN configuration that minimizes the impact of disturbances!

Disturbance impact $\hat{y} = \hat{C}\hat{x}[k]$ $\hat{x}[k] = \begin{bmatrix} \mathbf{x}[k] \\ \mathbf{z}[k] \end{bmatrix}$

- Model as a new system:

$$\hat{C} = \begin{bmatrix} \mathbf{I} & \mathbf{0} \end{bmatrix}$$

$$\hat{x}[k + 1] = \hat{A}\hat{x}[k] + \hat{B}u_w[k]$$

$$\hat{y}[k] = \hat{C}\hat{x}[k].$$

**closed-loop:
WCN & plant!**

where the goal is to minimize \hat{y}

How to capture size of discrete time signals?

$$\|\mathbf{v}\|_{\ell_2} \triangleq \left(\sum_{k=0}^{\infty} \|\mathbf{v}[k]\|^2 \right)^{1/2} \quad \|\mathbf{v}\|_{\ell_\infty} \triangleq \sup_{k \geq 0} \|\mathbf{v}[k]\|$$

System gains for the discrete-time system

$$\begin{aligned} \hat{\mathbf{x}}[k+1] &= \hat{\mathbf{A}}\hat{\mathbf{x}}[k] + \hat{\mathbf{B}}\mathbf{u}_w[k] \\ \hat{\mathbf{y}}[k] &= \hat{\mathbf{C}}\hat{\mathbf{x}}[k]. \end{aligned}$$

- **Energy-to-Peak Gain:** $\gamma_{ep} = \sup_{\|\mathbf{u}_w\|_{\ell_2} \leq 1} \|\hat{\mathbf{y}}\|_{\ell_\infty}$
- **Energy-to-Energy Gain:** $\gamma_{ee} = \sup_{\|\mathbf{u}_w\|_{\ell_2} \leq 1} \|\hat{\mathbf{y}}\|_{\ell_2}$

System gains for the discrete-time system

$$\begin{aligned}\hat{\mathbf{x}}[k+1] &= \hat{\mathbf{A}}\hat{\mathbf{x}}[k] + \hat{\mathbf{B}}\mathbf{u}_w[k] \\ \hat{\mathbf{y}}[k] &= \hat{\mathbf{C}}\hat{\mathbf{x}}[k].\end{aligned}$$

- **Energy-to-Peak Gain:** $\gamma_{ep} = \sup_{\|\mathbf{u}_w\|_{\ell_2} \leq 1} \|\hat{\mathbf{y}}\|_{\ell_\infty}$
- **Energy-to-Energy Gain:** $\gamma_{ee} = \sup_{\|\mathbf{u}_w\|_{\ell_2} \leq 1} \|\hat{\mathbf{y}}\|_{\ell_2}$

• **Theorem:**

a) $\gamma_{ep} < \gamma \Leftrightarrow \mathcal{X} \succ 0, \Upsilon \succeq 0$

$$\Upsilon \prec \gamma \mathbf{I} \quad \mathcal{R}(\mathcal{X}, \mathcal{Z}, \Upsilon, \mathcal{X}^{-1}) = \begin{bmatrix} \mathcal{X} & \mathcal{Z} & \hat{\mathbf{A}} & \hat{\mathbf{B}} \\ \mathcal{Z}^T & \Upsilon & \hat{\mathbf{C}} & 0 \\ \hat{\mathbf{A}}^T & \hat{\mathbf{C}}^T & \mathcal{X}^{-1} & 0 \\ \hat{\mathbf{B}}^T & 0 & 0 & \mathbf{I} \end{bmatrix} \succ 0$$

non-convex!

Linearization: $LIN(\mathcal{X}^{-1}, \mathcal{X}_k) = \mathcal{X}_k^{-1} - \mathcal{X}_k^{-1}(\mathcal{X} - \mathcal{X}_k)\mathcal{X}_k^{-1}$

Find feasible points

$$X_0, Z_0, W_0, H_0, G_0$$

Solve the LMI problem,
from X_k find

$$X_{k+1}, \Upsilon_{k+1}, W_{k+1}, H_{k+1}, G_{k+1}$$

no

$$\gamma_{k+1} < \epsilon$$

yes

Configure
WCN

Function Linearization

$$\mathcal{X}_{k+1} = \arg \min_{\mathcal{X}, \mathcal{Z}, \Upsilon, \mathbf{W}, \mathbf{H}, \mathbf{G}, \gamma_{k+1}} \gamma_{k+1}$$

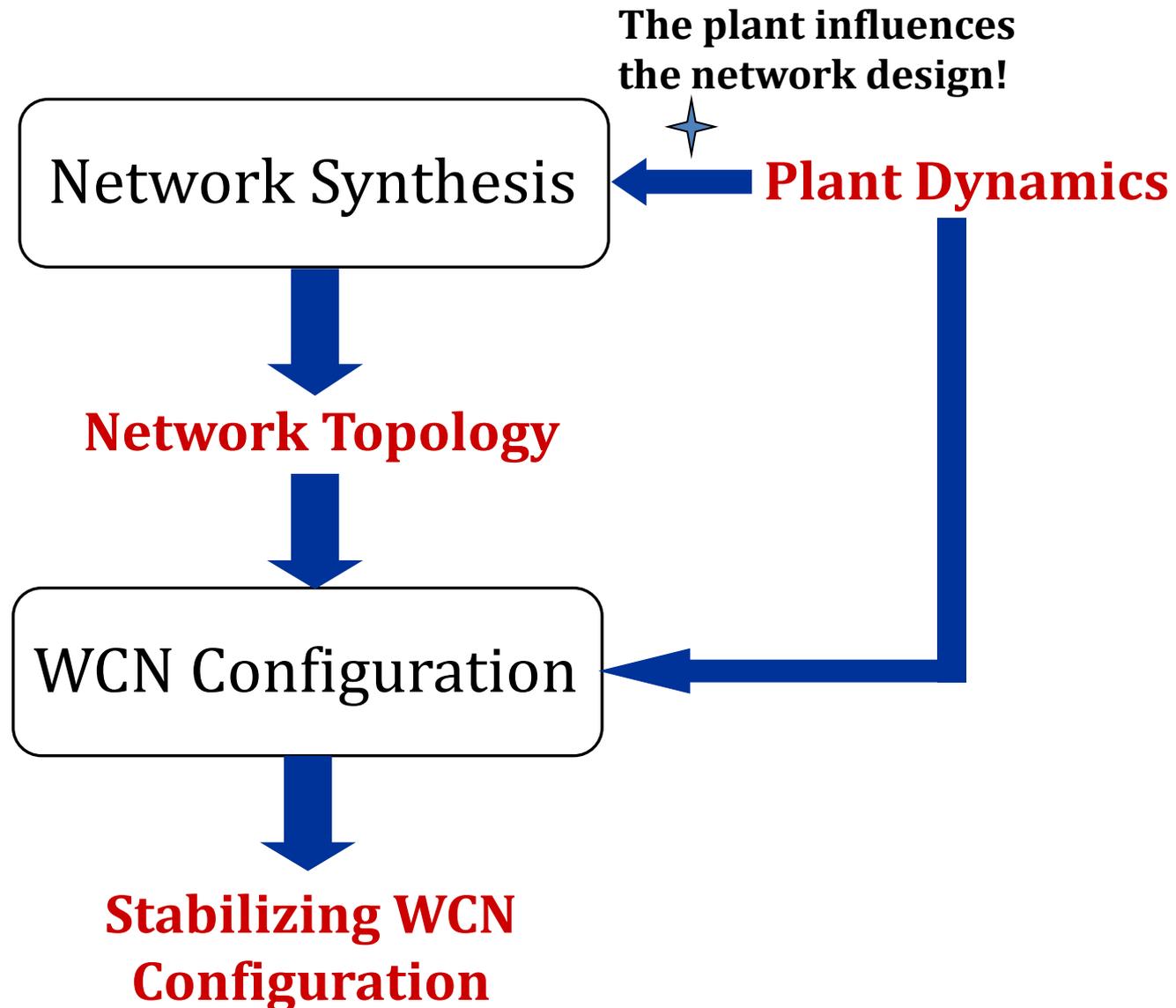
$$\mathcal{R}(\mathcal{X}, \mathcal{Z}, \Upsilon, \text{LIN}(\mathcal{X}^{-1}, \mathcal{X}_k)) \succ 0,$$

$$\Upsilon \prec \gamma_{k+1} \mathbf{I},$$

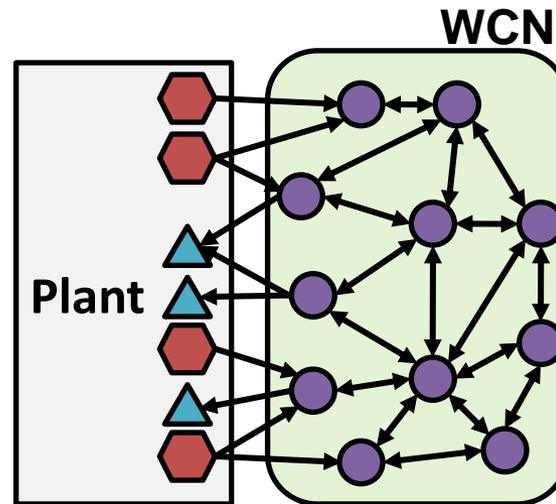
$$(\mathbf{W}, \mathbf{H}, \mathbf{G}) \in \Psi, \mathcal{X} \succ 0, \Upsilon \succeq 0$$

$$\text{For } \gamma_{ee}, \mathcal{Z} = \mathbf{0}$$

$$\mathcal{R}(\mathcal{X}, \mathcal{Z}, \Upsilon, \mathcal{X}^{-1}) = \begin{bmatrix} \mathcal{X} & \mathcal{Z} & \hat{\mathbf{A}} & \hat{\mathbf{B}} \\ \mathcal{Z}^T & \Upsilon & \hat{\mathbf{C}} & \mathbf{0} \\ \hat{\mathbf{A}}^T & \hat{\mathbf{C}}^T & \mathcal{X}^{-1} & \mathbf{0} \\ \hat{\mathbf{B}}^T & \mathbf{0} & \mathbf{0} & \mathbf{I} \end{bmatrix} \succ 0$$

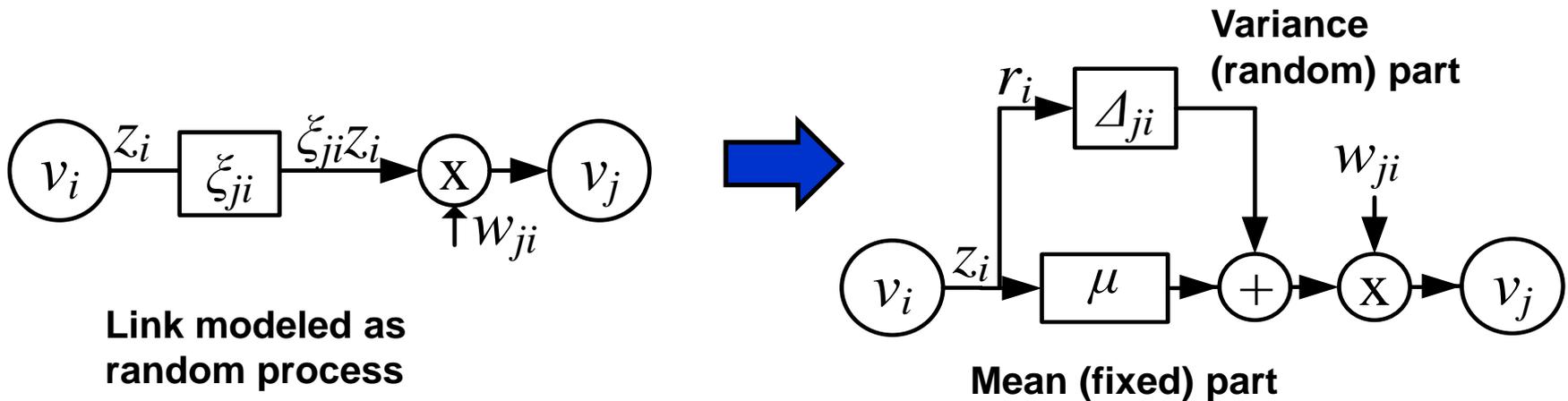


- Wireless control network



- Modeling
- Controller synthesis
- **Robustness & security**

- What happens if links in the network fail?
 - Bernoulli distribution: fails with some probability
- Many links in network: how to model concisely?
 - Use robust control [Elia, Sys & Control Letters, '05]

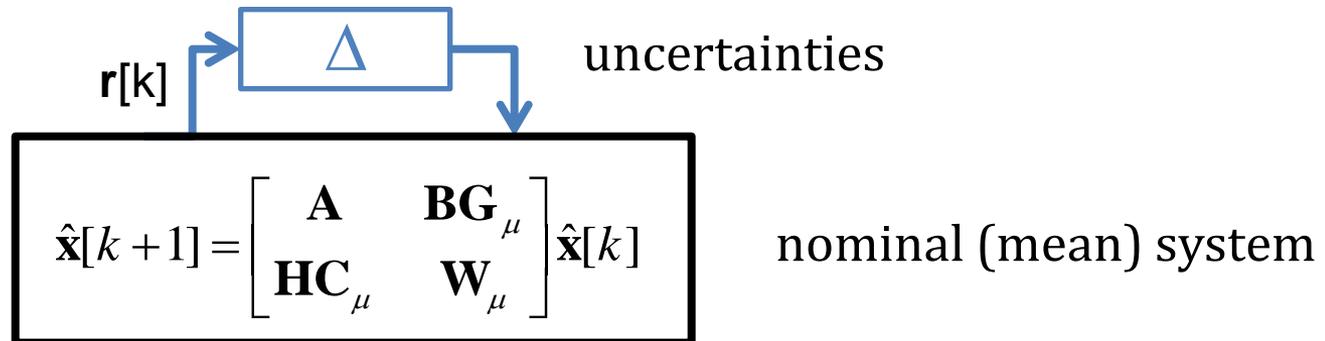


– Received value: $\xi_{ji}[k]z_i[k] = (\mu + \Delta_{ji}[k])z_i[k]$

random
variable

mean
(constant)

zero-mean
random variable



- Closed loop system with uncertainties:

$$\hat{\mathbf{x}}[k+1] = \begin{bmatrix} \mathbf{A} & \mathbf{B}\mathbf{G}_\mu \\ \mathbf{H}_\mu \mathbf{C} & \mathbf{W}_\mu \end{bmatrix} \hat{\mathbf{x}}[k] + \mathbf{J}\Delta[k]\mathbf{r}[k]$$

Mean (fixed) part

Random part

$$\mathbf{r}[k] = \hat{\mathbf{J}}^{\text{or}} \hat{\mathbf{x}}[k]$$

- Closed loop system with random Bernoulli failures

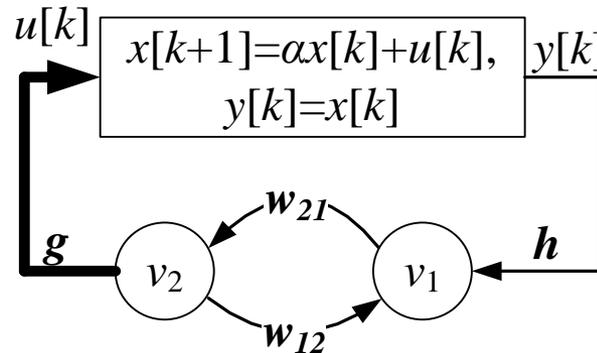
$$\hat{\mathbf{x}}[k+1] = \begin{bmatrix} \mathbf{A} & \mathbf{B}\mathbf{G}_\mu \\ \mathbf{H}_\mu\mathbf{C} & \mathbf{W}_\mu \end{bmatrix} \hat{\mathbf{x}}[k] + \mathbf{J}\Delta[k]\mathbf{r}[k]$$

System is mean square stable if and only if there exists \mathbf{X} , $\alpha_1, \alpha_2, \dots, \alpha_N$ such that

$$\mathbf{X} \succ \hat{\mathbf{A}}_\mu \mathbf{X} \hat{\mathbf{A}}_\mu^T + \mathbf{J} \text{diag}\{\alpha\} (\mathbf{J})^T$$
$$\alpha_i \geq \sigma_i^2 (\hat{\mathbf{J}}^{or})_i \mathbf{X} (\hat{\mathbf{J}}^{or})_i^T, \quad \forall i \in \{1, \dots, N_l\}$$

- Robustness requires
 - One additional constraint added for each link (Bernoulli failures)
 - More constraints for more general failure models
 - Significant improvements with observer style updates

- Example



For $\alpha=2$, maximal message drop probability which guarantees MSS

$$p_{max} \leq 1.18\%$$

How can we improve robustness of the WCN to link failures?

- Idea: Include **observer style updates**
 - different weights depending of the success of the transmission

$$z_j[k + 1] = (w_{jj} - \sum_{i \in \mathcal{N}_{v_j}} \xi_{ji} q_{ji}) z_j[k] + \sum_{i \in \mathcal{N}_{v_j}} \xi_{ji} w_{ji} z_i[k]$$

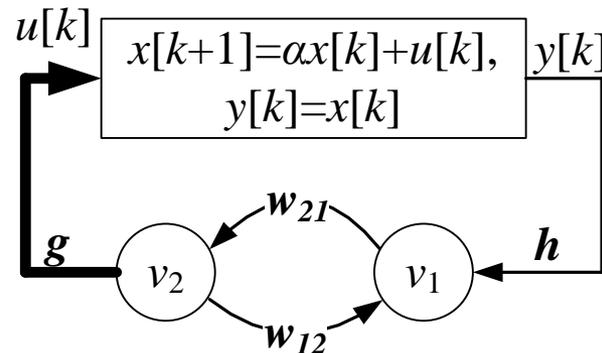
Observer Style Updates – for reliable communication links

$$z_j[k + 1] = w_{jj} z_j[k] + \sum_{i \in \mathcal{N}_{v_j}} (w_{ji} z_i[k] - q_{ji} z_j[k])$$

Standard observer

A similar **design-time** iterative algorithm can be used to extract robust WCN configurations!

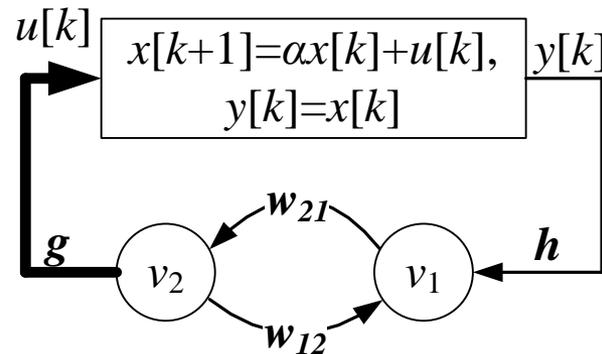
- Example



- Maximal message drop probability which guarantees MSS, $\alpha=2$

	<i>WCN</i> (scalar state)	<i>WCN</i> (\mathbb{R}^2 state)	<i>oWCN</i> (scalar state)	<i>oWCN</i> (\mathbb{R}^2 state)
$N = 2$	$p_m = 1.18\%$	$p_m = 1.30\%$	$p_m = 10.46\%$	$p_m = 17.82\%$
$N = 3$	$p_m = 1.32\%$	$p_m = 1.46\%$	$p_m = 11.24\%$	$p_m = 17.88\%$
$N = 4$	$p_m = 1.41\%$	$p_m = 1.54\%$	$p_m = 11.46\%$	$p_m = 17.88\%$
		<i>oWCN</i> (\mathbb{R}^3 state)	<i>oWCN</i> (\mathbb{R}^4 state)	<i>oWCN</i> (\mathbb{R}^5 state)
$N = 2$		$p_m = 20.40\%$	$p_m = 20.48\%$	$p_m = 20.64\%$

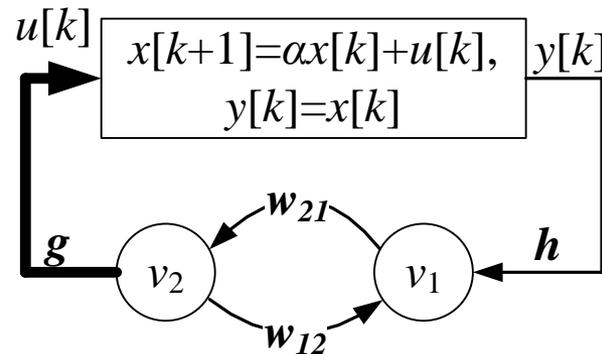
- Example



- Maximal message drop probability which guarantees MSS, $\alpha=2$

	WCN (scalar state)	WCN (\mathbb{R}^2 state)	$oWCN$ (scalar state)	$oWCN$ (\mathbb{R}^2 state)
$N = 2$	$p_m = 1.18\%$	$p_m = 1.30\%$	$p_m = 10.46\%$	$p_m = 17.82\%$
$N = 3$	$p_m = 1.32\%$	$p_m = 1.46\%$	$p_m = 11.24\%$	$p_m = 17.88\%$
$N = 4$	$p_m = 1.41\%$	$p_m = 1.54\%$	$p_m = 11.46\%$	$p_m = 17.88\%$
		$oWCN$ (\mathbb{R}^3 state)	$oWCN$ (\mathbb{R}^4 state)	$oWCN$ (\mathbb{R}^5 state)
$N = 2$		$p_m = 20.40\%$	$p_m = 20.48\%$	$p_m = 20.64\%$

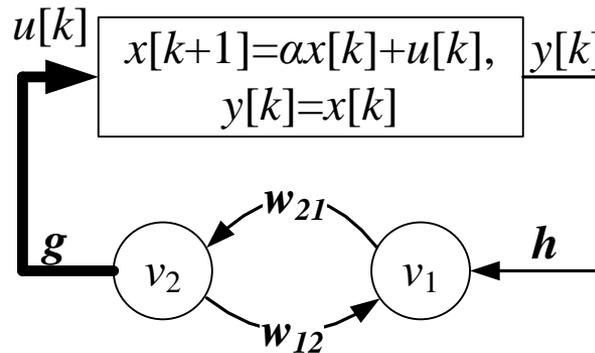
- Example



- Maximal message drop probability which guarantees MSS, $\alpha=2$

	<i>WCN</i> (scalar state)	<i>WCN</i> (\mathbb{R}^2 state)	<i>oWCN</i> (scalar state)	<i>oWCN</i> (\mathbb{R}^2 state)
$N = 2$	$p_m = 1.18\%$	$p_m = 1.30\%$	$p_m = 10.46\%$	$p_m = 17.82\%$
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		<i>oWCN</i> (\mathbb{R}^3 state)	<i>oWCN</i> (\mathbb{R}^4 state)	<i>oWCN</i> (\mathbb{R}^5 state)
$N = 2$		$p_m = 20.40\%$	$p_m = 20.48\%$	$p_m = 20.64\%$

- Example – WCN with observer style updates



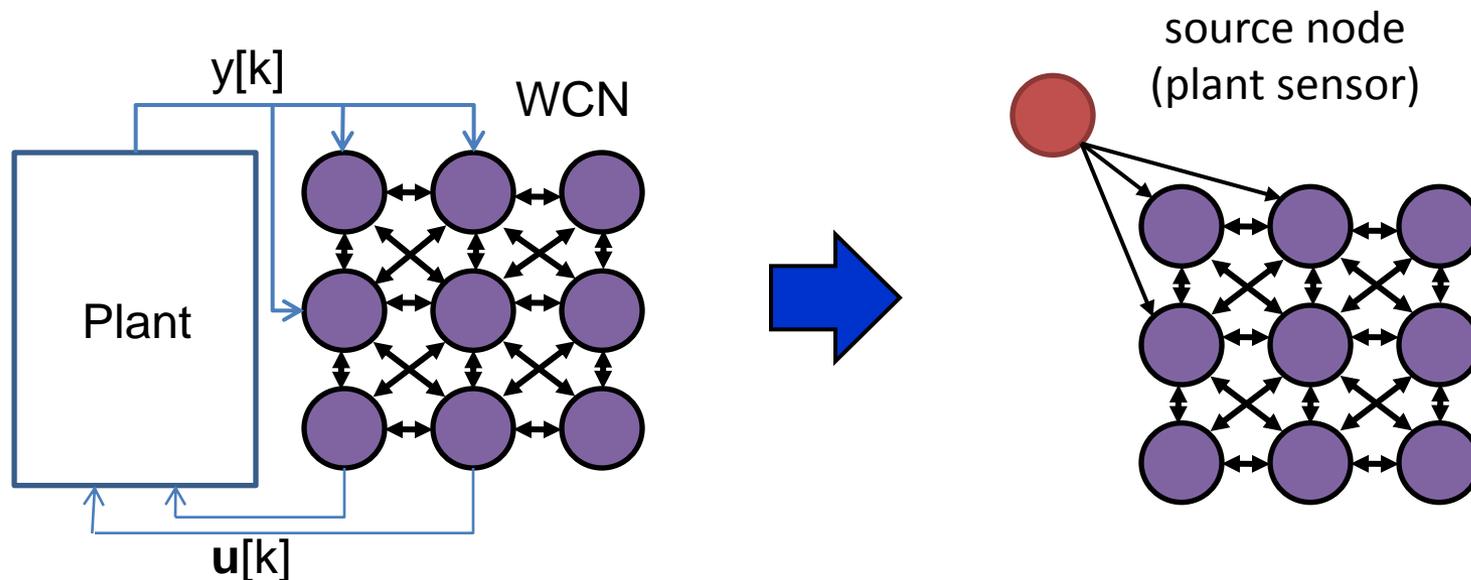
For $\alpha=2$, maximal message drop probability which guarantees MSS

$$p_{max} \approx 21\% < 25\%$$

Approaching theoretical limit for robustness with centralized controllers!

- What if certain nodes in the WCN become faulty or malicious?
- Security of control networks in industrial control systems is a major issue [NIST Technical Report, 2008]
 - Data Historian: Maintain and analyze logs of plant and network behavior
 - Intrusion Detection System: Detect and identify any abnormal activities
- Is it possible to design an Intrusion Detection System to determine if any nodes are not following WCN protocol?
- Can IDS scheme avoid listening all nodes? Under what conditions? Which nodes?

- Consider graph of wireless control network with plant sensors



- Denote transmissions of any set T of monitored nodes by

$$\mathbf{t}[k] = \mathbf{Tz}[k]$$

- T is a matrix with a single 1 in each row, indicating which nodes $z[k]$ are being monitored

- WCN model with set S of faulty/malicious nodes:

$$\mathbf{z}[k + 1] = \mathbf{W}\mathbf{z}[k] + \mathbf{H}\mathbf{y}[k] + \mathbf{B}_S\mathbf{f}_S[k]$$

$$\mathbf{t}[k] = \mathbf{T}\mathbf{z}[k]$$

- Objective: Recover $\mathbf{y}[k]$, $\mathbf{f}_S[k]$ and S (initial state $\mathbf{z}[0]$ known)
 - Almost equivalent to invertibility of system
- Problem: Don't know the set of faulty nodes S
 - Assumption: At most b faulty/malicious nodes
- Approach: Must ensure that output sequence cannot be generated by a different $\mathbf{y}[k]$ and possibly different set of b malicious nodes

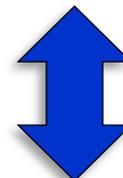
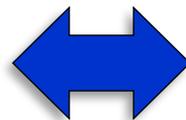
IDS can recover $\mathbf{y}[k]$ and identify up to b faulty nodes in the network by monitoring transmissions of set T

Can recover inputs and set S in system

$$\mathbf{z}[k+1] = \mathbf{W}\mathbf{z}[k] + \begin{bmatrix} \mathbf{H} & \mathbf{B}_S \end{bmatrix} \begin{bmatrix} \mathbf{y}[k] \\ \mathbf{f}_S[k] \end{bmatrix}$$

$$\mathbf{t}[k] = \mathbf{T}\mathbf{z}[k]$$

for **any unknown set** S of b nodes



Linear system

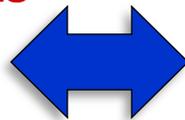
$$\mathbf{z}[k+1] = \mathbf{W}\mathbf{z}[k] + \begin{bmatrix} \mathbf{H} & \mathbf{B}_Q \end{bmatrix} \begin{bmatrix} \mathbf{y}[k] \\ \mathbf{f}_Q[k] \end{bmatrix}$$

$$\mathbf{t}[k] = \mathbf{T}\mathbf{z}[k]$$

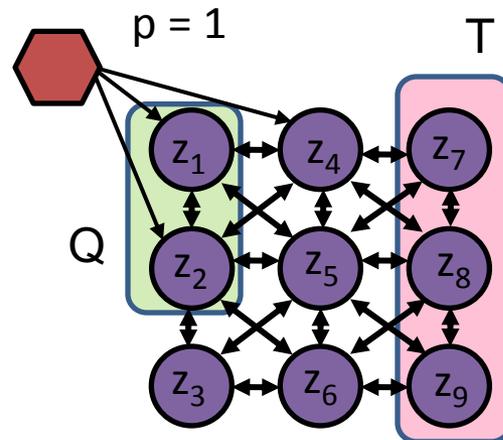
is invertible for **any known set** Q of $2b$ nodes

There are $p+2b$ node-disjoint paths from sensors and any set Q of $2b$ nodes to monitoring set T

(generically)

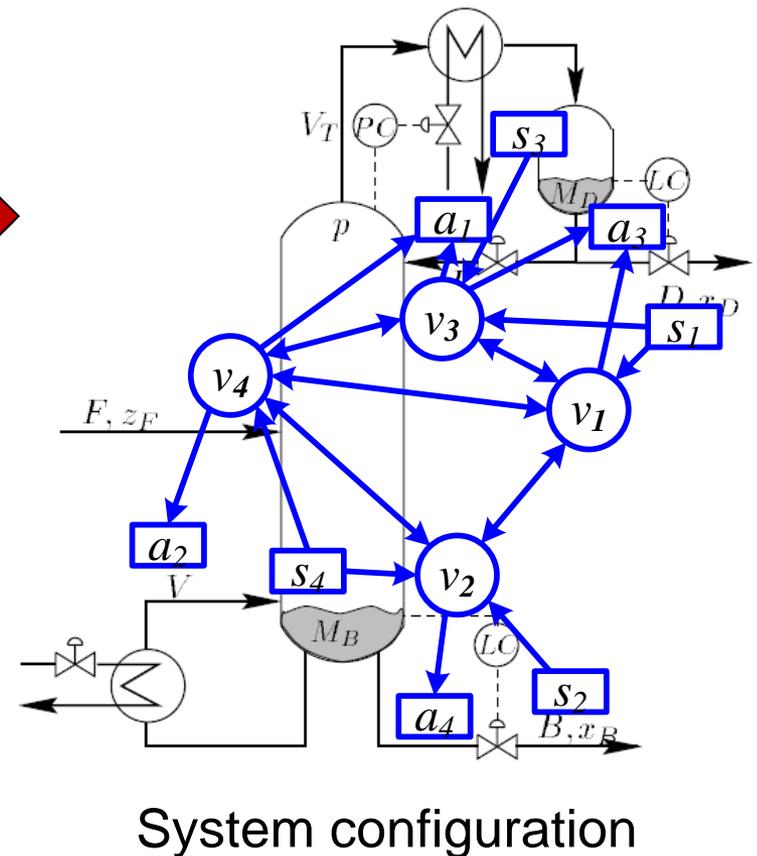
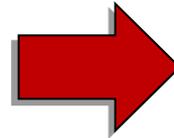
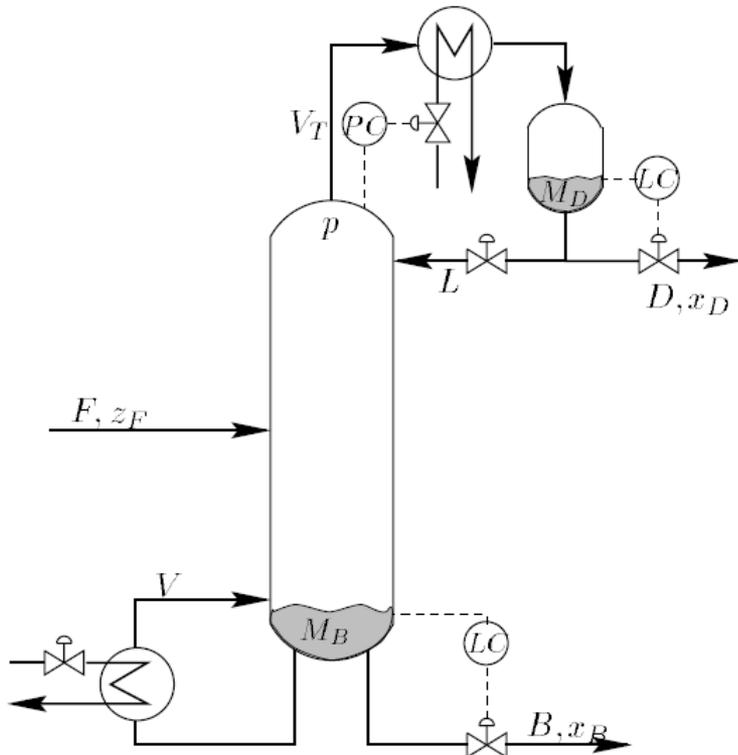


- Suppose we want to identify $b = 1$ faulty/malicious node and recover the plant outputs in this setting:



- Consider set $Q = \{v_1, v_2\}$
 - $p+2b$ vertex disjoint paths from sensor and Q to T
- Can verify that this holds for any set Q of $2b$ nodes
- Sufficient condition: Network is $p+2b$ connected

- Distillation column control
 - Plant **continuous-time** model contains 8 states, 4 inputs, 4 outputs
- Distillation column structure



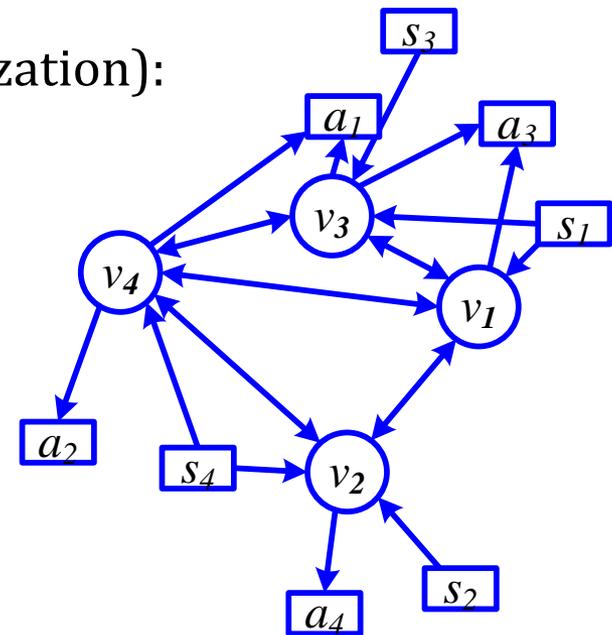
- Distillation column control
 - Plant model contains 8 states, 4 inputs, 4 outputs
- WCN contains 4 nodes

Stable configuration (obtained after plant discretization):

$$\text{node} \rightarrow \text{node} \quad \mathbf{W} = \begin{bmatrix} -0.470 & 0.339 & -0.260 & -0.390 \\ -1.117 & -0.145 & 0 & -0.269 \\ 0.0514 & 0 & -0.703 & 0.600 \\ 0.854 & 0.277 & -0.086 & -0.112 \end{bmatrix}$$

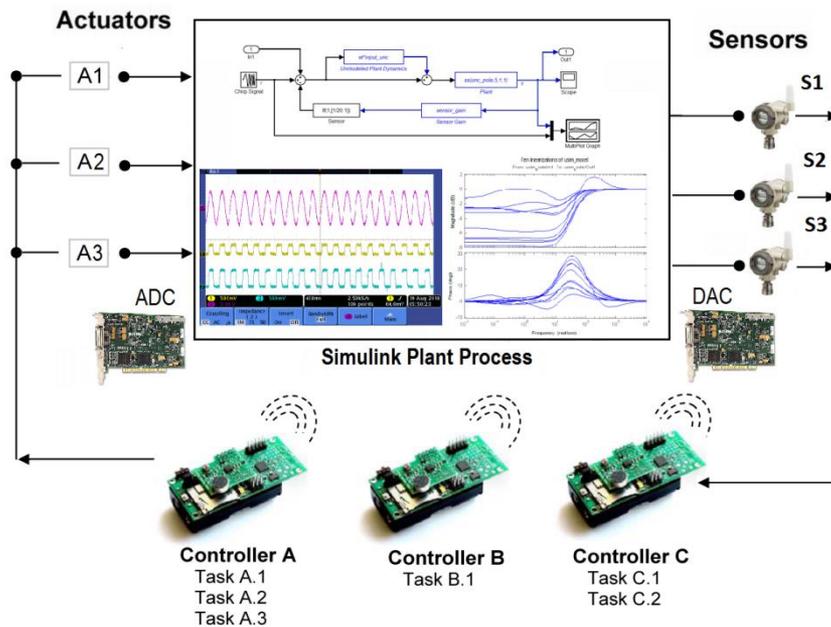
$$\text{sensor} \rightarrow \text{node} \quad \mathbf{H} = \begin{bmatrix} 1.260 & 0 & 0 & 0 \\ 0 & 0.104 & 0 & 0.075 \\ 0 & 0 & 0.421 & 0 \\ 0 & 0 & 0 & -0.034 \end{bmatrix}$$

$$\text{node} \rightarrow \text{actuator} \quad \mathbf{G} = \begin{bmatrix} 0 & 0 & -0.226 & -0.459 \\ 0 & 0 & 0 & 0.102 \\ 0.120 & 0 & 1.072 & 0 \\ 0 & 2.549 & 0 & 0 \end{bmatrix}$$

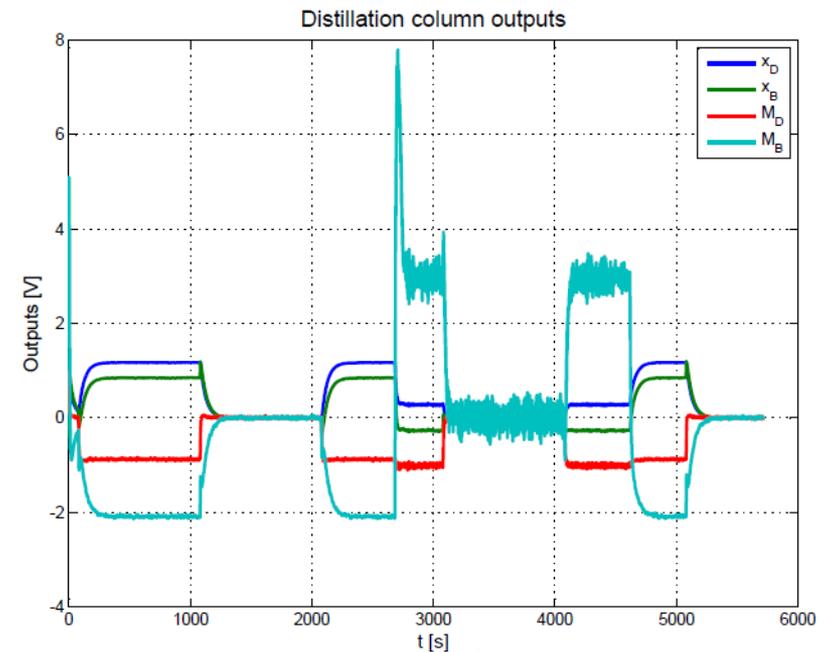


Network topology

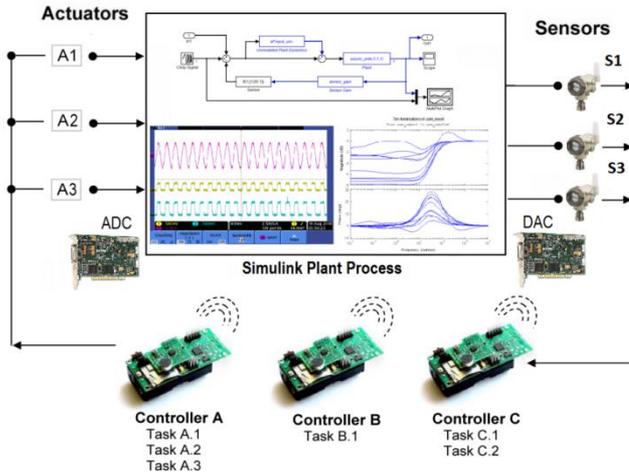
Process-in-the-loop test-bed



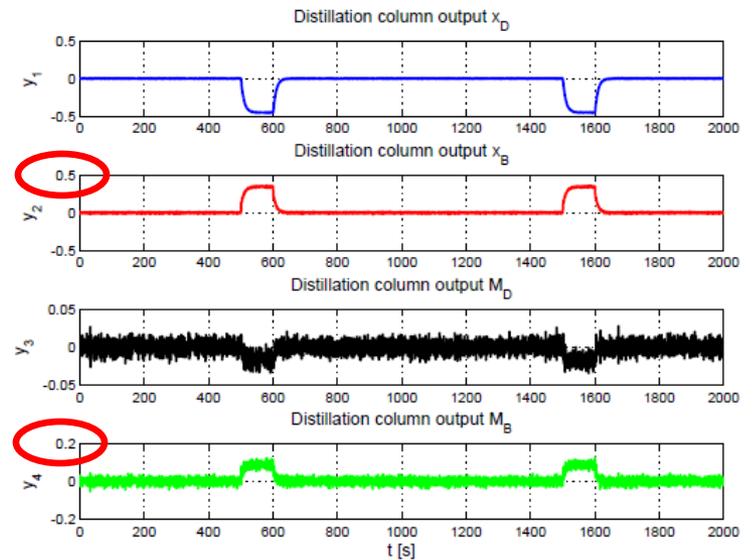
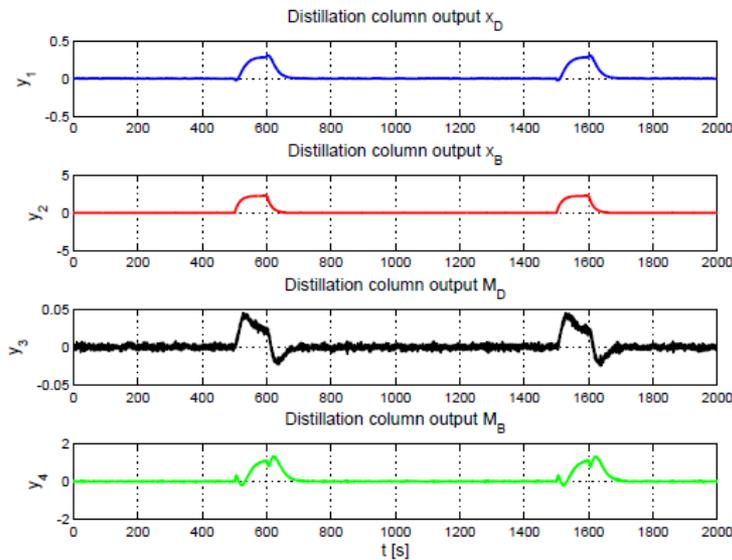
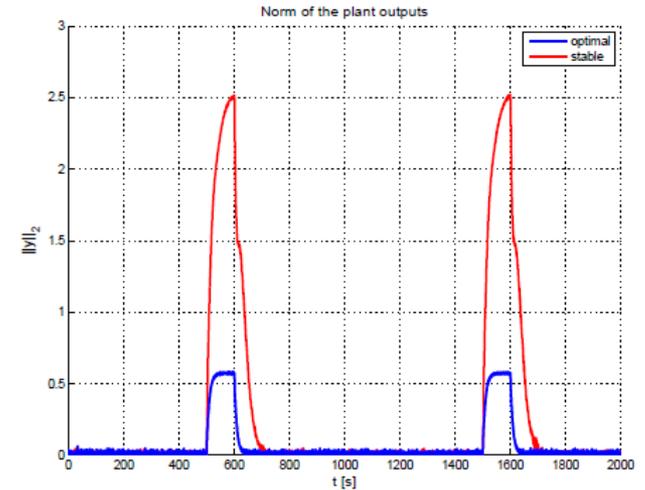
Scenario I: v_1 turned OFF/ON



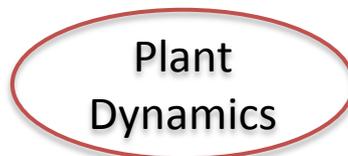
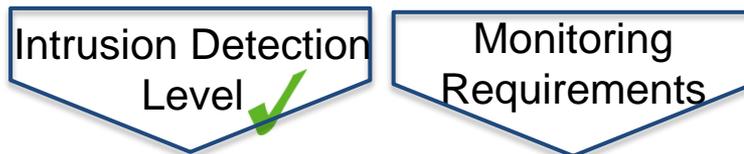
Process-in-the-loop test-bed



Scenario II: Optimal control



[CDC'10]



[JSAC'13, TAC'11,
CDC'11, CDC'10]

Network
Topology

Communication
schedule

[IPSN'12]



[ACC'13]



Many thanks for your attention!

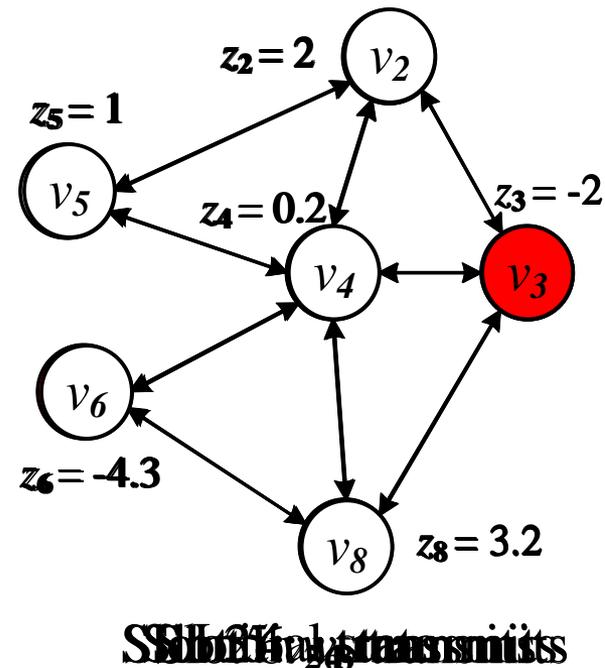
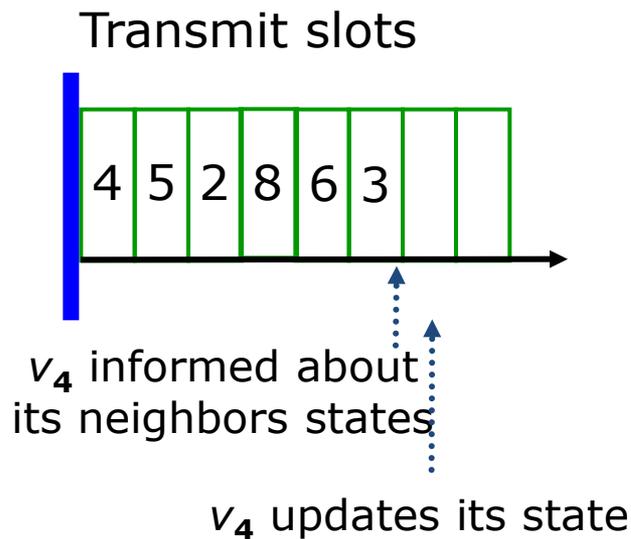


PRECISE
PENN RESEARCH IN EMBEDDED COMPUTING AND INTEGRATED SYSTEMS ENGINEERING

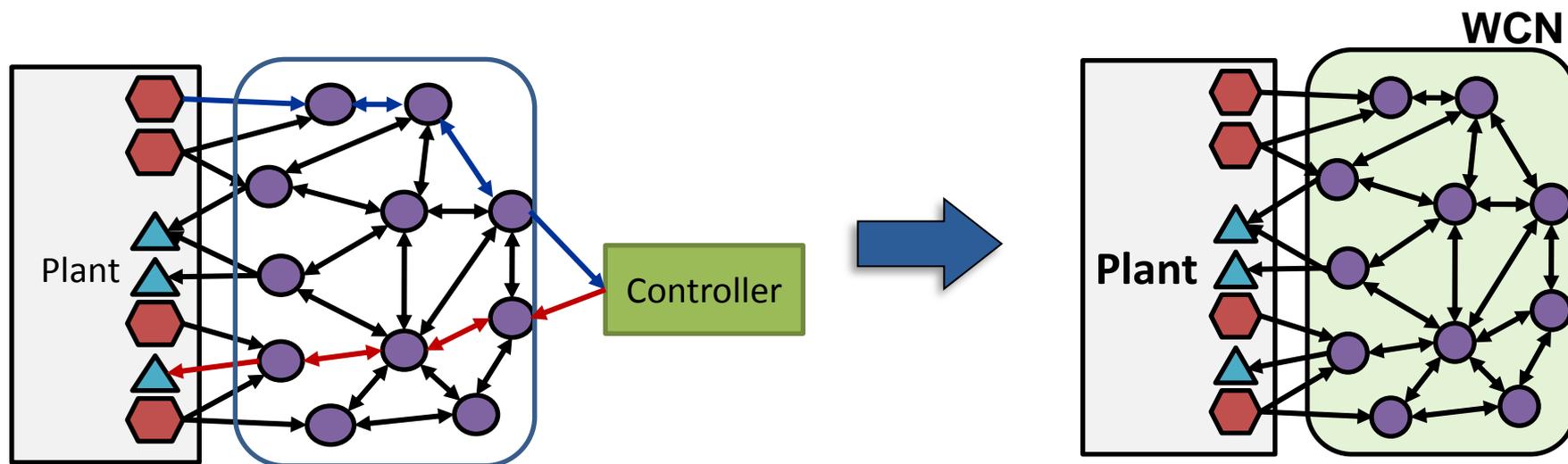
ABB
Honeywell



- Each node maintains its (possible vector) state
 - Transmits state exactly once in each step (per frame)
 - Updates own state using linear iterative strategy
- Example:



- In multi-hop control, nodes route information to controller



- Can we leverage computation of the network?
- Can we distribute the controller to nodes of the network?
- Reminiscent of network coding